

# 29152/29302/29502/29752

## High-Current Low-Dropout Regulators

### General Description

The 29150/29300/29500/29750 are high current, high accuracy, low-dropout voltage regulators. Using proprietary Super  $\beta$  eta PNP<sup>®</sup> process with a PNP pass element, these regulators feature 350mV to 425mV (full load) typical dropout voltages and very low ground current. Designed for high current loads, these devices also find applications in lower current, extremely low dropout-critical systems, where their tiny dropout voltage and ground current values are important attributes.

The 29150/29300/29500/29750 are fully protected against overcurrent faults, reversed input polarity, reversed lead insertion, overtemperature operation, and positive and negative transient voltage spikes. Five pin fixed voltage versions feature logic level ON/OFF control and an error flag which signals whenever the output falls out of regulation. Flagged states include low input voltage (dropout), output current limit, overtemperature shutdown, and extremely high voltage spikes on the input.

On the 29xx1 and 29xx2, the ENABLE pin may be tied to VIN if it is not required for ON/OFF control. The 29150/29300/29500 are available in 3-pin and 5-pin TO-220 and surface mount TO-263 (D<sup>2</sup>Pak) packages. The 29750 7.5A regulators are available in 3-pin and 5-pin TO-247 packages. The 1.5A, adjustable output 29152 is available in a 5-pin power D-Pak (TO-252) package.

For applications with input voltage 6V or below, see HG37xxx LDOs.

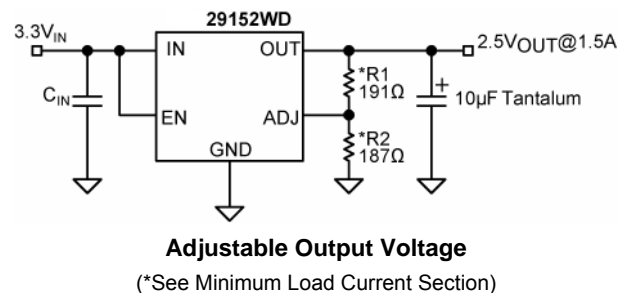
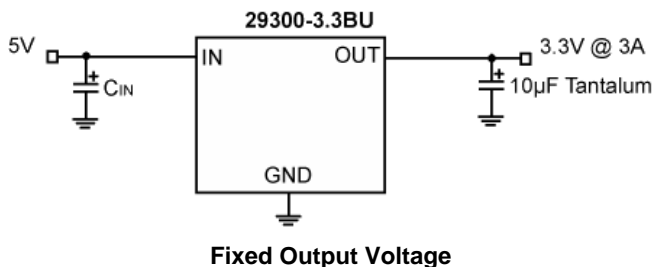
### Features

- High current capability  
29150/29151/29152/29153.....1.5A  
29300/29301/29302/29303.....3A  
29500/29501/29502/29503.....5A  
29750/29751/29752.....7.5A
- Low-dropout voltage
- Low ground current
- Accurate 1% guaranteed tolerance
- Extremely fast transient response
- Reverse-battery and “Load Dump” protection
- Zero-current shutdown mode (5-pin versions)
- Error flag signals output out-of-regulation (5-pin versions)
- Also characterized for smaller loads with industry-leading performance specifications
- Fixed voltage and adjustable versions

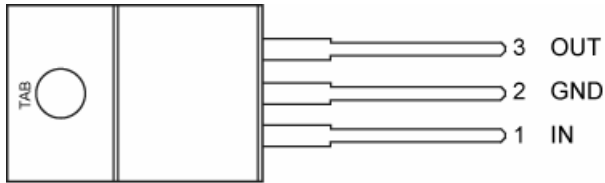
### Applications

- Battery powered equipment
- High-efficiency “Green” computer systems
- Automotive electronics
- High-efficiency linear lower supplies
- High-efficiency lost-regulator for switching supply

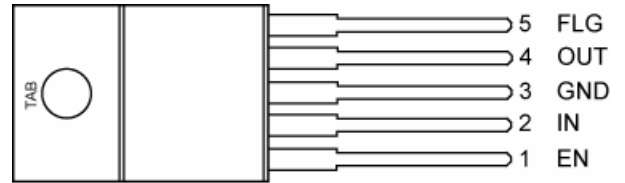
### Typical Application\*\*



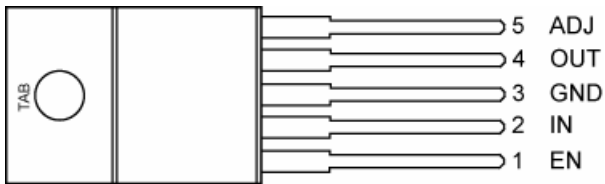
**Pin Configuration**



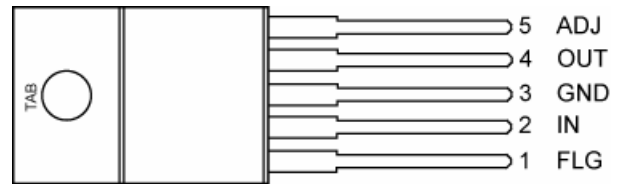
**3-Pin TO-220 (T)**  
29150/29300/29500



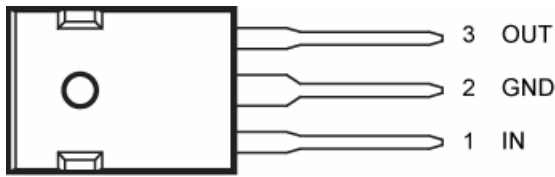
**5-Pin TO-220 Fixed Voltage (T)**  
29151/29301/29501/29751



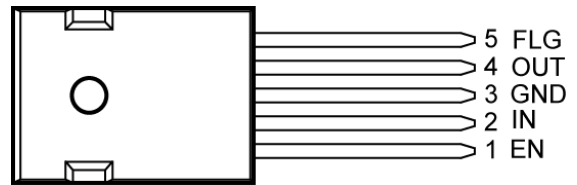
**5-Pin TO-220 Adjustable Voltage (T)**  
29152/29302/29502



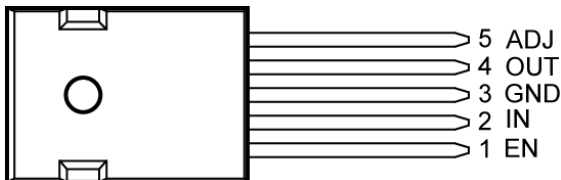
**5-Pin TO-220 Adjustable with Flag (T)**  
29153/29303/29503



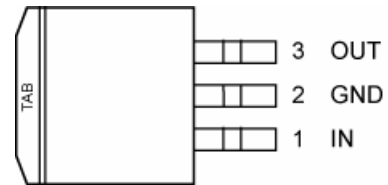
**3-Pin TO-247 (WT)**  
29750



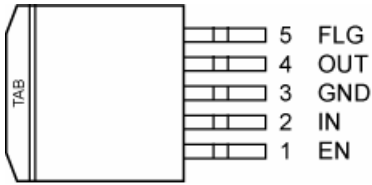
**5-Pin TO-247 Fixed Voltage (WT)**  
29751



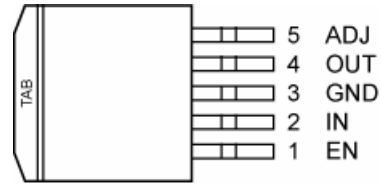
**5-Pin TO-247 Adjustable Voltage (WT)**  
29752



