

## 1 Features

- Low-Voltage Operation/Wide Adjust Range (1.24 V/30 V)
- 0.5% Initial Tolerance (XB431)
- Temperature Compensated for Industrial Temperature Range (39 PPM/°C for the XB431)
- Low Operation Current (55  $\mu$ A)
- Low Output Impedance (0.25  $\Omega$ )
- Fast Turn-On Response
- Low Cost

## 2 Applications

- Shunt Regulator
- Series Regulator
- Current Source or Sink
- Voltage Monitor
- Error Amplifier
- 3-V Off-Line Switching Regulator
- Low Dropout N-Channel Series Regulator

## 3 Description

The XB431 are precision 1.24 V shunt regulators capable of adjustment to 30 V. Negative feedback from the cathode to the adjust pin controls the cathode voltage, much like a non-inverting op amp configuration (Refer to [Symbol and Functional Diagrams](#)). A two-resistor voltage divider terminated at the adjust pin controls the gain of a 1.24 V band-gap reference. Shorting the cathode to the adjust pin (voltage follower) provides a cathode voltage of a 1.24 V.

The XB431 have respective initial tolerances of 1.5%, 1%, and 0.5%, and functionally lend themselves to several applications that require zener diode type performance at low voltages. Applications include a 3 V to 2.7 V low drop-out regulator, an error amplifier in a 3 V off-line switching regulator and even as a voltage detector. These parts are typically stable with capacitive loads greater than 10 nF and less than 50 pF.

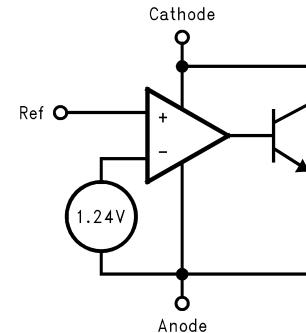
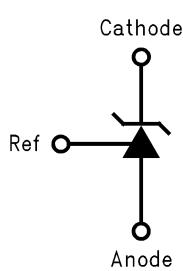
The XB431 provide performance at a competitive price.

## 4 Device Information<sup>(1)</sup>

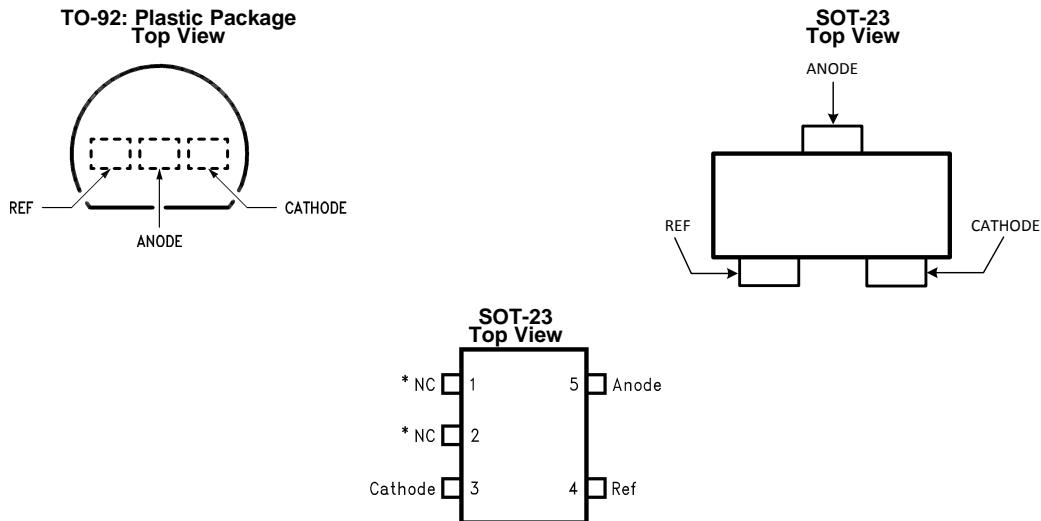
PART NUMBER	PACKAGE	BODY SIZE (NOM)
XB431	SOT-23 (5)	2.90 mm x 1.60 mm
	TO-92 (3)	4.30 mm x 4.30 mm
	SOT-23 (3)	2.92 mm x 1.30 mm

(1) For all available packages, see the orderable addendum at the end of the datasheet.

## 5 Symbol and Functional Diagrams



## 6 Pin Configurations and Functions



\*Pin 1 is not internally connected.

\*Pin 2 is internally connected to Anode pin. Pin 2 should be either floating or connected to Anode pin.

## 7 Specifications

### 7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		MIN	MAX	UNIT
Operating temperature	Industrial (XB431)	-40	85	°C
	Commercial (XB431)	0	70	
Lead temperature	TO-92 Package/SOT-23 -5,-3 Package (Soldering, 10 sec.)		265	
Internal power dissipation <sup>(2)</sup>	TO-92		0.78	W
	SOT-23-5, -3 Package		0.28	W
Cathode voltage			35	V
Continuous cathode current		-30	30	mA
Reference input current		-0.05	3	

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) Ratings apply to ambient temperature at 25°C. Above this temperature, derate the TO-92 at 6.2 mW/°C, and the SOT-23-5 at 2.2 mW/°C. See derating curve in [Operating Condition](#) section.

### 7.2 Handling Ratings

		MIN	MAX	UNIT
T <sub>stg</sub>	Storage temperature range	-65	150	°C
V <sub>(ESD)</sub>	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins <sup>(1)</sup>	2000	V

- (1) The human body model is a 100 pF capacitor discharged through a 1.5kΩ resistor into each pin. The machine model is a 200 pF capacitor discharged directly into each pin. MIL-STD-883 3015.7.

### 7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	NOM	MAX	UNIT															
Cathode voltage	V <sub>REF</sub>	30		V															
Cathode current		0.1	15	mA															
Temperature	XB431	-40	85	°C															
Derating Curve (Slope = -1/R <sub>θJA</sub> )				<p>The graph plots Maximum Continuous Dissipation (mW) on the y-axis (log scale from 100 to 1000) against Temperature (°C) on the x-axis (linear scale from 25 to 125). Two curves are shown: a steeper one for TO-92 and a flatter one for SOT-23. Both curves start at approximately 850 mW at 25°C and decrease as temperature increases.</p> <table border="1"> <caption>Approximate data points from Derating Curve graph</caption> <thead> <tr> <th>Temperature (°C)</th> <th>TO-92 (mW)</th> <th>SOT-23 (mW)</th> </tr> </thead> <tbody> <tr> <td>25</td> <td>850</td> <td>850</td> </tr> <tr> <td>70</td> <td>550</td> <td>450</td> </tr> <tr> <td>85</td> <td>450</td> <td>350</td> </tr> <tr> <td>125</td> <td>250</td> <td>200</td> </tr> </tbody> </table>	Temperature (°C)	TO-92 (mW)	SOT-23 (mW)	25	850	850	70	550	450	85	450	350	125	250	200
Temperature (°C)	TO-92 (mW)	SOT-23 (mW)																	
25	850	850																	
70	550	450																	
85	450	350																	
125	250	200																	

### 7.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>	XB431	XB431	XB431	UNIT
	SOT-23	SOT-23	TO-92	
	3 PINS	5 PINS	3 PINS	
R <sub>θJA</sub> Junction-to-ambient thermal resistance <sup>(2)</sup>	455	455	161	°C/W

- (1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

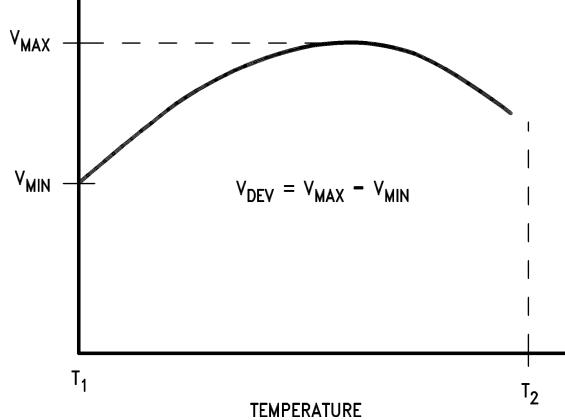
- (2) T<sub>J Max</sub> = 150°C, T<sub>J</sub> = T<sub>A</sub> + (R<sub>θJA</sub> P<sub>D</sub>), where P<sub>D</sub> is the operating power of the device.

