FEATURES

- Low Cost
- Rail-to-Rail Input and Output
  0.8mV Typical $V_{OS}$
- Unity Gain Stable
- Gain-Bandwidth Product: 3MHz
- Very Low Input Bias Current: 10pA
- Supply Voltage Range: 2.1V to 5.5V
- Input Voltage Range:
  -0.1V to +5.6V with $(V_{DD} - V_{SS}) = 5.5V$
- Low Supply Current: 200μA/Amplifier
- Small Packaging
  MD1621 Available in SOT-23-5
  MD1622 Available in SOIC-8

APPLICATIONS

- ASIC Input or Output Amplifier
- Sensor Interface
- Piezoelectric Transducer Amplifier
- Medical Instrumentation
- Mobile Communication
- Audio Output
- Portable Systems
- Smoke Detectors
- Notebook PC
- PCMCIA Cards
- Battery-Powered Equipment
- DSP Interface

PRODUCT DESCRIPTION

The MD1621 (single), MD1622 (dual) are low cost, rail-to-rail input and output voltage feedback amplifiers. They have a wide input common mode voltage range and output voltage swing, and take the minimum operating supply voltage down to 2.1V. The maximum recommended supply voltage is 5.5V. It is specified over the extended -40°C to +85°C temperature range.

The MD1621/2 provide 3MHz bandwidth at a low current consumption of 200μA per amplifier. Very low input bias currents of 10pA enable MD1621/2 to be used for integrators, photodiode amplifiers, and piezoelectric sensors. Rail-to-rail input and output are useful to designers for buffering ASIC in single-supply systems.

Applications for this series of amplifiers include safety monitoring, portable equipment, battery and power supply control, and signal conditioning and interfacing for transducers in very low power systems.

The MD1621 is available in the Green SOT-23-5 Package. The MD1622 comes in the Green SOIC-8 package.

PIN CONFIGURATIONS (TOP VIEW)
MD1621/2

ORDER INFORMATION

<table>
<thead>
<tr>
<th>MODEL</th>
<th>ORDER NUMBER</th>
<th>PACKAGE DESCRIPTION</th>
<th>PACKAGE OPTION</th>
<th>MARKING INFORMATION</th>
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<tbody>
<tr>
<td>MD1621</td>
<td>--</td>
<td>SOT23-5</td>
<td>Tape and Reel, 2500</td>
<td>--</td>
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<tr>
<td>MD1622</td>
<td>--</td>
<td>SOIC-8</td>
<td>Tape and Reel, 2500</td>
<td>--</td>
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</tbody>
</table>

ABSOLUTE MAXIMUM RATINGS

Supply Voltage, \( V_{DD} \) to \( V_{SS} \)...............................................................6V
Common Mode Input Voltage.................\( V_{SS} \) - 0.3V to \( V_{DD} \) + 0.3V
Storage Temperature Range.........................-65℃ to +150℃
Junction Temperature.................................................................150℃
Operating Temperature Range..............................-40℃ to +85℃

Package Thermal Resistance @ \( T_A = +25\)℃
SOIC-8, 0JA.................................................................125℃/W
Lead Temperature (Soldering 10sec)....................260℃

NOTE:
Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ESD, Electrostatic Discharge Protection

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Condition</th>
<th>Minimum Level</th>
<th>Unit</th>
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<tr>
<td>HBM</td>
<td>Human Body Model ESD</td>
<td></td>
<td>4</td>
<td>kV</td>
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<tr>
<td>MM</td>
<td>Machine Model ESD</td>
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<td>300</td>
<td>V</td>
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### ELECTRICAL CHARACTERISTICS

The • denotes the specifications which apply over the full operating temperature range, otherwise specifications are At T_A=25°C, V_{DD} = +5V, V_{SS} = GND, R_L = 100kΩ connected to V_{DD}/2, and V_{OUT} = V_{DD}/2.

<table>
<thead>
<tr>
<th>PARAMETER</th>
<th>SYMBOL</th>
<th>CONDITIONS</th>
<th>MIN</th>
<th>TYP</th>
<th>MAX</th>
<th>UNITS</th>
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<tr>
<td>INPUT CHARACTERISTICS</td>
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<tr>
<td>Input Offset Voltage</td>
<td>V_{OS}</td>
<td>V_{CM} = V_{DD}/2</td>
<td>0.8</td>
<td>5</td>
<td>6.6</td>
<td>mV</td>
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<tr>
<td>Input Offset Voltage</td>
<td></td>
<td>V_{CM} = V_{DD}/2</td>
<td>•</td>
<td></td>
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<td>Input Bias Current</td>
<td>I_B</td>
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<td>10</td>
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<td></td>
<td>pA</td>
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<td>10</td>
<td></td>
<td></td>
<td>pA</td>
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<tr>
<td>Input Offset Voltage Drift</td>
<td>ΔV_{OS}/ΔT</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>μV/°C</td>
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<td>Input Common Mode Voltage Range</td>
<td>V_{CM}</td>
<td>V_{DD} = 5.5V</td>
<td>-0.1-5.6</td>
<td></td>
<td></td>
<td>V</td>
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<td>Common Mode Rejection Ratio</td>
<td>CMRR</td>
<td>V_{DD} = 5.5V, V_{CM} = -0.1V to 4V</td>
<td>76</td>
<td></td>
<td></td>
<td>dB</td>
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<td>Open-Loop Voltage Gain</td>
<td>A_{OL}</td>
<td>R_L = 100kΩ, V_{OUT} = 2.5V</td>
<td>100</td>
<td></td>
<td></td>
<td>dB</td>
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<td></td>
<td>R_L = 100kΩ, V_{OUT} = +0.2V to +4.8V</td>
<td>98</td>
<td></td>
<td></td>
<td>dB</td>
</tr>
</tbody>
</table>