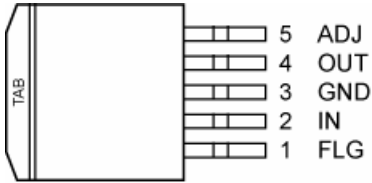
**3-Pin TO-263 (D<sup>2</sup>Pak) (UT)**  
29150/29300



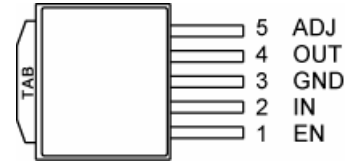
**5-Pin TO-263 (D<sup>2</sup>Pak) Fixed Voltage (U)**  
 29151/29301/29501



**5-Pin TO-263 (D<sup>2</sup>Pak) Adjustable Voltage (U)**  
 29302/29502



**5-Pin TO-263 (D<sup>2</sup>Pak) Adjustable with Flag (U)**  
 29153/29303/29503



**5-Pin TO-252 (D-Pak) Adjustable Voltage (D)**  
 29152

### Pin Description

Pin Number TO-220 TO-247 TO-263	Pin Name
1	INPUT: Supplies the current to the output power device
2	GND: TAB is also connected internally to the IC's ground on D-PAK.
3	OUTPUT: The regulator output voltage

### Pin Description

Pin Number Fixed TO-220 TO-247 TO-263	Pin Number Adjustable TO-220 TO-247 TO-252 TO-263	Pin Number Adj. with Flag TO-220 TO-247 TO-263	Pin Name
1	1	—	ENABLE: CMOS compatible control input. Logic high = enable, logic low = shutdown.
2	2	2	INPUT: Supplies the current to the output power device
3, TAB	3, TAB	3, TAB	GND: TAB is also connected internally to the IC's ground on D-PAK.
4	4	4	OUTPUT: The regulator output voltage
—	5	5	ADJUST: Adjustable regulator feedback input that connects to the resistor voltage divider that is placed from OUTPUT to GND in order to set the output voltage.
5	—	1	FLAG: Active low error flag output signal that indicates an output fault condition

**Absolute Maximum Ratings<sup>(1)</sup>**

Input Supply Voltage ( $V_{IN}$ ) <sup>(1)</sup> .....	-20V to +60V
Enable Input Voltage ( $V_{EN}$ ).....	-0.3V to $V_{IN}$
Lead Temperature (soldering, 5sec.).....	260°C
Power Dissipation .....	Internally Limited
Storage Temperature Range .....	-65°C to +150°C
ESD Rating.....	<b>Note 3</b>

**Operating Ratings<sup>(2)</sup>**

Operating Junction Temperature .....	-40°C to +125°C
Maximum Operating Input Voltage .....	26V
Package Thermal Resistance	
TO-220 ( $\theta_{JC}$ ).....	2°C/W
TO-263 ( $\theta_{JC}$ ).....	2°C/W
TO-247 ( $\theta_{JC}$ ) .....	1.5°C/W
TO-252 ( $\theta_{JC}$ ).....	3°C/W
TO-252 ( $\theta_{JA}$ ) .....	56°C/W

**Electrical Characteristics<sup>(4,13)</sup>**

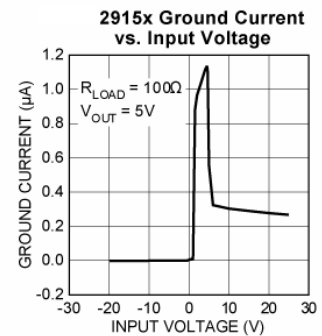
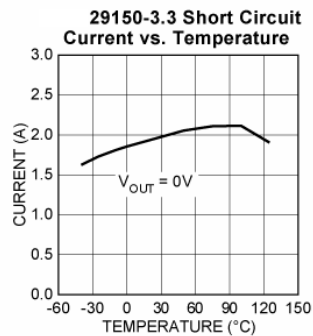
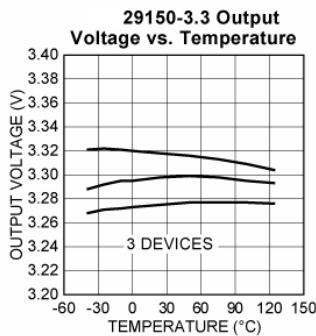
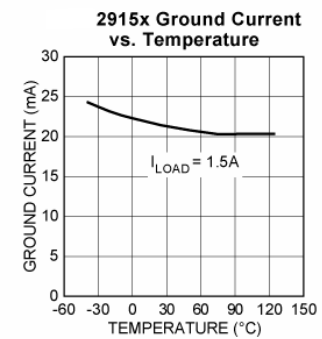
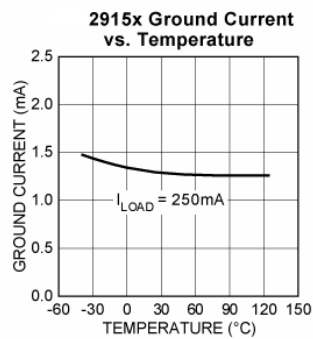
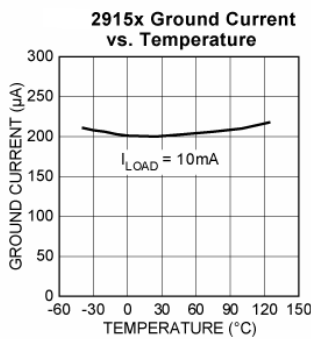
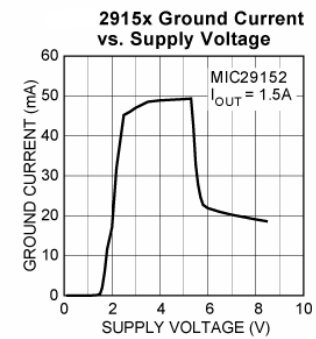
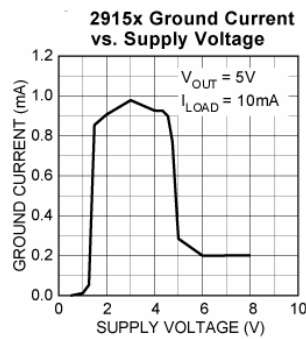
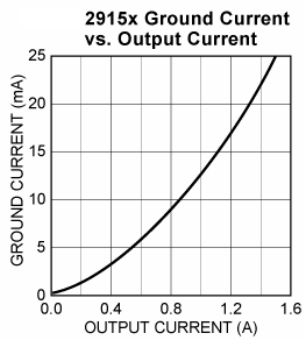
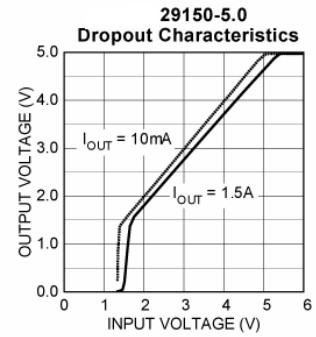
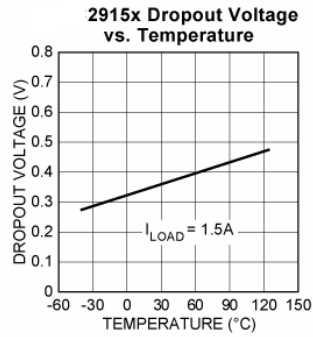
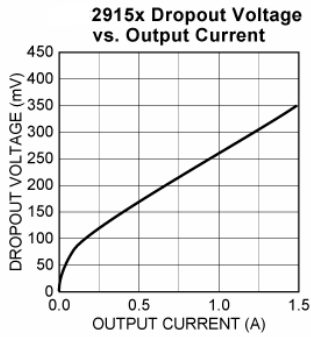
$V_{IN} = V_{OUT} + 1V$ ;  $I_{OUT} = 10mA$ ;  $T_J = 25^\circ C$ , bold values indicate  $-40^\circ C \leq T_J \leq +125^\circ C$ , unless noted.