## 7.5 XB431 Electrical Characteristics

$T_A = 25^\circ\text{C}$  unless otherwise specified

SYMBOL	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{\text{REF}}$	Reference Voltage	$V_Z = V_{\text{REF}}, I_Z = 10 \text{ mA}$ (See <a href="#">Figure 32</a> )	$T_A = 25^\circ\text{C}$	1.222	1.24	1.258
			$T_A = \text{Full Range}$	1.21		1.27
$V_{\text{DEV}}$	Deviation of Reference Input Voltage Over Temperature <sup>(1)</sup>	$V_Z = V_{\text{REF}}, I_Z = 10 \text{ mA},$ $T_A = \text{Full Range}$ (See <a href="#">Figure 32</a> )		4	12	mV
$\frac{\Delta V_{\text{REF}}}{\Delta V_Z}$	Ratio of the Change in Reference Voltage to the Change in Cathode Voltage	$I_Z = 10 \text{ mA}$ (see <a href="#">Figure 33</a> ) $V_Z$ from $V_{\text{REF}}$ to 6 V $R_1 = 10 \text{ k}\Omega, R_2 = \infty$ and $2.6 \text{ k}\Omega$		-1.5	-2.7	mV/V
$I_{\text{REF}}$	Reference Input Current	$R_1 = 10 \text{ k}\Omega, R_2 = \infty$ $I_I = 10 \text{ mA}$ (see <a href="#">Figure 33</a> )		0.15	0.5	$\mu\text{A}$
$\alpha I_{\text{REF}}$	Deviation of Reference Input Current over Temperature	$R_1 = 10 \text{ k}\Omega, R_2 = \infty,$ $I_I = 10 \text{ mA}, T_A = \text{Full Range}$ (see <a href="#">Figure 33</a> )		0.05	0.3	$\mu\text{A}$
$I_{Z(\text{MIN})}$	Minimum Cathode Current for Regulation	$V_Z = V_{\text{REF}}$ (see <a href="#">Figure 32</a> )		55	80	$\mu\text{A}$
$I_{Z(\text{OFF})}$	Off-State Current	$V_Z = 6 \text{ V}, V_{\text{REF}} = 0 \text{ V}$ (see <a href="#">Figure 34</a> )		0.001	0.1	$\mu\text{A}$
$r_Z$	Dynamic Output Impedance <sup>(2)</sup>	$V_Z = V_{\text{REF}}, I_Z = 0.1 \text{ mA to } 15 \text{ mA}$ Frequency = 0 Hz (see <a href="#">Figure 32</a> )		0.25	0.4	$\Omega$

- (1) Deviation of reference input voltage,  $V_{\text{DEV}}$ , is defined as the maximum variation of the reference input voltage over the full temperature range. See the following:



The average temperature coefficient of the reference input voltage,  $\alpha V_{\text{REF}}$ , is defined as:

$$\alpha V_{\text{REF}} \frac{\text{ppm}}{^\circ\text{C}} = \frac{\pm \left( \frac{V_{\text{Max}} - V_{\text{Min}}}{V_{\text{REF}}(\text{at } 25^\circ\text{C})} \right) 10^6}{T_2 - T_1} = \frac{\pm \left( \frac{V_{\text{DEV}}}{V_{\text{REF}}(\text{at } 25^\circ\text{C})} \right) 10^6}{T_2 - T_1}$$

Where:  $T_2 - T_1$  = full temperature change.  $\alpha V_{\text{REF}}$  can be positive or negative depending on whether the slope is positive or negative.

Example:  $V_{\text{DEV}} = 6 \text{ mV}$ ,  $V_{\text{REF}} = 1240 \text{ mV}$ ,  $T_2 - T_1 = 125^\circ\text{C}$ .

$$\alpha V_{\text{REF}} = \frac{\left( \frac{6.0 \text{ mV}}{1240 \text{ mV}} \right) 10^6}{125^\circ\text{C}} = +39 \text{ ppm / } ^\circ\text{C}$$

- (2) The dynamic output impedance,  $r_Z$ , is defined as:

$$r_Z = \frac{\Delta V_Z}{\Delta I_Z}$$

When the device is programmed with two external resistors,  $R_1$  and  $R_2$ , (see [Figure 33](#)), the dynamic output impedance of the overall circuit,  $r_Z$ , is defined as:

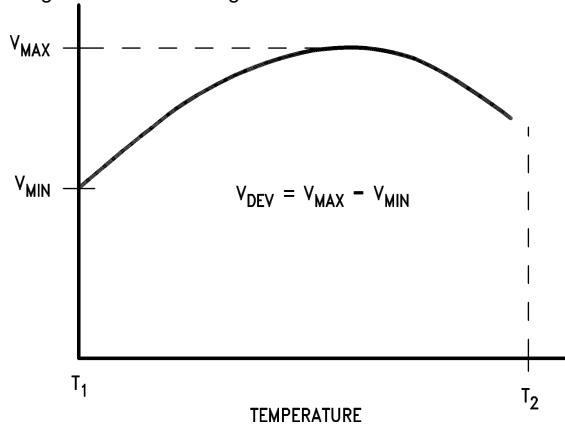
$$r_Z = \frac{\Delta V_Z}{\Delta I_Z} \approx \left[ r_Z \left( 1 + \frac{R_1}{R_2} \right) \right]$$

## 7.6 XB431 Electrical Characteristics

$T_A = 25^\circ\text{C}$  unless otherwise specified

SYMBOL	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{\text{REF}}$	Reference Voltage	$V_Z = V_{\text{REF}}, I_Z = 10 \text{ mA}$ (See <a href="#">Figure 32</a> )	$T_A = 25^\circ\text{C}$	1.222	1.24	1.258
			$T_A = \text{Full Range}$	1.202		1.278
$V_{\text{DEV}}$	Deviation of Reference Input Voltage Over Temperature <sup>(1)</sup>	$V_Z = V_{\text{REF}}, I_Z = 10 \text{ mA}, T_A = \text{Full Range}$ (See <a href="#">Figure 32</a> )		6	20	mV
$\frac{\Delta V_{\text{REF}}}{\Delta V_Z}$	Ratio of the Change in Reference Voltage to the Change in Cathode Voltage	$I_Z = 10 \text{ mA}$ (see <a href="#">Figure 33</a> ) $V_Z$ from $V_{\text{REF}}$ to 6V $R_1 = 10 \text{ k}\Omega, R_2 = \infty$ and $2.6 \text{ k}\Omega$		-1.5	-2.7	mV/V
$I_{\text{REF}}$	Reference Input Current	$R_1 = 10 \text{ k}\Omega, R_2 = \infty$ $I_I = 10 \text{ mA}$ (see <a href="#">Figure 33</a> )		0.15	0.5	$\mu\text{A}$
$\alpha I_{\text{REF}}$	Deviation of Reference Input Current over Temperature	$R_1 = 10 \text{ k}\Omega, R_2 = \infty$ , $I_I = 10 \text{ mA}, T_A = \text{Full Range}$ (see <a href="#">Figure 33</a> )		0.1	0.4	$\mu\text{A}$
$I_{Z(\text{MIN})}$	Minimum Cathode Current for Regulation	$V_Z = V_{\text{REF}}$ (see <a href="#">Figure 32</a> )		55	80	$\mu\text{A}$
$I_{Z(\text{OFF})}$	Off-State Current	$V_Z = 6 \text{ V}, V_{\text{REF}} = 0 \text{ V}$ (see <a href="#">Figure 34</a> )		0.001	0.1	$\mu\text{A}$
$r_Z$	Dynamic Output Impedance <sup>(2)</sup>	$V_Z = V_{\text{REF}}, I_Z = 0.1 \text{ mA to } 15 \text{ mA}$ Frequency = 0 Hz (see <a href="#">Figure 32</a> )		0.25	0.4	$\Omega$

- (1) Deviation of reference input voltage,  $V_{\text{DEV}}$ , is defined as the maximum variation of the reference input voltage over the full temperature range. See the following:



The average temperature coefficient of the reference input voltage,  $\alpha V_{\text{REF}}$ , is defined as:

$$\alpha V_{\text{REF}} \frac{\text{ppm}}{^\circ\text{C}} = \frac{\pm \left( \frac{V_{\text{Max}} - V_{\text{Min}}}{V_{\text{REF}}(\text{at } 25^\circ\text{C})} \right) 10^6}{T_2 - T_1} = \frac{\pm \left( \frac{V_{\text{DEV}}}{V_{\text{REF}}(\text{at } 25^\circ\text{C})} \right) 10^6}{T_2 - T_1}$$

Where:  $T_2 - T_1$  = full temperature change.  $\alpha V_{\text{REF}}$  can be positive or negative depending on whether the slope is positive or negative.