| OUTPUT CHARACTERISTICS | | | | | | |
| Output Voltage Swing from Rail | | | | | | |
| V_{OH} | R_L = 100kΩ | 4.980 | 4.995 | | V |
| V_{OL} | R_L = 100kΩ | 25 | 5 | | mV |
| V_{OH} | R_L = 10kΩ | 4.970 | 4.994 | | V |
| V_{OL} | R_L = 10kΩ | 35 | 6 | | mV |
| Output Current | I_{SOURCE} | R_L = 10Ω to V_{DD}/2 | 40 | | | mA |
| | I_{SINK} | | 40 | | | |

| POWER SUPPLY | | | | | | |
| Operating Voltage Range | | | 2.1 | 2.5 | 5.5 | V |
| Power Supply Rejection Ratio | PSRR | V_{DD} = +2.5V to +5.5V, V_{CM} = +0.5V | 85 | | | dB |
| Quiescent Current/Amplifier | I_Q | | 200 | | | μA |
| | | • | 120 | 270 | | |

| DYNAMIC PERFORMANCE (C_L = 100pF) | | | | | | |
| Gain-Bandwidth Product | GBP | | 3 | | | MHz |
| Phase Margin | PM | R_L = 100kΩ, C_L = 100pF | 45 | | | ° |
| Harmonic Distortion | HD2 | f = 10kHz, G = +1, R_L=100kΩ,V_{OUT}=2V_{PP} | >80 | | | dBc |
| | HD3 | f = 10kHz, G = +1, R_L=100kΩ,V_{OUT}=2V_{PP} | >80 | | | |
| Slew Rate | SR | G = +1, 2V Output Step | 1.74 | | | V/μs |
| Settling Time to 0.1% | t_s | G = +1, 2V Output Step | 6 | | | μs |
| Overload Recovery Time | V_{IN}/G = V_{DD} | | 2.5 | | | μs |

| NOISE PERFORMANCE | | | | | | |
| Voltage Noise Density | e_n | f = 1kHz | 20 | | | nV/√Hz |
| | | f = 10kHz | 12 | | | nV/√Hz |
TYPICAL PERFORMANCE CHARACTERISTICS

At $T_A = +25^\circ C$, $V_{DD} = +5V$, $V_{SS} = GND$, and $R_L = 100k\Omega$ connected to $V_{DD}/2$, unless otherwise specified.

Figure 1. Supply Current vs. Supply Voltage

Figure 2. Input Offset Voltage Distribution

Figure 3. Output Short Circuit Current vs. Supply Voltage

Figure 4. Open-Loop Gain vs. Temperature

Figure 5. CMRR vs. Temperature

Figure 6. PSRR vs. Temperature
TYPICAL PERFORMANCE CHARACTERISTICS

At TA = +25°C, VDD = +5V, VSS = GND, and RL = 100kΩ connected to VDD/2, unless otherwise specified.

Figure 7. Small-Signal Overshoot vs. Load Capacitance

Figure 8. Small-Signal Overshoot vs. Load Capacitance

Figure 9. Output Voltage vs. Output Current

Figure 10. Output Voltage vs. Output Current

Figure 11. Input Voltage Noise Spectral Density vs. Frequency

Figure 12. Maximum Output Voltage vs. Frequency
TYPICAL PERFORMANCE CHARACTERISTICS

At $T_A = +25^\circ C$, $V_{DD} = +5V$, $V_{SS} = \text{GND}$, and $R_L = 100\,\Omega$ connected to $V_{DD}/2$, unless otherwise specified.

Figure 13. Positive Overload Recovery Time

Figure 14. Negative Overload Recovery Time

Figure 15. Phase Reversal

Figure 16. Large-Signal Step Response

Figure 17. Small-Signal Step Response

Figure 18. Small-Signal Step Response
TYPICAL PERFORMANCE CHARACTERISTICS

At $T_A = +25^\circ C$, $V_{DD} = +5V$, $V_{SS} = GND$, and $R_L = 100k\Omega$ connected to $V_{DD}/2$, unless otherwise specified.

Figure 19. Gain and Phase vs. Frequency

Figure 20. CMRR and PSRR vs. Frequency

Figure 21. Channel Separation vs. Frequency

Figure 22. $Z_{ol}$ vs. Frequency
MD1621/2 are CMOS, rail-to-rail input and output voltage feedback amplifiers designed for general purpose applications.