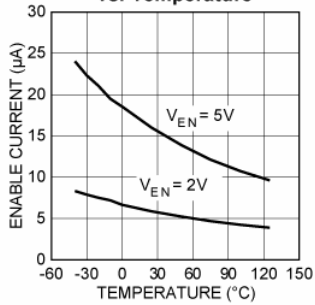
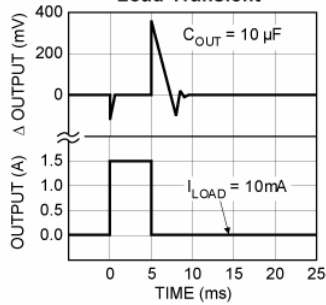
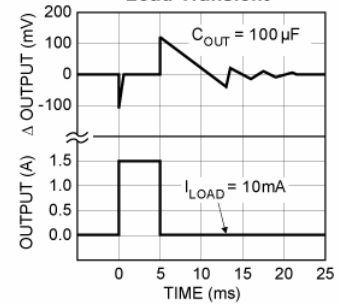
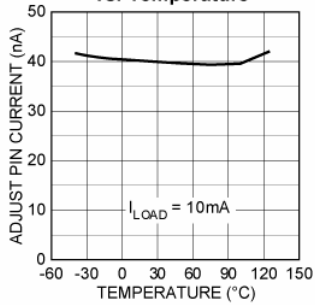
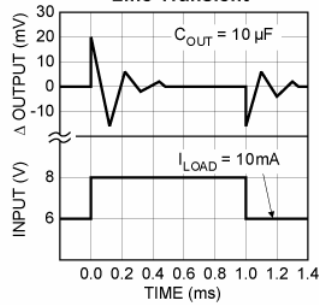
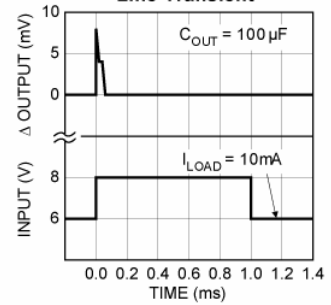
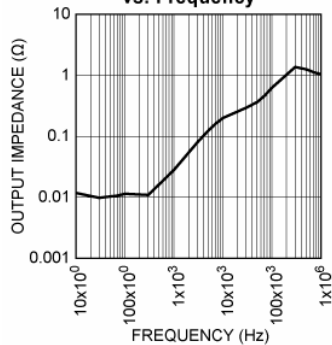
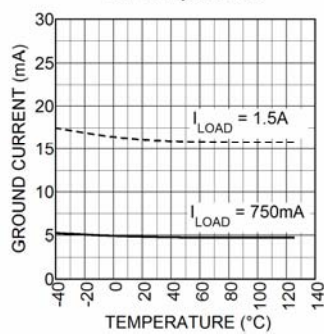
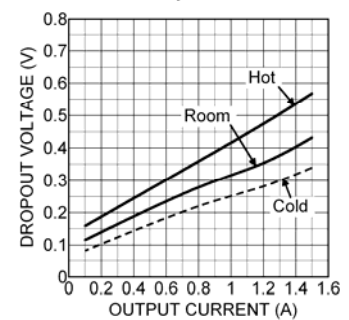
Parameter	Condition	Min	Typ	Max	Units	
Output Voltage	$I_{OUT} = 10mA$	-1		1	%	
	$10mA \leq I_{OUT} \leq I_{FL}$ , $(V_{OUT} + 1V) \leq V_{IN} \leq 26V$ <sup>(5)</sup>	-2		2	%	
Line Regulation	$I_{OUT} = 10mA$ , $(V_{OUT} + 1V) \leq V_{IN} \leq 26V$		0.06	0.5	%	
Load Regulation	$V_{IN} = V_{OUT} + 5V$ , $10mA \leq I_{OUT} \leq 1.5A$ <sup>(5,9)</sup>		0.2	1	%	
$\frac{\Delta V_O}{\Delta T}$	Output Voltage <sup>(9)</sup> Temperature Coefficient.		<b>20</b>	<b>100</b>	ppm°C	
Dropout Voltage	$\Delta V_{OUT} = -1\%$ <sup>(6)</sup>					
	29150	$I_{OUT} = 100mA$ $I_{OUT} = 750mA$	80 220	<b>200</b>	mV	
	29300	$I_{OUT} = 1.5A$ $I_{OUT} = 100mA$	350 80	<b>600</b> <b>175</b>		
		$I_{OUT} = 1.5A$ $I_{OUT} = 3A$	250 370	<b>600</b>		
	29500	$I_{OUT} = 250mA$ $I_{OUT} = 2.5A$	125 250	<b>250</b>		
		$I_{OUT} = 5A$ $I_{OUT} = 250mA$	370 80	<b>600</b> <b>200</b>		
29750	$I_{OUT} = 4A$ $I_{OUT} = 7.5A$	270 425	<b>750</b>			
	Ground Current	29150	$I_{OUT} = 750mA$ , $V_{IN} = V_{OUT} + 1V$ $I_{OUT} = 1.5A$	8 22	<b>20</b>	mA
		29300	$I_{OUT} = 1.5A$ , $V_{IN} = V_{OUT} + 1V$ $I_{OUT} = 3A$	10 37	<b>35</b>	mA
		29500	$I_{OUT} = 2.5A$ , $V_{IN} = V_{OUT} + 1V$ $I_{OUT} = 5A$	15 70	<b>50</b>	mA
29750		$I_{OUT} = 4A$ , $V_{IN} = V_{OUT} + 1V$ $I_{OUT} = 7.5A$	35 120	<b>75</b>	mA	
<b>Note 8</b>						
$I_{GRNDDO}$ Ground Pin Current at Droupout	$V_{IN} = 0.5V$ less than specified $V_{OUT} \times I_{OUT} = 10mA$					
	29150		0.9		mA	
	29300		1.7		mA	
	29500		2.1		mA	
29750		3.1		mA		
Current Limit	29150	$V_{OUT} = 0V$ <sup>(7)</sup>	2.1	<b>3.5</b>	A	
	29300	$V_{OUT} = 0V$ <sup>(7)</sup>	4.5	<b>5.0</b>	A	
	29500	$V_{OUT} = 0V$ <sup>(7)</sup>	7.5	<b>10.0</b>	A	
	29750	$V_{OUT} = 0V$ <sup>(7)</sup>	9.5	<b>15</b>	A	