Example:  $V_{\text{DEV}} = 6 \text{ mV}, V_{\text{REF}} = 1240 \text{ mV}, T_2 - T_1 = 125^\circ\text{C}$ .

$$\alpha V_{\text{REF}} = \frac{\left( \frac{6.0 \text{ mV}}{1240 \text{ mV}} \right) 10^6}{125^\circ\text{C}} = +39 \text{ ppm / } ^\circ\text{C}$$

- (2) The dynamic output impedance,  $r_Z$ , is defined as:

$$r_Z = \frac{\Delta V_Z}{\Delta I_Z}$$

When the device is programmed with two external resistors,  $R_1$  and  $R_2$ , (see [Figure 33](#)), the dynamic output impedance of the overall circuit,  $r_Z$ , is defined as:

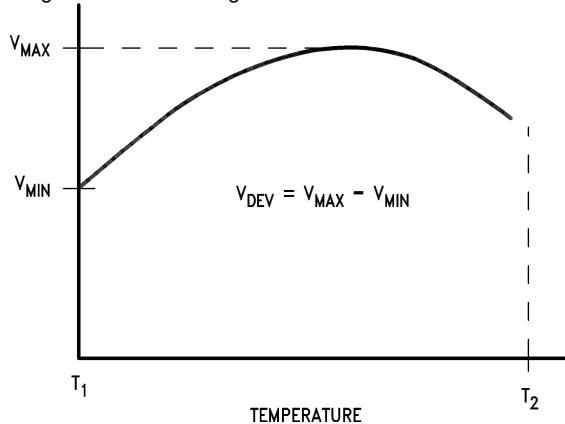
$$r_Z = \frac{\Delta V_Z}{\Delta I_Z} \cong \left[ r_Z \left( 1 + \frac{R_1}{R_2} \right) \right]$$

## 7.7 XB431 Electrical Characteristics

$T_A = 25^\circ\text{C}$  unless otherwise specified

SYMBOL	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{\text{REF}}$	Reference Voltage	$V_Z = V_{\text{REF}}, I_Z = 10 \text{ mA}$ (See <a href="#">Figure 32</a> )	$T_A = 25^\circ\text{C}$	1.228	1.24	1.252
			$T_A = \text{Full Range}$	1.221		1.259
$V_{\text{DEV}}$	Deviation of Reference Input Voltage Over Temperature <sup>(1)</sup>	$V_Z = V_{\text{REF}}, I_Z = 10 \text{ mA},$ $T_A = \text{Full Range}$ (See <a href="#">Figure 32</a> )		4	12	mV
$\frac{\Delta V_{\text{REF}}}{\Delta V_Z}$	Ratio of the Change in Reference Voltage to the Change in Cathode Voltage	$I_Z = 10 \text{ mA}$ (see <a href="#">Figure 33</a> ) $V_Z$ from $V_{\text{REF}}$ to 6 V $R_1 = 10 \text{ k}\Omega, R_2 = \infty$ and $2.6 \text{ k}\Omega$		-1.5	-2.7	mV/V
$I_{\text{REF}}$	Reference Input Current	$R_1 = 1 \text{ k}\Omega, R_2 = \infty$ $I_I = 10 \text{ mA}$ (see <a href="#">Figure 33</a> )		0.15	0.50	$\mu\text{A}$
$\alpha I_{\text{REF}}$	Deviation of Reference Input Current over Temperature	$R_1 = 10 \text{ k}\Omega, R_2 = \infty,$ $I_I = 10 \text{ mA}, T_A = \text{Full Range}$ (see <a href="#">Figure 33</a> )		0.05	0.3	$\mu\text{A}$
$I_{Z(\text{MIN})}$	Minimum Cathode Current for Regulation	$V_Z = V_{\text{REF}}$ (see <a href="#">Figure 32</a> )		55	80	$\mu\text{A}$
$I_{Z(\text{OFF})}$	Off-State Current	$V_Z = 6 \text{ V}, V_{\text{REF}} = 0\text{V}$ (see <a href="#">Figure 34</a> )		0.001	0.1	$\mu\text{A}$
$r_Z$	Dynamic Output Impedance <sup>(2)</sup>	$V_Z = V_{\text{REF}}, I_Z = 0.1\text{mA to } 15\text{mA}$ Frequency = 0 Hz (see <a href="#">Figure 32</a> )		0.25	0.4	$\Omega$

- (1) Deviation of reference input voltage,  $V_{\text{DEV}}$ , is defined as the maximum variation of the reference input voltage over the full temperature range. See the following:



The average temperature coefficient of the reference input voltage,  $\alpha V_{\text{REF}}$ , is defined as:

$$\alpha V_{\text{REF}} \frac{\text{ppm}}{^\circ\text{C}} = \frac{\pm \left( \frac{V_{\text{Max}} - V_{\text{Min}}}{V_{\text{REF}}(\text{at } 25^\circ\text{C})} \right) 10^6}{T_2 - T_1} = \frac{\pm \left( \frac{V_{\text{DEV}}}{V_{\text{REF}}(\text{at } 25^\circ\text{C})} \right) 10^6}{T_2 - T_1}$$

Where:  $T_2 - T_1$  = full temperature change.  $\alpha V_{\text{REF}}$  can be positive or negative depending on whether the slope is positive or negative.  
Example:  $V_{\text{DEV}} = 6 \text{ mV}$ ,  $V_{\text{REF}} = 1240 \text{ mV}$ ,  $T_2 - T_1 = 125^\circ\text{C}$ .

$$\alpha V_{\text{REF}} = \frac{\left( \frac{6.0 \text{ mV}}{1240 \text{ mV}} \right) 10^6}{125^\circ\text{C}} = +39 \text{ ppm / } ^\circ\text{C}$$

- (2) The dynamic output impedance,  $r_Z$ , is defined as:

$$r_Z = \frac{\Delta V_Z}{\Delta I_Z}$$

When the device is programmed with two external resistors,  $R_1$  and  $R_2$ , (see [Figure 33](#)), the dynamic output impedance of the overall circuit,  $r_Z$ , is defined as:

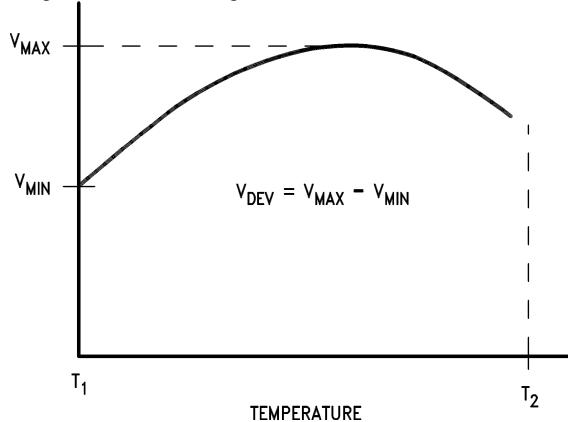
$$r_Z = \frac{\Delta V_Z}{\Delta I_Z} \approx \left[ r_Z \left( 1 + \frac{R_1}{R_2} \right) \right]$$

## 7.8 XB431 Electrical Characteristics

$T_A = 25^\circ\text{C}$  unless otherwise specified

SYMBOL	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
$V_{\text{REF}}$	Reference Voltage	$V_Z = V_{\text{REF}}, I_Z = 10\text{mA}$ (See <a href="#">Figure 32</a> )	$T_A = 25^\circ\text{C}$	1.228	1.24	1.252	V
			$T_A = \text{Full Range}$	1.215		1.265	V
$V_{\text{DEV}}$	Deviation of Reference Input Voltage Over Temperature <sup>(1)</sup>	$V_Z = V_{\text{REF}}, I_Z = 10\text{mA},$ $T_A = \text{Full Range}$ (See <a href="#">Figure 32</a> )		6	20	mV	
$\frac{\Delta V_{\text{REF}}}{\Delta V_Z}$	Ratio of the Change in Reference Voltage to the Change in Cathode Voltage	$I_Z = 10\text{mA}$ (see <a href="#">Figure 33</a> ) $V_Z$ from $V_{\text{REF}}$ to 6 V $R_1 = 10\text{k}\Omega, R_2 = \infty$ and $2.6\text{k}\Omega$		-1.5	-2.7	mV/V	
$I_{\text{REF}}$	Reference Input Current	$R_1 = 10\text{k}\Omega, R_2 = \infty$ $I_I = 10\text{mA}$ (see <a href="#">Figure 33</a> )		0.15	0.5	$\mu\text{A}$	
$\alpha I_{\text{REF}}$	Deviation of Reference Input Current over Temperature	$R_1 = 10\text{k}\Omega, R_2 = \infty,$ $I_I = 10\text{mA}, T_A = \text{Full Range}$ (see <a href="#">Figure 33</a> )		0.1	0.4	$\mu\text{A}$	
$I_{Z(\text{MIN})}$	Minimum Cathode Current for Regulation	$V_Z = V_{\text{REF}}$ (see <a href="#">Figure 32</a> )		55	80	$\mu\text{A}$	
$I_{Z(\text{OFF})}$	Off-State Current	$V_Z = 6\text{V}, V_{\text{REF}} = 0\text{V}$ (see <a href="#">Figure 34</a> )		0.001	0.1	$\mu\text{A}$	
$r_Z$	Dynamic Output Impedance <sup>(2)</sup>	$V_Z = V_{\text{REF}}, I_Z = 0.1\text{mA}$ to $15\text{mA}$ Frequency = 0 Hz (see <a href="#">Figure 32</a> )		0.25	0.4	$\Omega$	