Operating Voltage
The MD1621/2 are specified over a power-supply range of +2.1V to +5.5V (±1.05V to ±2.75V), Supply voltages higher than 6V (absolute maximum) can permanently damage the amplifier.

Parameters that vary over supply voltage or temperature are shown in the typical characteristics section of this datasheet.

Rail-to-Rail Input
The input stage of the amplifiers is a true rail-to-rail architecture, allowing the input common-mode voltage range of the op amp to extend to both positive and negative supply rails. This maximizes the usable voltage range of the amplifier, an important feature for single-supply and low voltage applications. This rail-to-rail input range is achieved with a complementary input stage—an NMOS input differential pair in parallel with a PMOS differential pair. The NMOS pair is active at the upper end of the common-mode voltage range, typically VDD – 1.2V to 100mV above the positive supply, while the PMOS pair is active for inputs from 100mV below the negative supply to approximately VDD – 1.2V.

Rail-to-Rail Output
A class AB output stage with common-source transistors is used to achieve rail-to-rail output. The maximum output voltage swing is proportional to the output current, and larger currents will limit how close the output voltage can get to the proximity of the output voltage to the supply rail. This is a characteristic of all rail-to-rail output amplifiers. See the typical performance characteristic Figure 9, Output Voltage Swing vs. Output Current.

Capacitive Loads
The MD1621/2 op amps can directly drive large capacitive loads. As the load capacitance increases, the feedback loop’s phase margin decreases and the closed-loop’s bandwidth is reduced. This produces gain peaking in the frequency response, with overshoot and ringing in the step response. While a op amp in unity gain configuration (G = +1 V/V) is most susceptible to the effects of capacitive loading.

When driving large capacitive loads with the MD1621/2 amplifiers (e.g., > 100pF when G = +1 V/V), a small series resistor at the output (RISO in Figure 23) improves the feedback loop’s phase margin (stability) by making the output load resistive at higher frequencies.

PCB Surface Leakage
In Applications where low input bias current is critical, PC board surface leakage effects need to be considered. Surface leakage is caused by humidity, dust or other contamination on the board. Under low humidity conditions, a typical resistance between nearby traces is $10^{12} \Omega$. A 5V difference would cause 5pA of current to flow; which is similar to the MD1621/2 op amps’ bias current at +25°C (±10pA, typical).

The best way to reduce surface leakage is to use a guard ring around sensitive pins (or traces). The guard ring is biased at the same voltage as the sensitive pin. An example of this type of layout is shown in Figure 24.

1. Non-inverting Gain and Unity-Gain Buffer:
   a) Connect the non-inverting pin (VIN+) to the input with a wire that does not touch the PCB surface.
   b) Connect the guard ring to the inverting input pin (VIN-). This biases the guard ring to the Common Mode input voltage.

2. Inverting Gain and Transimpedance Gain Amplifiers (convert current to voltage, such as photo detectors):
   a) Connect the guard ring to the non-inverting input pin (VIN+). This biases the guard ring to the reference voltage as the op amp (e.g., VDD/2 or ground).
   b) Connect the inverting pin (VIN-) to the input with a wire that does not touch the PCB surface.
TYPICAL APPLICATION

Differential Amplifier
The circuit shown in Figure 25 performs the difference function. If the resistor ratios are equal to \((R_4 / R_3 = R_2 / R_1)\), then \(V_{OUT} = (V_P - V_N) \times R_2 / R_1 + V_{REF}\).

![Differential Amplifier](image)

Figure 25. Differential Amplifier

Photodiode Application
The MD1621/2 have very high impedance with an input bias current typically around 10 pA. This characteristic allows the MD1621/2 op amp to be used in photodiode applications and other applications that require high input impedance. Note that the MD1621/2 have significant voltage offset that can be removed by capacitive coupling or software calibration.

Figure 26 illustrates a photodiode or current measurement application. The feedback resistor is limited to 10 MΩ to avoid excessive output offset. In addition, a resistor is not needed on the noninverting input to cancel bias current offset because the bias current-related output offset is not significant when compared to the voltage offset contribution. For best performance, follow the standard high impedance layout techniques, which include the following:

- Shielding the circuit.
- Cleaning the circuit board.
- Putting a trace connected to the noninverting input around the inverting input.
- Using separate analog and digital power supplies.

![Photodiode Application](image)

Figure 26. High Input Impedance Application—Photodiode Amplifier
3MHz, 200μA, Rail-to-Rail I/O
CMOS Operational Amplifiers

PACKAGE OUTLINE DIMENSIONS

COMPLIANT TO JEDEC STANDARD MS-012-AA
Figure 27 8-Lead Small Outline Package [SOIC]
Dimensions shown in millimeters
3MHz, 200μA, Rail-to-Rail I/O
CMOS Operational Amplifiers

PACKAGE OUTLINE DIMENSIONS

COMPLIANT TO JEDEC STANDARD MO-178-AA

Figure 28 5-Lead Small Outline Transistor Package [SOT-23]
Dimensions shown in millimeters