Parameter	Condition	Min	Typ	Max	Units
e <sub>n</sub> , Output Noise Voltage (10Hz to 100kHz) I <sub>L</sub> = 100mA	C <sub>L</sub> = 10μF		400		μV (rms)
	C <sub>L</sub> = 33μF		260		
Ground Current in Shutdown	29150/1/2/3 only V <sub>EN</sub> = 0.4V		2	10 <b>30</b>	μA μA
<b>Reference 29xx2/29xx3</b>					
Reference Voltage		1.228 <b>1.215</b>	1.240	1.252 <b>1.265</b>	V V
Reference Voltage		<b>1.203</b>		<b>1.277</b>	V
Adjust Pin Bias Current			40	80 <b>120</b>	nA
Reference Voltage Temperature Coefficient	<sup>(10)</sup>		20		ppm°C
Adjust Pin Bias Current Temperature Coefficient			0.1		nA/°C
<b>Flag Output (Error Comparator) 29xx1/29xx3</b>					
Output Leakage Current	V <sub>OH</sub> = 26V		0.01	1.00 <b>2.00</b>	μA
Output Low Voltage	Device set for 5V, V <sub>IN</sub> = 4.5V I <sub>OL</sub> = 250μA		220	300 <b>400</b>	mV
Upper Threshold Voltage	Device set for 5V <sup>(11)</sup>	40 <b>25</b>	60		mV
Lower Threshold Voltage	Device set for 5V <sup>(11)</sup>		75	95 <b>140</b>	mV
Hysteresis	Device set for 5V <sup>(11)</sup>		15		mV
<b>ENABLE Input 29xx1/29xx2</b>					
Input Logic Voltage Low (OFF) High (ON)		<b>2.4</b>		<b>0.8</b>	V
Enable Pin Input Current	V <sub>EN</sub> = 26V		100	600 <b>750</b>	μA
	V <sub>EN</sub> = 0.8V	0.7		2 <b>4</b>	μA
Regulator Output Current in Shutdown	<sup>(12)</sup>		10	<b>500</b>	μA

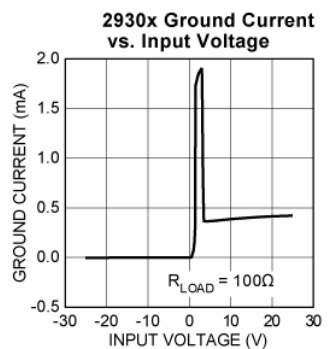
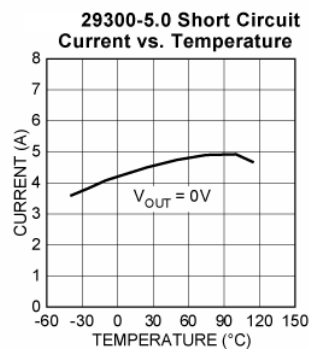
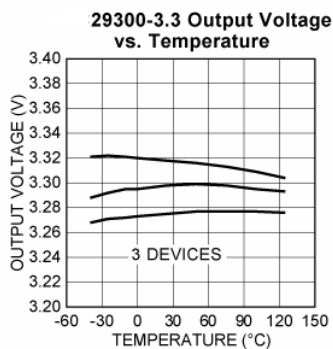
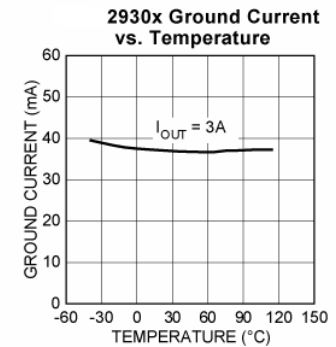
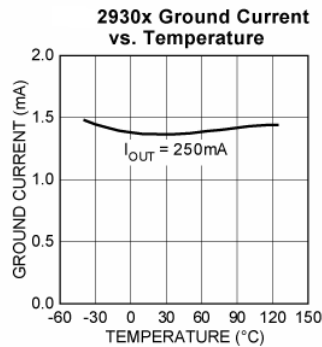
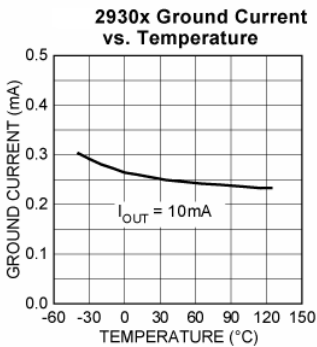
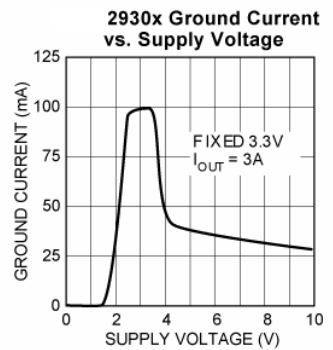
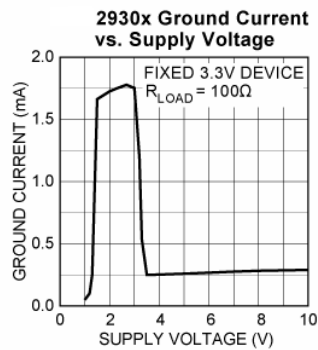
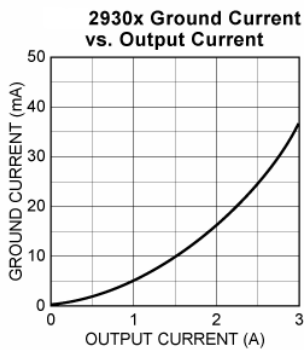
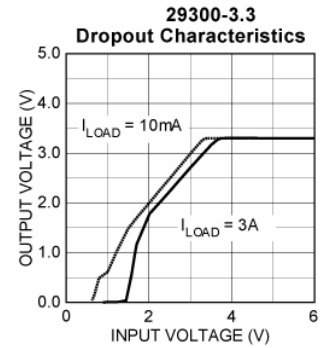
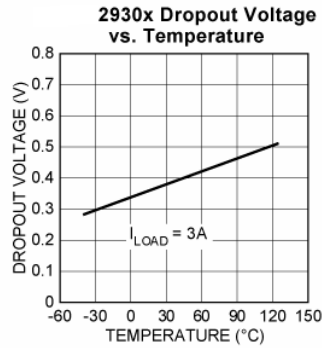
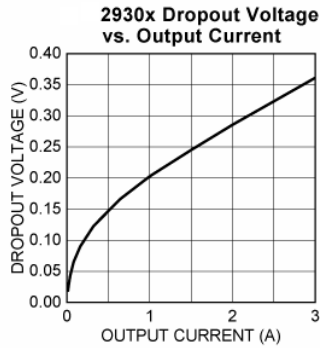
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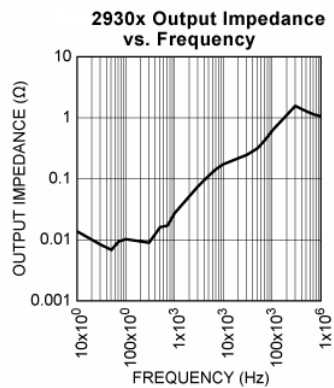
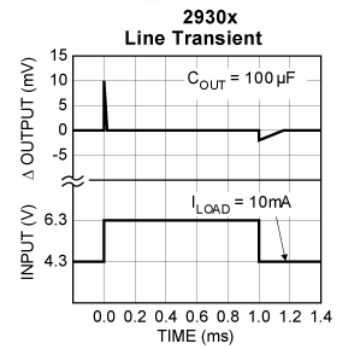
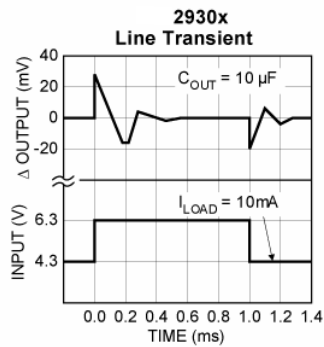
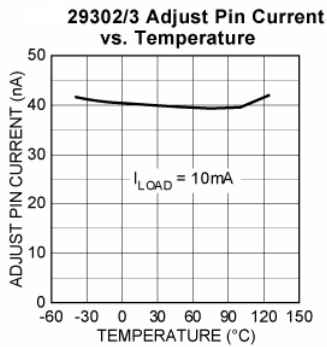
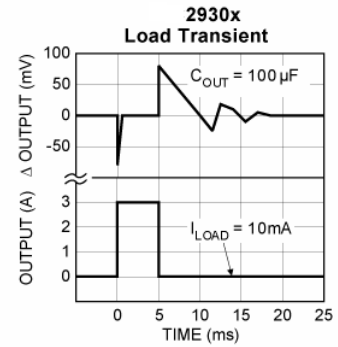
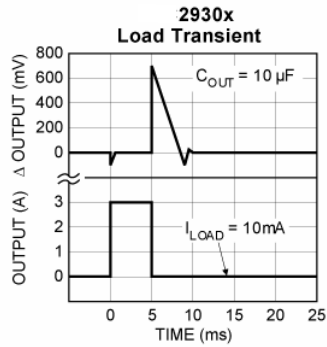
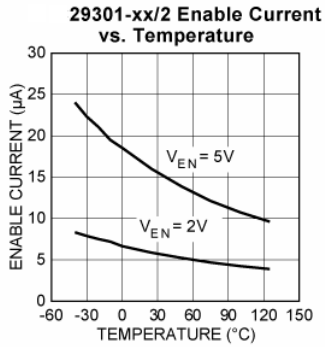
1. Maximum positive supply voltage of 60V must be of limited duration (<100msec) and duty cycle ( $\leq 1\%$ ). The maximum continuous supply voltage is 26V. Exceeding the absolute maximum rating may damage the device.
2. The device is not guaranteed to function outside its operating rating.
3. Devices are ESD sensitive. Handling precautions recommended.
4. Specification for packaged product only.
5. Full load current ( $I_{FL}$ ) is defined as 1.5A for the 29150, 3A for the 29300, 5A for the 29500, and 7.5A for the 29750 families.
6. Dropout voltage is defined as the input-to-output differential when the output voltage drops to 99% of its normal value with  $V_{OUT} + 1V$  applied to  $V_{IN}$ .
7.  $V_{IN} = V_{OUT(nominal)} + 1V$ . For example, use  $V_{IN} = 4.3V$  for a 3.3V regulator or use 6V for a 5V regulator. Employ pulse-testing procedures to pin current.
8. Ground pin current is the regulator quiescent current. The total current drawn from the source is the sum of the load current plus the ground pin current.
9. Output voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.
10. Thermal regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 200mA load pulse at  $V_{IN} = 20V$  (a 4W pulse) for  $T = 10ms$ .
11. Comparator thresholds are expressed in terms of a voltage differential at the Adjust terminal below the nominal reference voltage measured at 6V input. To express these thresholds in terms of output voltage change, multiply by the error amplifier gain  $= V_{OUT}/V_{REF} = (R1 + R2)/R2$ . For example, at a programmed output voltage of 5V, the Error output is guaranteed to go low when the output drops by  $95mV \times 5V/1.240V = 384mV$ . Thresholds remain constant as a percent of  $V_{OUT}$  as  $V_{OUT}$  is varied, with the dropout warning occurring at typically 5% below nominal, 7.7% guaranteed.
12.  $V_{EN} \leq 0.8V$  and  $V_{IN} \leq 26V$ ,  $V_{OUT} = 0$ .
13. When used in dual supply systems where the regulator load is returned to a negative supply, the output voltage must be diode clamped to ground.