- (1) Deviation of reference input voltage,  $V_{\text{DEV}}$ , is defined as the maximum variation of the reference input voltage over the full temperature range. See the following:



The average temperature coefficient of the reference input voltage,  $\alpha V_{\text{REF}}$ , is defined as:

$$\alpha V_{\text{REF}} \frac{\text{ppm}}{^\circ\text{C}} = \frac{\pm \left( \frac{V_{\text{Max}} - V_{\text{Min}}}{V_{\text{REF}}(\text{at } 25^\circ\text{C})} \right) 10^6}{T_2 - T_1} = \frac{\pm \left( \frac{V_{\text{DEV}}}{V_{\text{REF}}(\text{at } 25^\circ\text{C})} \right) 10^6}{T_2 - T_1}$$

Where:  $T_2 - T_1$  = full temperature change.  $\alpha V_{\text{REF}}$  can be positive or negative depending on whether the slope is positive or negative.  
Example:  $V_{\text{DEV}} = 6\text{mV}$ ,  $V_{\text{REF}} = 1240\text{mV}$ ,  $T_2 - T_1 = 125^\circ\text{C}$ .

$$\alpha V_{\text{REF}} = \frac{\left( \frac{6.0\text{mV}}{1240\text{mV}} \right) 10^6}{125^\circ\text{C}} = +39\text{ ppm / }^\circ\text{C}$$

- (2) The dynamic output impedance,  $r_Z$ , is defined as:

$$r_Z = \frac{\Delta V_Z}{\Delta I_Z}$$

When the device is programmed with two external resistors,  $R_1$  and  $R_2$ , (see [Figure 33](#)), the dynamic output impedance of the overall circuit,  $r_Z$ , is defined as:

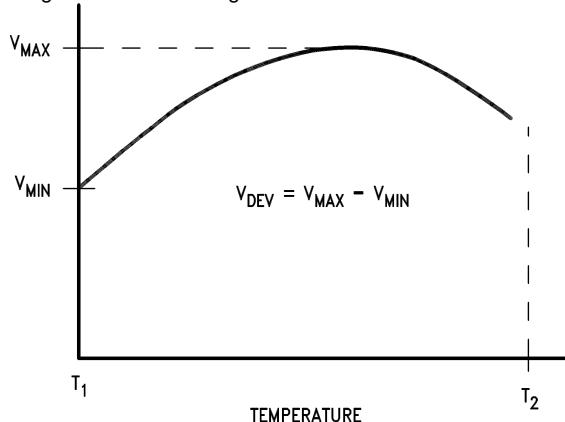
$$r_Z = \frac{\Delta V_Z}{\Delta I_Z} \approx \left[ r_Z \left( 1 + \frac{R_1}{R_2} \right) \right]$$

## 7.9 XB431 Electrical Characteristics

$T_A = 25^\circ\text{C}$  unless otherwise specified

SYMBOL	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
$V_{\text{REF}}$	Reference Voltage	$V_Z = V_{\text{REF}}, I_Z = 10 \text{ mA}$ (See <a href="#">Figure 32</a> )	$T_A = 25^\circ\text{C}$	1.234	1.24	1.246	V
			$T_A = \text{Full Range}$	1.227		1.253	V
$V_{\text{DEV}}$	Deviation of Reference Input Voltage Over Temperature <sup>(1)</sup>	$V_Z = V_{\text{REF}}, I_Z = 10 \text{ mA},$ $T_A = \text{Full Range}$ (See <a href="#">Figure 32</a> )		4	12	mV	
$\frac{\Delta V_{\text{REF}}}{\Delta V_Z}$	Ratio of the Change in Reference Voltage to the Change in Cathode Voltage	$I_Z = 10 \text{ mA}$ (see <a href="#">Figure 33</a> ) $V_Z$ from $V_{\text{REF}}$ to 6 V $R_1 = 10 \text{ k}\Omega, R_2 = \infty$ and $2.6 \text{ k}\Omega$		-1.5	-2.7	mV/V	
$I_{\text{REF}}$	Reference Input Current	$R_1 = 10 \text{ k}\Omega, R_2 = \infty$ $I_I = 10 \text{ mA}$ (see <a href="#">Figure 33</a> )		0.15	0.50	$\mu\text{A}$	
$\alpha I_{\text{REF}}$	Deviation of Reference Input Current over Temperature	$R_1 = 10 \text{ k}\Omega, R_2 = \infty,$ $I_I = 10 \text{ mA}, T_A = \text{Full Range}$ (see <a href="#">Figure 33</a> )		0.05	0.3	$\mu\text{A}$	
$I_{Z(\text{MIN})}$	Minimum Cathode Current for Regulation	$V_Z = V_{\text{REF}}$ (see <a href="#">Figure 32</a> )		55	80	$\mu\text{A}$	
$I_{Z(\text{OFF})}$	Off-State Current	$V_Z = 6 \text{ V}, V_{\text{REF}} = 0\text{V}$ (see <a href="#">Figure 34</a> )		0.001	0.1	$\mu\text{A}$	
$r_Z$	Dynamic Output Impedance <sup>(2)</sup>	$V_Z = V_{\text{REF}}, I_Z = 0.1\text{mA to } 15\text{mA}$ Frequency = 0 Hz (see <a href="#">Figure 32</a> )		0.25	0.4	$\Omega$	

- (1) Deviation of reference input voltage,  $V_{\text{DEV}}$ , is defined as the maximum variation of the reference input voltage over the full temperature range. See the following:



The average temperature coefficient of the reference input voltage,  $\alpha V_{\text{REF}}$ , is defined as:

$$\alpha V_{\text{REF}} \frac{\text{ppm}}{^\circ\text{C}} = \frac{\pm \left( \frac{V_{\text{Max}} - V_{\text{Min}}}{V_{\text{REF}}(\text{at } 25^\circ\text{C})} \right) 10^6}{T_2 - T_1} = \frac{\pm \left( \frac{V_{\text{DEV}}}{V_{\text{REF}}(\text{at } 25^\circ\text{C})} \right) 10^6}{T_2 - T_1}$$

Where:  $T_2 - T_1$  = full temperature change.  $\alpha V_{\text{REF}}$  can be positive or negative depending on whether the slope is positive or negative.  
Example:  $V_{\text{DEV}} = 6 \text{ mV}$ ,  $V_{\text{REF}} = 1240 \text{ mV}$ ,  $T_2 - T_1 = 125^\circ\text{C}$ .

$$\alpha V_{\text{REF}} = \frac{\left( \frac{6.0 \text{ mV}}{1240 \text{ mV}} \right) 10^6}{125^\circ\text{C}} = +39 \text{ ppm / } ^\circ\text{C}$$

- (2) The dynamic output impedance,  $r_Z$ , is defined as:

$$r_Z = \frac{\Delta V_Z}{\Delta I_Z}$$

When the device is programmed with two external resistors,  $R_1$  and  $R_2$ , (see [Figure 33](#)), the dynamic output impedance of the overall circuit,  $r_Z$ , is defined as:

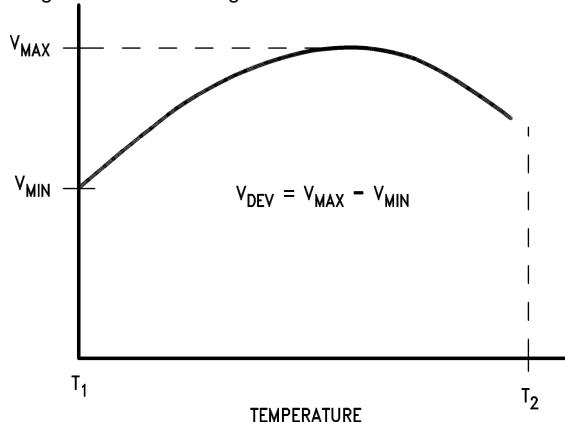
$$r_Z = \frac{\Delta V_Z}{\Delta I_Z} \approx \left[ r_Z \left( 1 + \frac{R_1}{R_2} \right) \right]$$