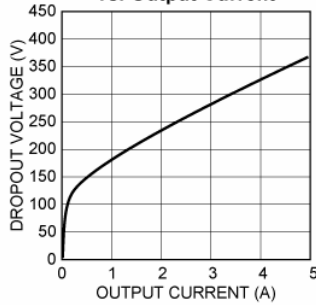
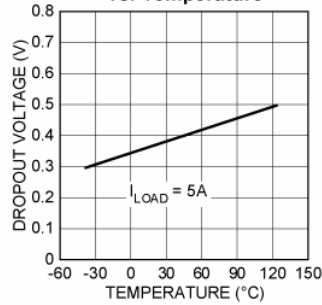
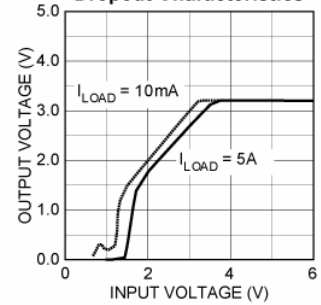
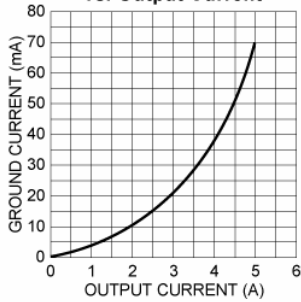
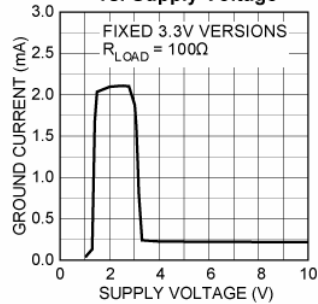
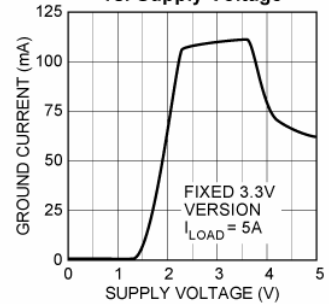
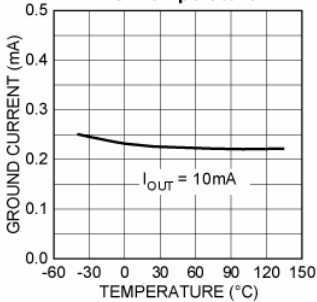
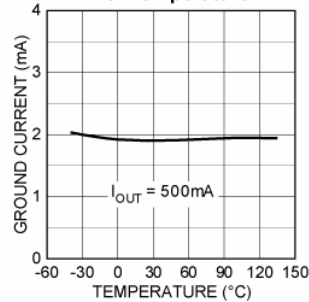
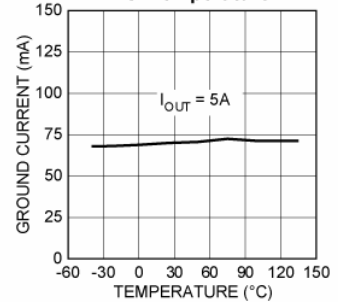
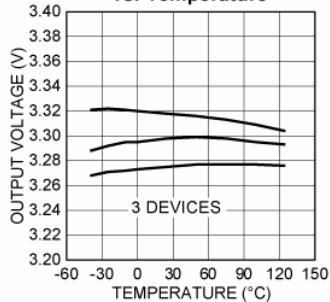
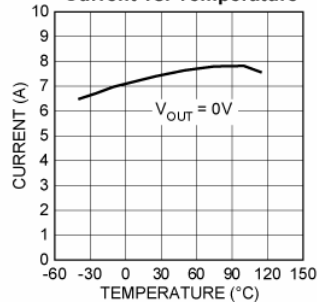
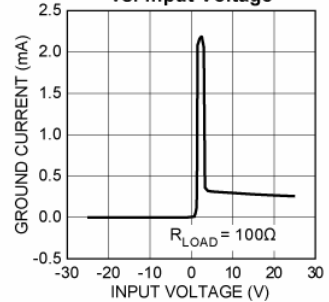
**Typical Characteristics 2915x**