## 7.10 XB431 Electrical Characteristics

$T_A = 25^\circ\text{C}$  unless otherwise specified

SYMBOL	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
$V_{\text{REF}}$	Reference Voltage	$V_Z = V_{\text{REF}}, I_Z = 10 \text{ mA}$ (See <a href="#">Figure 32</a> )	$T_A = 25^\circ\text{C}$	1.234	1.24	1.246	V
			$T_A = \text{Full Range}$	1.224		1.259	V
$V_{\text{DEV}}$	Deviation of Reference Input Voltage Over Temperature <sup>(1)</sup>	$V_Z = V_{\text{REF}}, I_Z = 10 \text{ mA},$ $T_A = \text{Full Range}$ (See <a href="#">Figure 32</a> )		6	20	mV	
$\frac{\Delta V_{\text{REF}}}{\Delta V_Z}$	Ratio of the Change in Reference Voltage to the Change in Cathode Voltage	$I_Z = 10 \text{ mA}$ (see <a href="#">Figure 33</a> ) $V_Z$ from $V_{\text{REF}}$ to 6V $R_1 = 10 \text{ k}\Omega, R_2 = \infty$ and $2.6 \text{ k}\Omega$		-1.5	-2.7	mV/V	
$I_{\text{REF}}$	Reference Input Current	$R_1 = 10 \text{ k}\Omega, R_2 = \infty$ $I_I = 10 \text{ mA}$ (see <a href="#">Figure 33</a> )		0.15	0.50	$\mu\text{A}$	
$\alpha I_{\text{REF}}$	Deviation of Reference Input Current over Temperature	$R_1 = 10 \text{ k}\Omega, R_2 = \infty,$ $I_I = 10 \text{ mA}, T_A = \text{Full Range}$ (see <a href="#">Figure 33</a> )		0.1	0.4	$\mu\text{A}$	
$I_{Z(\text{MIN})}$	Minimum Cathode Current for Regulation	$V_Z = V_{\text{REF}}$ (see <a href="#">Figure 32</a> )		55	80	$\mu\text{A}$	
$I_{Z(\text{OFF})}$	Off-State Current	$V_Z = 6 \text{ V}, V_{\text{REF}} = 0 \text{ V}$ (see <a href="#">Figure 34</a> )		0.001	0.1	$\mu\text{A}$	
$r_Z$	Dynamic Output Impedance <sup>(2)</sup>	$V_Z = V_{\text{REF}}, I_Z = 0.1 \text{ mA to } 15 \text{ mA}$ Frequency = 0 Hz (see <a href="#">Figure 32</a> )		0.25	0.4	$\Omega$	

- (1) Deviation of reference input voltage,  $V_{\text{DEV}}$ , is defined as the maximum variation of the reference input voltage over the full temperature range. See the following:



The average temperature coefficient of the reference input voltage,  $\alpha V_{\text{REF}}$ , is defined as:

$$\alpha V_{\text{REF}} \frac{\text{ppm}}{^\circ\text{C}} = \frac{\pm \left( \frac{V_{\text{Max}} - V_{\text{Min}}}{V_{\text{REF}}(\text{at } 25^\circ\text{C})} \right) 10^6}{T_2 - T_1} = \frac{\pm \left( \frac{V_{\text{DEV}}}{V_{\text{REF}}(\text{at } 25^\circ\text{C})} \right) 10^6}{T_2 - T_1}$$

Where:  $T_2 - T_1$  = full temperature change.  $\alpha V_{\text{REF}}$  can be positive or negative depending on whether the slope is positive or negative.  
Example:  $V_{\text{DEV}} = 6 \text{ mV}$ ,  $V_{\text{REF}} = 1240 \text{ mV}$ ,  $T_2 - T_1 = 125^\circ\text{C}$ .

$$\alpha V_{\text{REF}} = \frac{\left( \frac{6.0 \text{ mV}}{1240 \text{ mV}} \right) 10^6}{125^\circ\text{C}} = +39 \text{ ppm / } ^\circ\text{C}$$

- (2) The dynamic output impedance,  $r_Z$ , is defined as:

$$r_Z = \frac{\Delta V_Z}{\Delta I_Z}$$

When the device is programmed with two external resistors,  $R_1$  and  $R_2$ , (see [Figure 33](#)), the dynamic output impedance of the overall circuit,  $r_Z$ , is defined as:

$$r_Z = \frac{\Delta V_Z}{\Delta I_Z} \approx \left[ r_Z \left( 1 + \frac{R_1}{R_2} \right) \right]$$