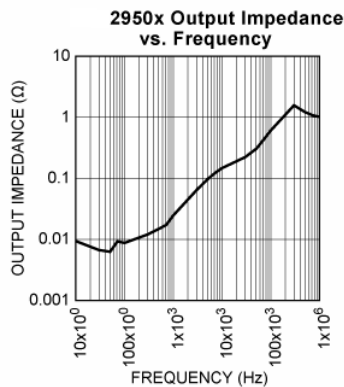
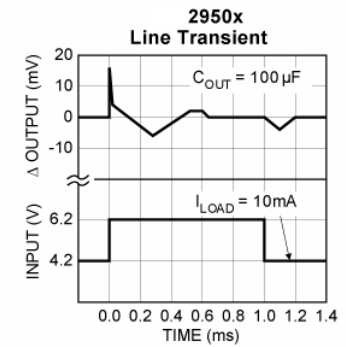
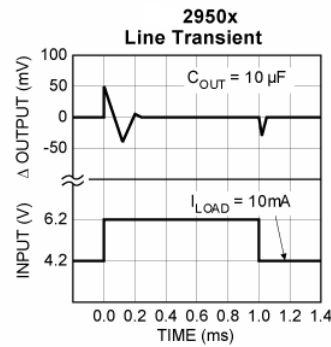
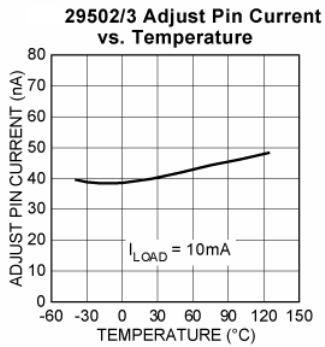
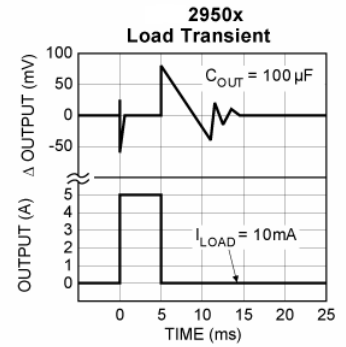
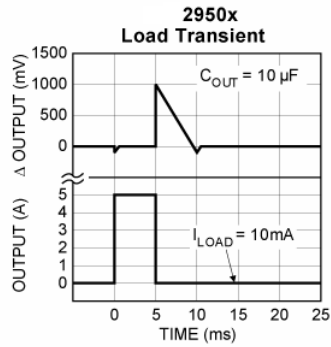
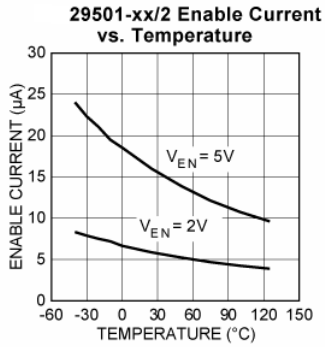
**29151-xx/2 Enable Current vs. Temperature**

**2915x Load Transient**

**2915x Load Transient**

**29152/3 Adjust Pin Current vs. Temperature**

**2915x Line Transient**

**2915x Line Transient**

**2915x Output Impedance vs. Frequency**

**29152 Ground Current vs. Temperature**

**29152 Dropout Voltage vs. Output Current**


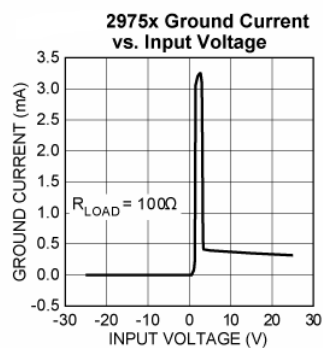
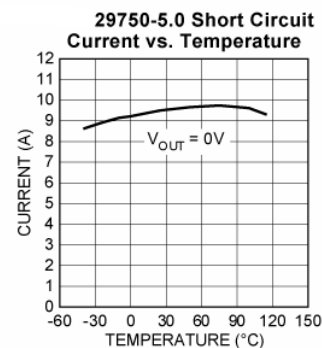
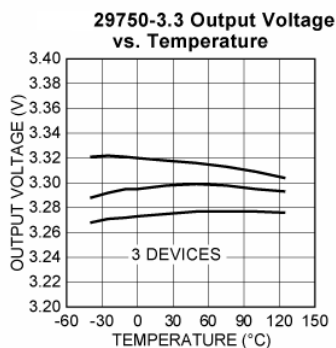
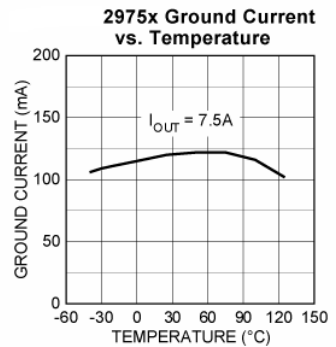
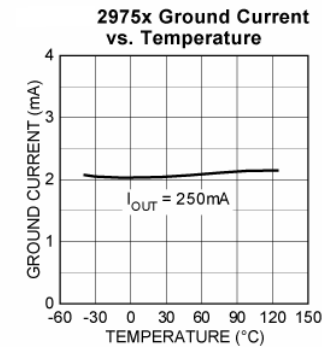
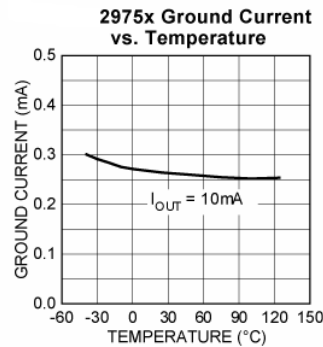
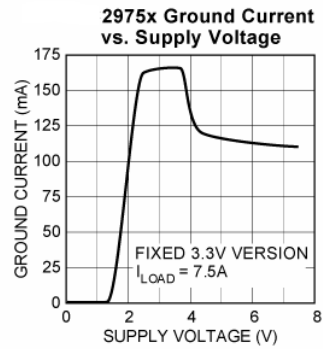
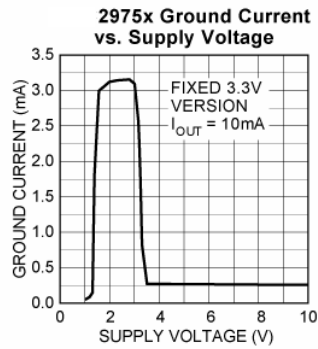
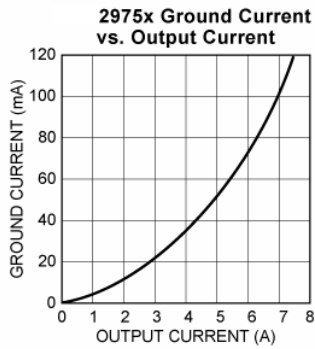
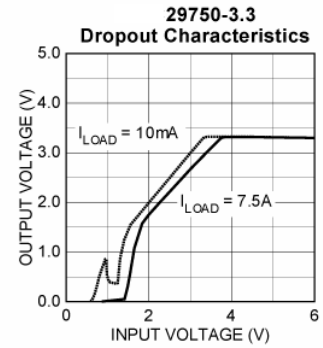
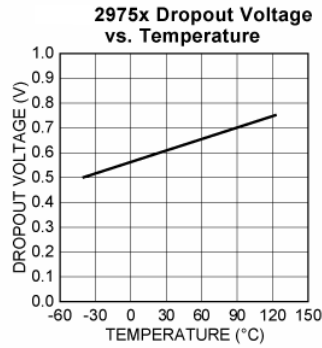
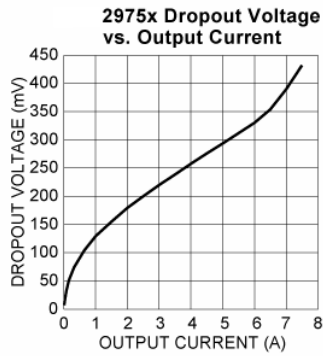