## 7.11 Typical Performance Characteristics

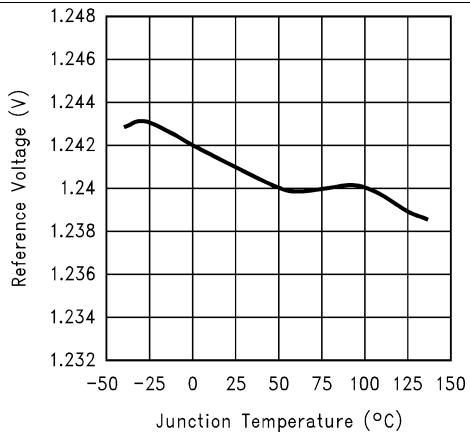


Figure 1. Reference Voltage vs. Junction Temperature

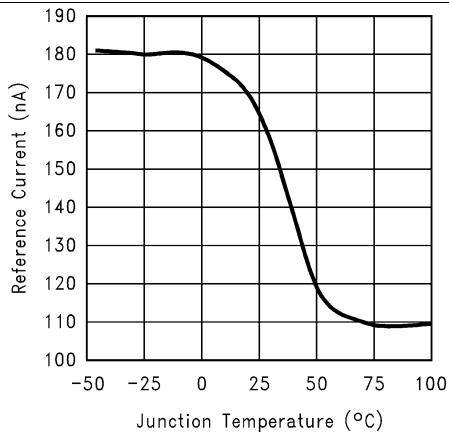


Figure 2. Reference Input Current vs. Junction Temperature

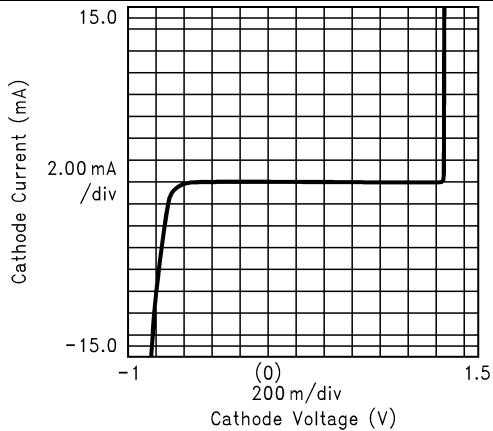


Figure 3. Cathode Current vs. Cathode Voltage 1

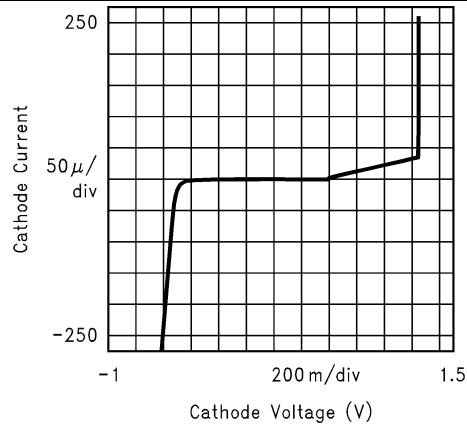


Figure 4. Cathode Current vs. Cathode Voltage 2

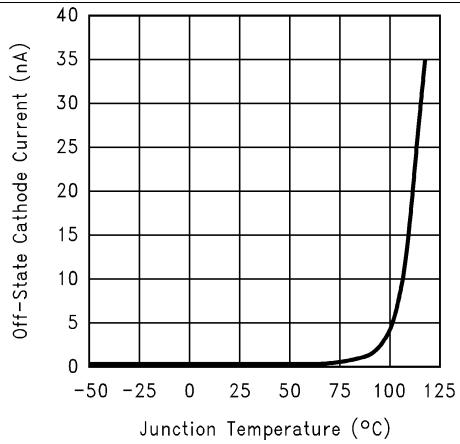


Figure 5. Off-State Cathode Current vs. Junction Temperature

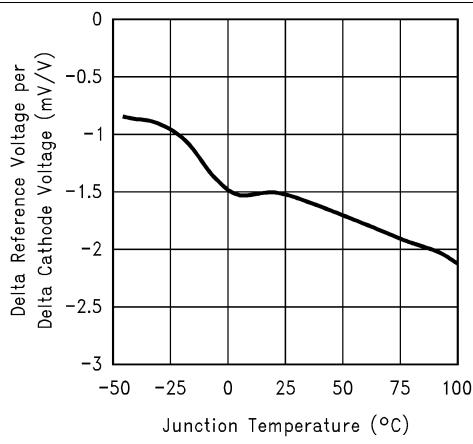
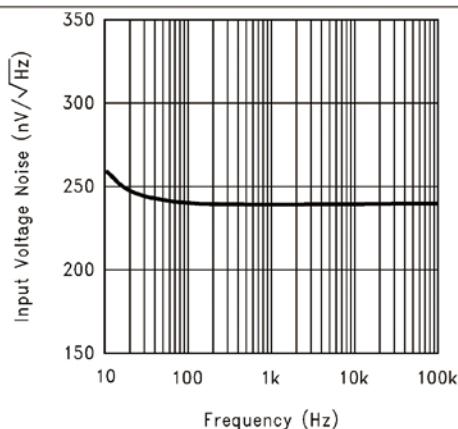
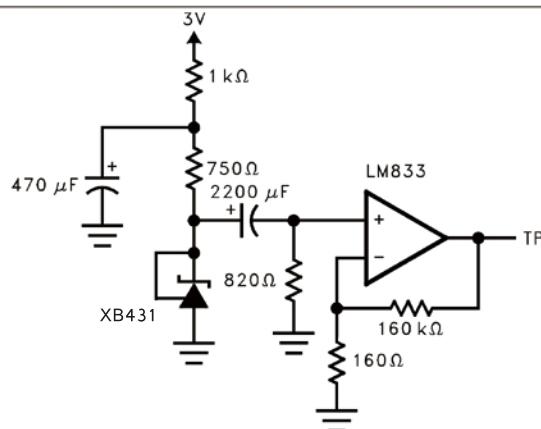


Figure 6. Delta Reference Voltage Per Delta Cathode Voltage vs. Junction Temperature

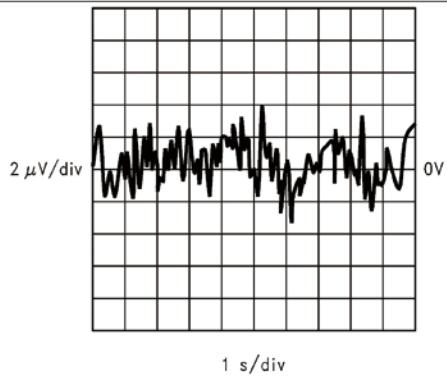
## Typical Performance Characteristics (continued)



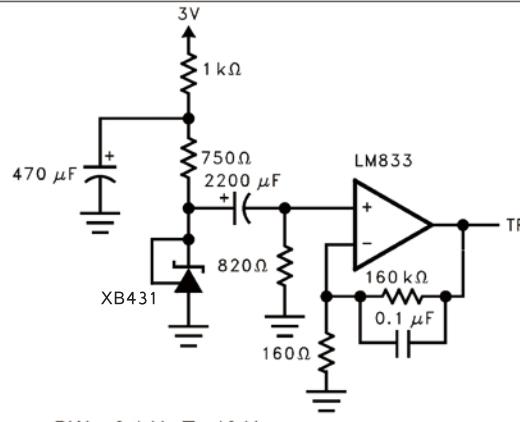
**Figure 7. Input Voltage Noise vs. Frequency**



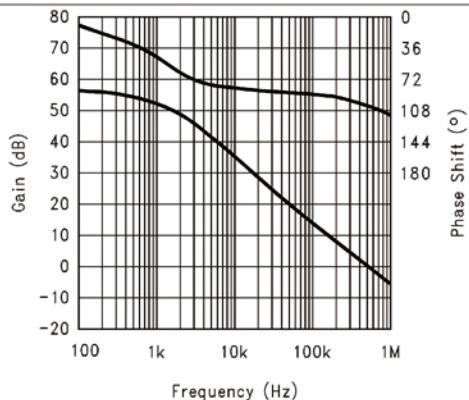
**Figure 8. Test Circuit For Input Voltage Noise vs. Frequency**



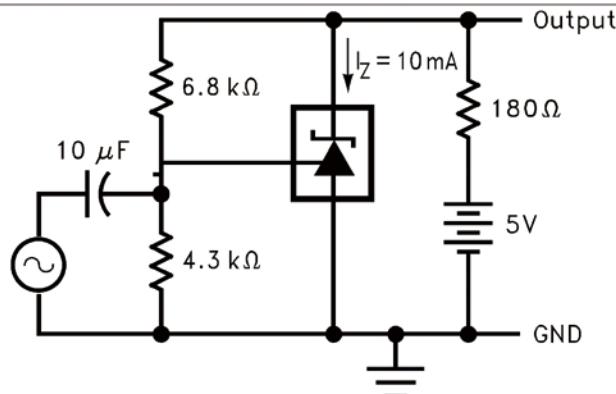
**Figure 9. Low Frequency Peak To Peak Noise**



**Figure 10. Test Circuit For Peak To Peak Noise**



**Figure 11. Small Signal Voltage Gain And Phase Shift vs. Frequency**



**Figure 12. Test Circuit For Voltage Gain And Phase Shift vs. Frequency**

## Typical Performance Characteristics (continued)

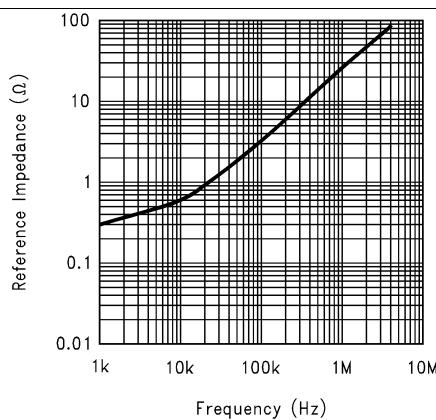


Figure 13. Reference Impedance vs. Frequency

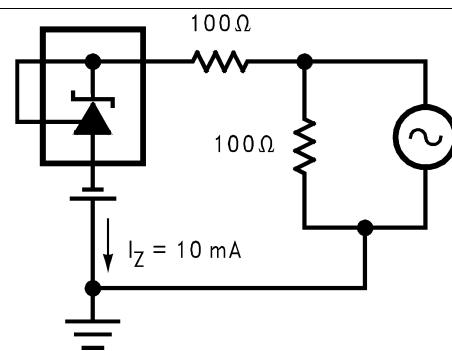


Figure 14. Test Circuit For Reference Impedance vs. Frequency

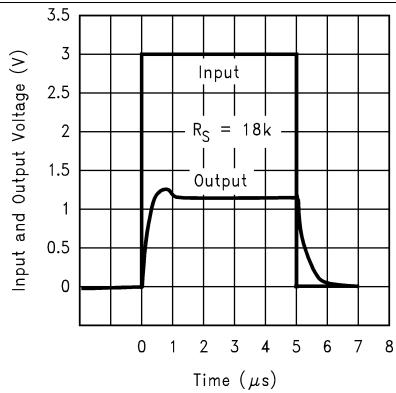


Figure 15. Pulse Response 1

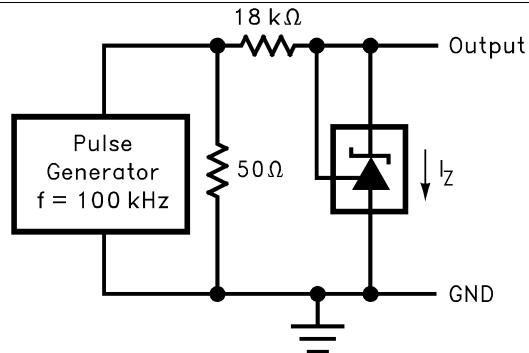


Figure 16. Test Circuit For Pulse Response 1

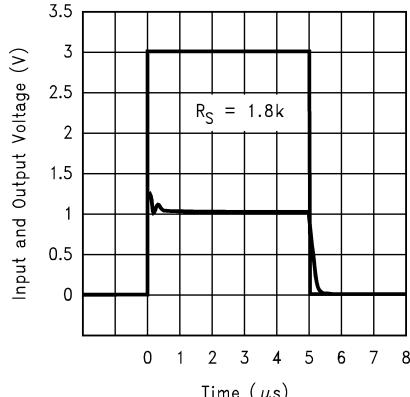


Figure 17. Pulse Response 2

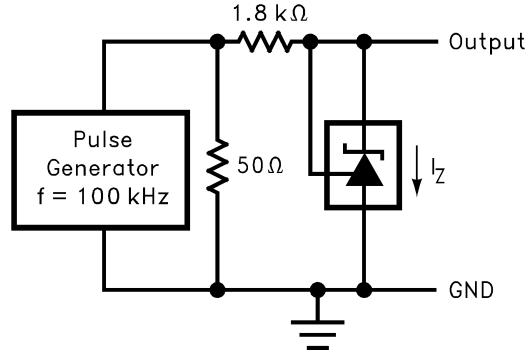
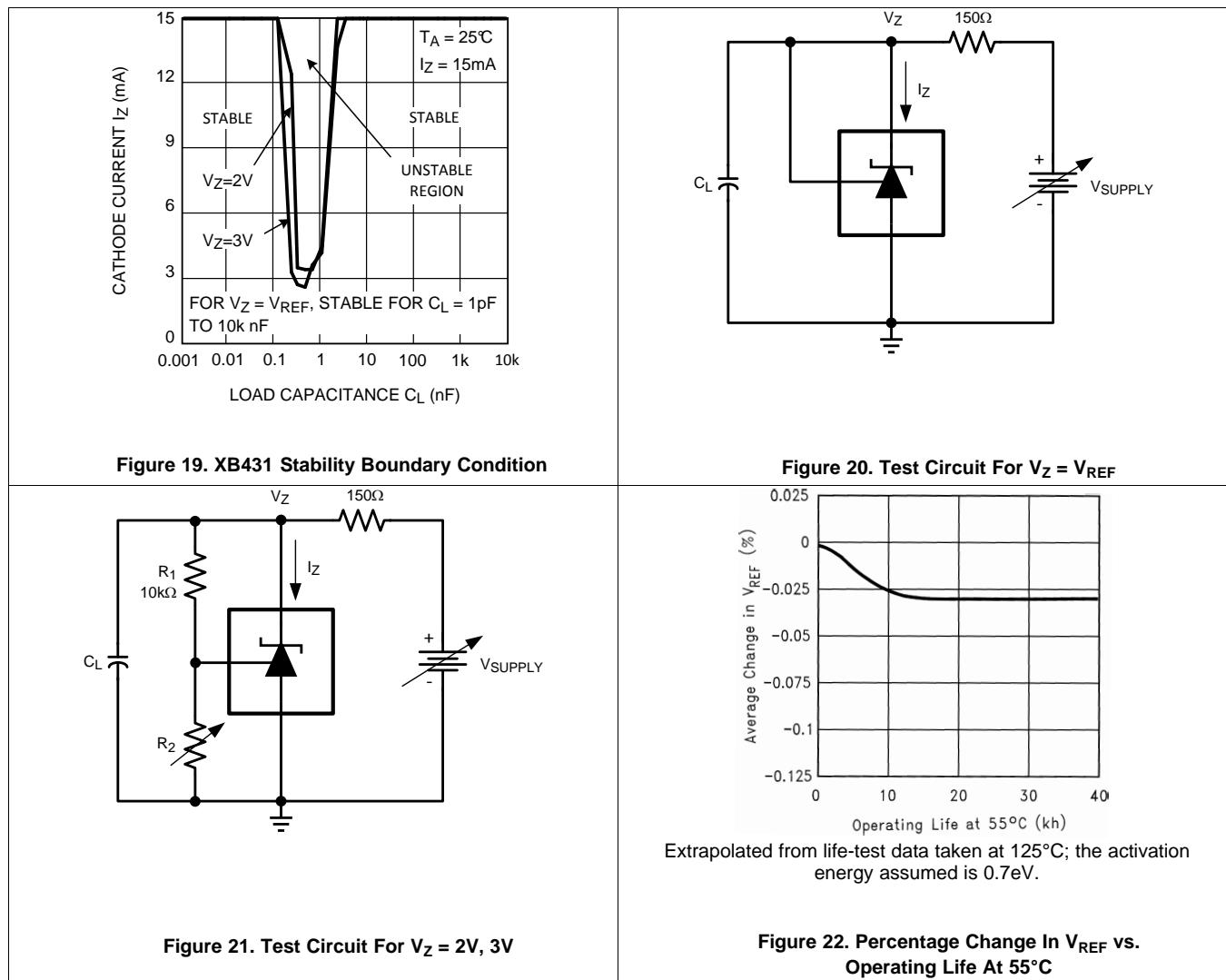


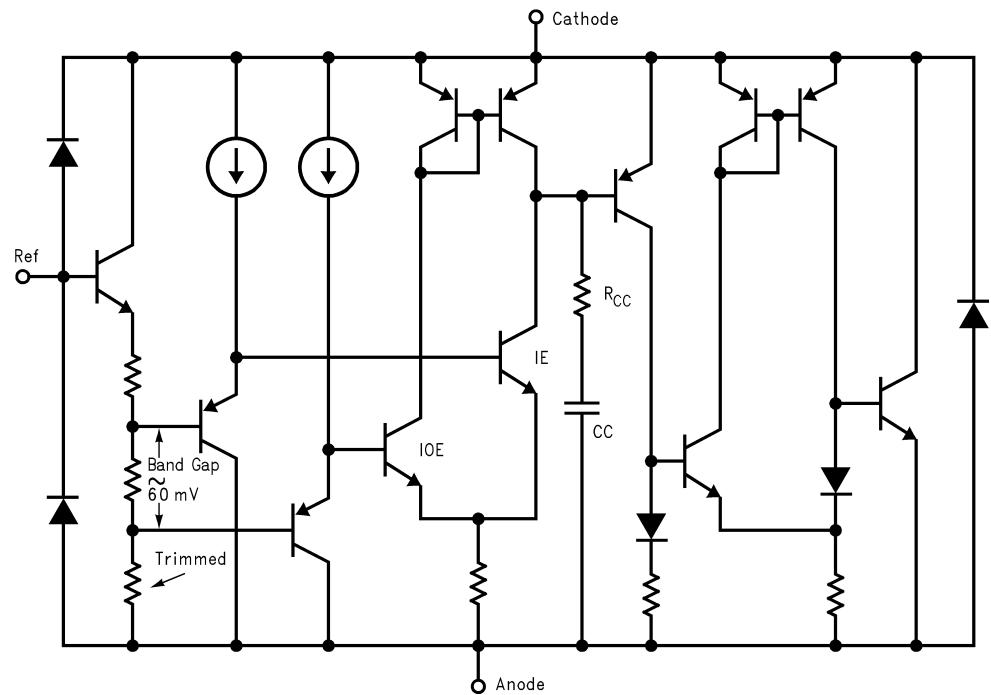
Figure 18. Test Circuit For Pulse Response 2