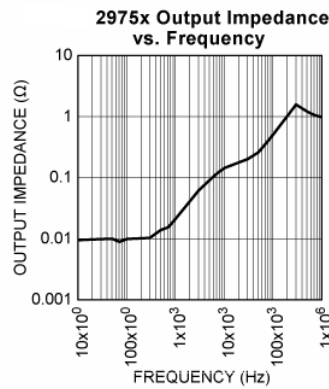
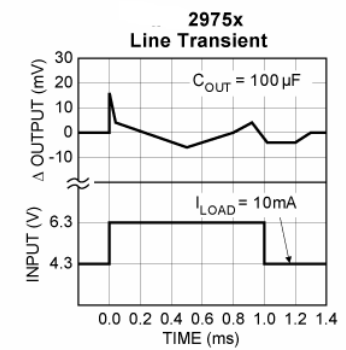
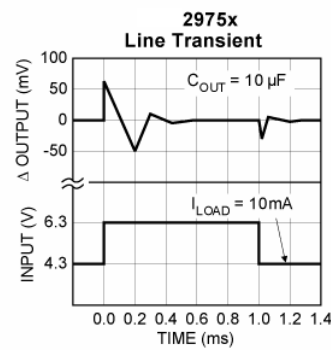
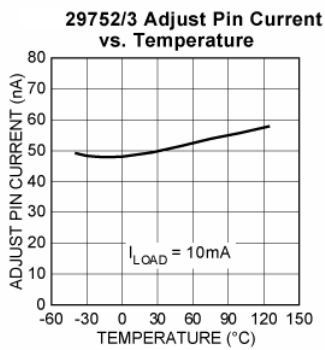
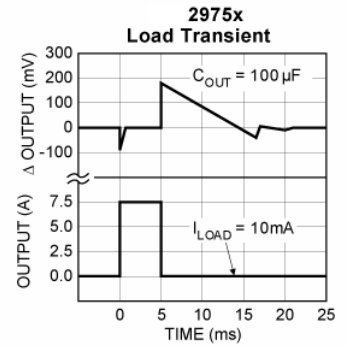
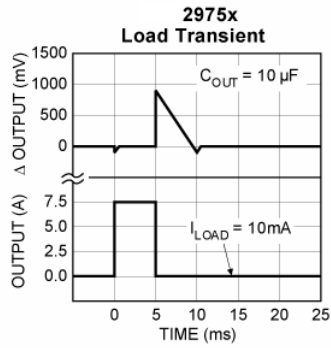
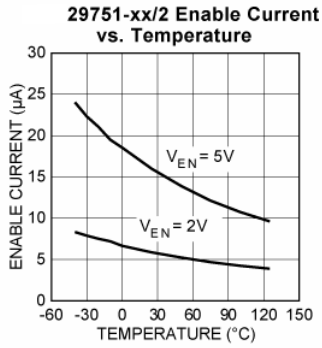
**Typical Characteristics 2930x**




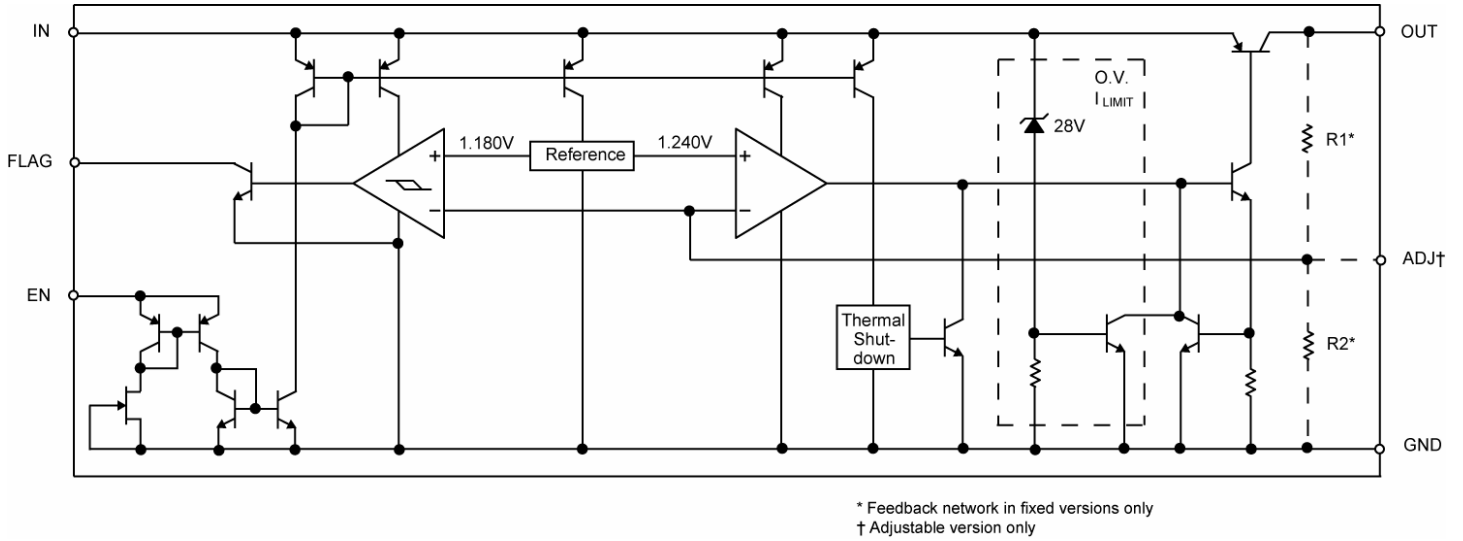
**Typical Characteristics 2950x**
**2950x Dropout Voltage vs. Output Current**

**2950x Dropout Voltage vs. Temperature**

**29500-3.3 Dropout Characteristics**

**2950x Ground Current vs. Output Current**

**2950x Ground Current vs. Supply Voltage**

**2950x Ground Current vs. Supply Voltage**

**2950x Ground Current vs. Temperature**

**2950x Ground Current vs. Temperature**

**2950x Ground Current vs. Temperature**

**29500-3.3 Output Voltage vs. Temperature**

**2950x-5.0 Short Circuit Current vs. Temperature**

**2950x Ground Current vs. Input Voltage**




**Typical Characteristics 2975x**




**Functional Diagram**



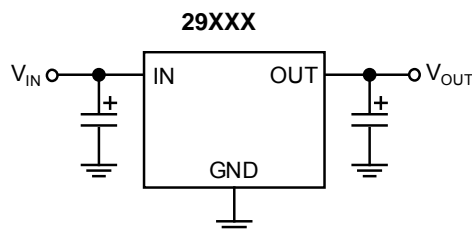
## Application Information

The 29150/29300/29500/29750 are high performance low-dropout voltage regulators suitable for all moderate to high-current voltage regulator applications. Their 350mV to 425mV typical dropout voltage at full load make them especially valuable in battery powered systems and as high efficiency noise filters in “post-regulator” applications. Unlike older NPN-pass transistor designs, where the minimum dropout voltage is limited by the base-emitter voltage drop and collector-emitter saturation voltage, dropout performance of the PNP output of these devices is limited merely by the low  $V_{CE}$  saturation voltage.

A trade-off for the low-dropout voltage is a varying base driver requirement. But Super Beta PNP<sup>®</sup> process reduces this drive requirement to merely 1% of the load current.

The 29150/29300/29500/29750 family of regulators are fully protected from damage due to fault conditions. Current limiting is provided. This limiting is linear; output current under overload conditions is constant. Thermal shutdown disables the device when the die temperature exceeds the 125°C maximum safe operating temperature. Transient protection allows device (and load) survival even when the input voltage spikes between -20V and +60V. When the input voltage exceeds about 35V to 40V, the over voltage sensor temporarily disables the regulator. The output structure of these regulators allows voltages in excess of the desired output voltage to be applied without reverse current flow. 29xx1 and 29xx2 versions offer a logic level ON/OFF control: when disabled, the devices draw nearly zero current.

An additional feature of this regulator family is a common pinout: a design’s current requirement may change up or down yet use the same board layout, as all of these regulators have identical pinouts.



**Figure 3. Linear regulators require only two capacitors for operation.**

### Thermal Design

Linear regulators are simple to use. The most complicated design parameters to consider are thermal characteristics. Thermal design requires the following application-specific parameters:

- Maximum ambient temperature,  $T_A$

- Output Current,  $I_{OUT}$
- Output Voltage,  $V_{OUT}$
- Input Voltage,  $V_{IN}$

First, we calculate the power dissipation of the regulator from these numbers and the device parameters from this datasheet.

$$P_D = I_{OUT}(1.01 V_{IN} - V_{OUT})$$

Where the ground current is approximated by 1% of  $I_{OUT}$ . Then the heat sink thermal resistance is determined with this formula:

$$\theta_{SA} = \frac{T_{JMAX} - T_A}{P_D} - (\theta_{JC} + \theta_{CS})$$

Where  $T_{JMAX} \leq 125^\circ\text{C}$  and  $\theta_{CS}$  is between 0 and  $2^\circ\text{C/W}$ .

The heat sink may be significantly reduced in applications where the minimum input voltage is known and is large compared with the dropout voltage. Use a series input resistor to drop excessive voltage and distribute the heat between this resistor and the regulator. The low-dropout properties of Super Beta PNP<sup>®</sup> regulators allow very significant reductions in regulator power dissipation and the associated heat sink without compromising performance. When this technique is employed, a capacitor of at least 0.1µF is needed directly between the input and regulator ground.

Please refer to Application Note 9 and Application Hint 17 for further details and examples on thermal design and heat sink specification.

With no heat sink in the application, calculate the junction temperature to determine the maximum power dissipation that will be allowed before exceeding the maximum junction temperature of the 29152. The maximum power allowed can be calculated using the thermal resistance ( $\theta_{JA}$ ) of the D-Pak adhering to the following criteria for the PCB design: 2 oz. copper and 100mm<sup>2</sup> copper area for the 29152.

For example, given an expected maximum ambient temperature ( $T_A$ ) of 75°C with  $V_{IN} = 3.3\text{V}$ ,  $V_{OUT} = 2.5\text{V}$ , and  $I_{OUT} = 1.5\text{A}$ , first calculate the expected  $P_D$  using Equation (1);

$$P_D = (3.3\text{V} - 2.5\text{V})1.5\text{A} - (3.3\text{V})(0.016\text{A}) = 1.1472\text{W}$$

Next, calculate the junction temperature for the expected power dissipation.

$$T_J = (\theta_{JA} \times P_D) + T_A = (56^\circ\text{C/W} \times 1.1472\text{W}) + 75^\circ\text{C} = 139.24^\circ\text{C}$$

Now determine the maximum power dissipation allowed that would not exceed the IC’s maximum junction temperature (125°C) without the use of a heat sink by

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA} = (125^\circ\text{C} - 75^\circ\text{C}) / (56^\circ\text{C/W}) = 0.893\text{W}$$



### Capacitor Requirements

For stability and minimum output noise, a capacitor on the regulator output is necessary. The value of this capacitor is dependent upon the output current; lower currents allow smaller capacitors.

This capacitor need not be an expensive low ESR type: aluminum electrolytics are adequate. In fact, extremely low ESR capacitors may contribute to instability. Tantalum capacitors are recommended for systems where fast load transient response is important.

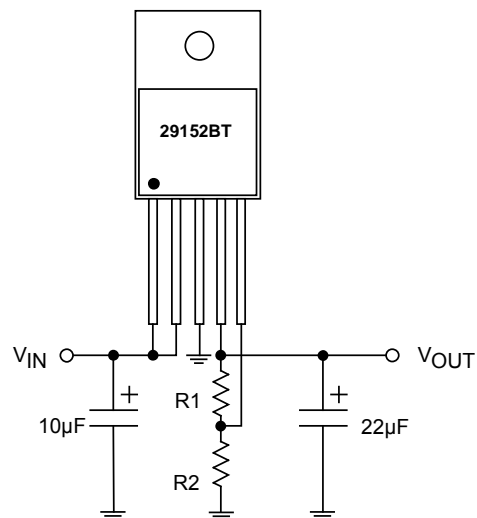
Where the regulator is powered from a source with a high AC impedance, a 0.1μF capacitor connected between Input and GND is recommended. This capacitor should have good characteristics to above 250kHz.

### Minimum Load Current

The 29150–29750 regulators are specified between finite loads. If the output current is too small, leakage currents dominate and the output voltage rises. The following minimum load current swamps any expected leakage current across the operating temperature range:

Device	Minimum Load
29150.....	5mA
29300.....	7mA
29500.....	10mA
29750.....	10mA

### Adjustable Regulator Design



**Figure 4. Adjustable Regulator with Resistors**

The adjustable regulator versions, 29xx2 and 29xx3, allow programming the output voltage anywhere between 1.25V and the 25V. Two resistors are used. The resistor values are calculated by:

$$R_1 = R_2 \times \left( \frac{V_{OUT}}{1.240} - 1 \right)$$

where V<sub>OUT</sub> is the desired output voltage. Figure 4 shows component definition. Applications with widely varying load currents may scale the resistors to draw the minimum load current required for proper operation (see “Minimum Load Current” section).

### Error Flag

29xx1 and 29xx3 versions feature an Error Flag, which looks at the output voltage and signals an error condition when this voltage drops 5% below its expected value. The error flag is an open-collector output that pulls low under fault conditions. It may sink 10mA. Low output voltage signifies a number of possible problems, including an overcurrent fault (the device is in current limit) and low input voltage. The flag output is inoperative during overtemperature shutdown conditions.

### Enable Input

29xx1 and 29xx2 versions feature an enable (EN) input that allows ON/OFF control of the device. Special design allows “zero” current drain when the device is disabled—only microamperes of leakage current flows. The EN input has TTL/CMOS compatible thresholds for simple interfacing with logic, or may be directly tied to ≤30V. Enabling the regulator requires approximately 20μA of current.