## Typical Performance Characteristics (continued)



## 8 Detailed Description

### 8.1 Functional Block Diagram



## 9 Application and Implementation

### 9.1 Typical Application

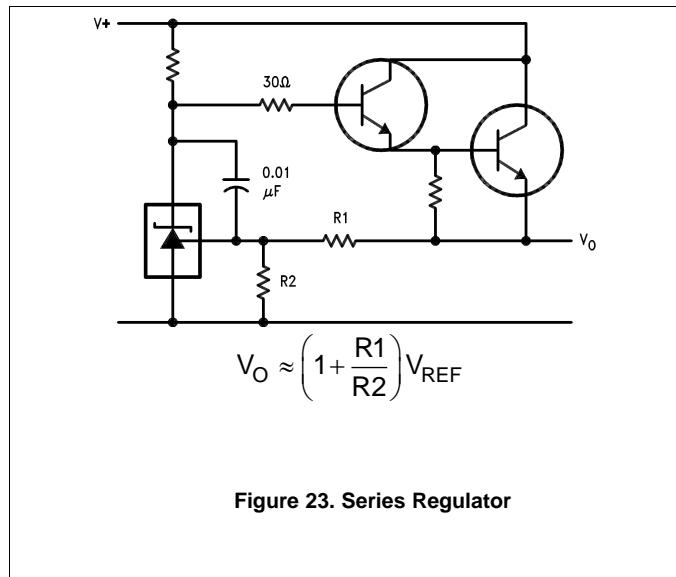


Figure 23. Series Regulator

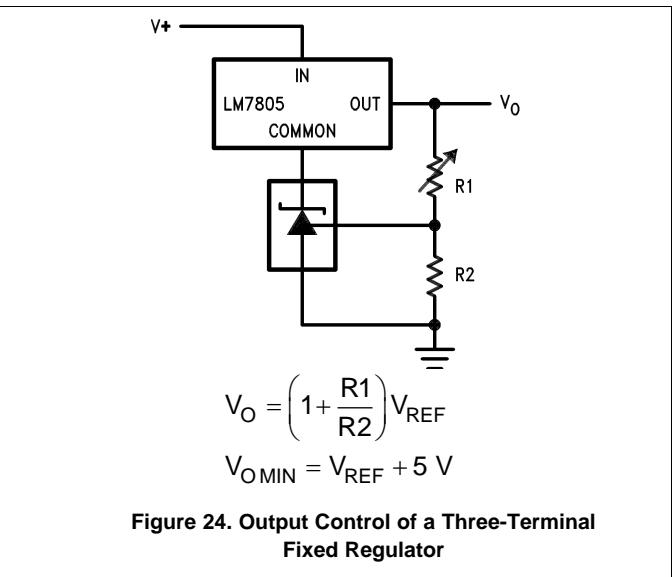


Figure 24. Output Control of a Three-Terminal Fixed Regulator

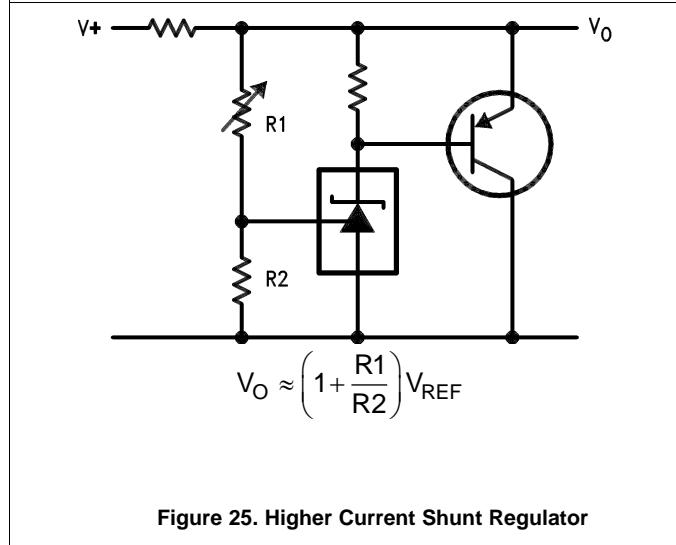


Figure 25. Higher Current Shunt Regulator

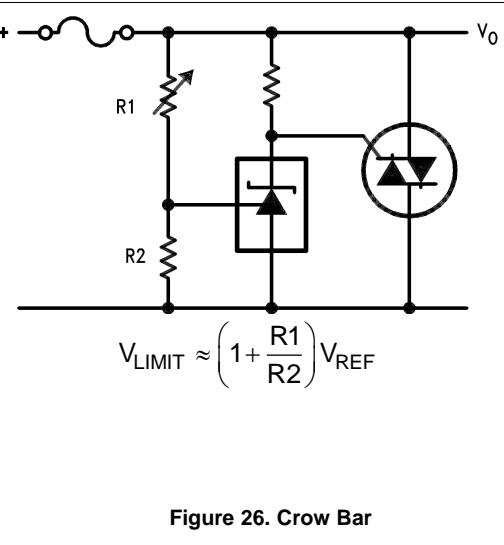
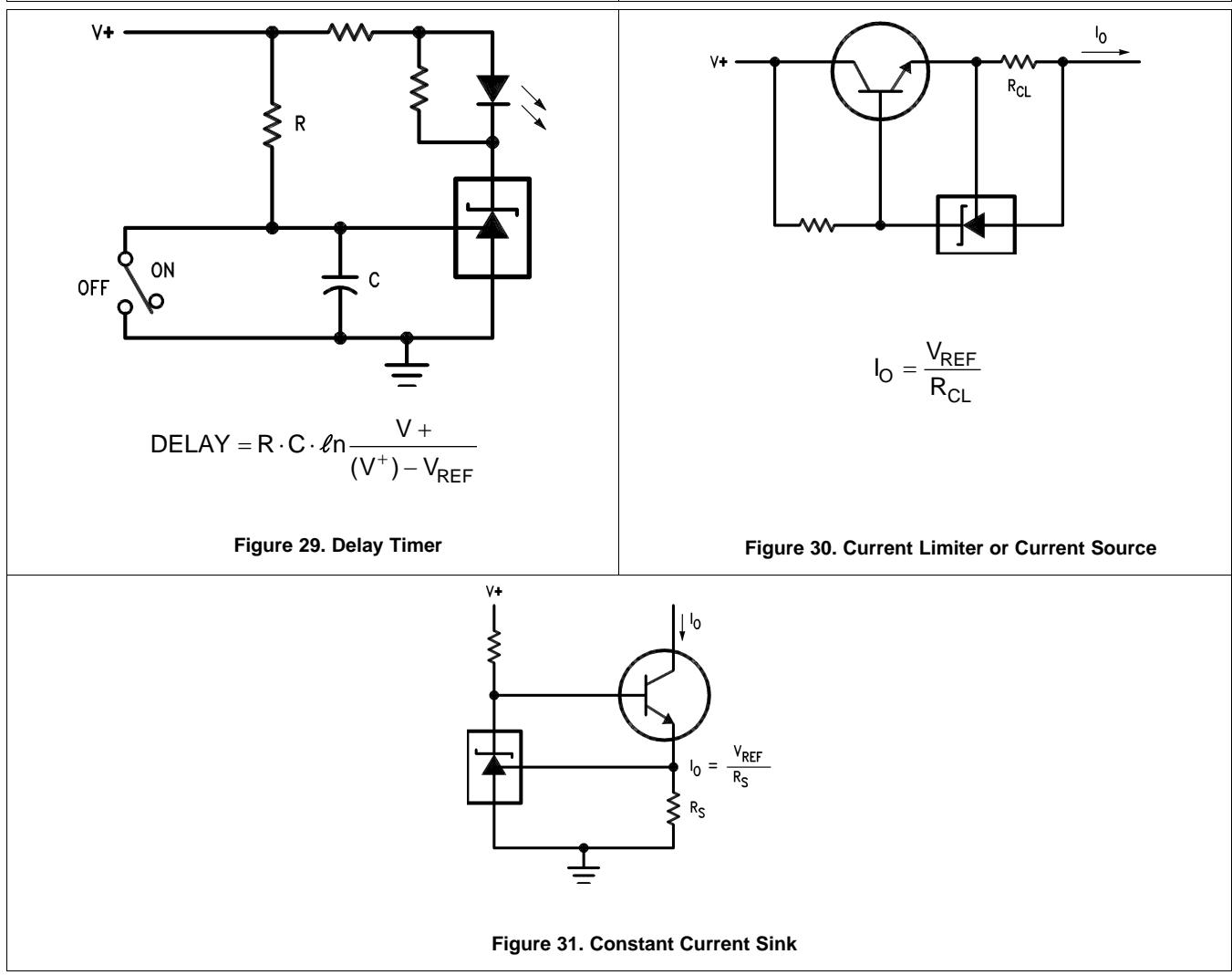
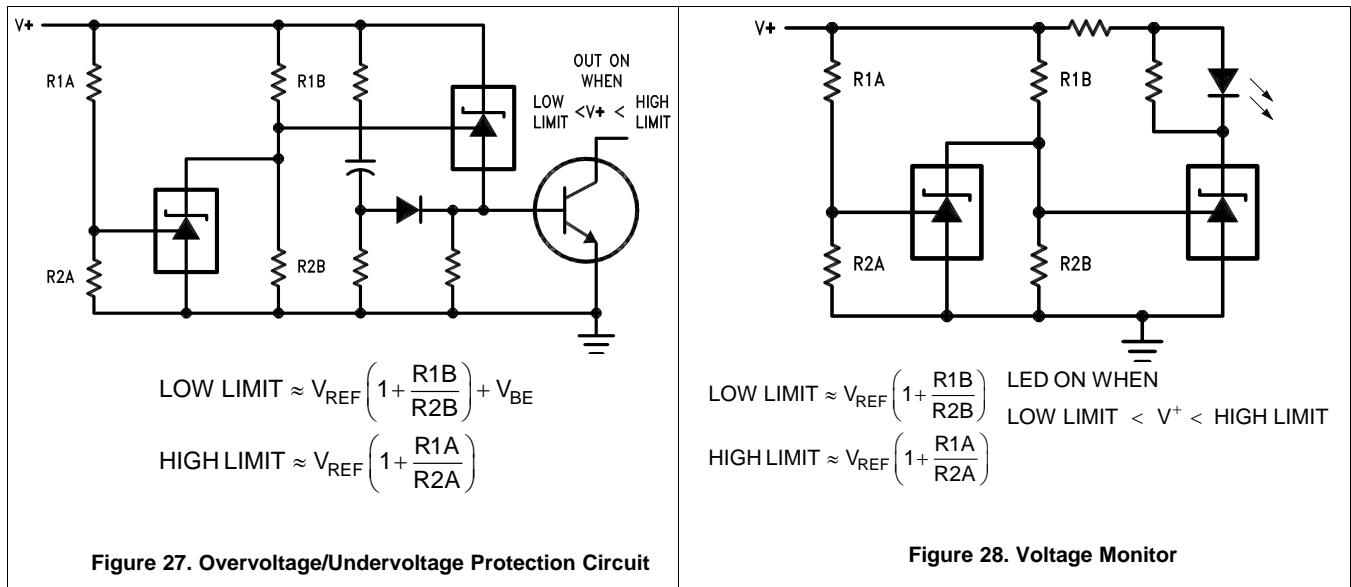
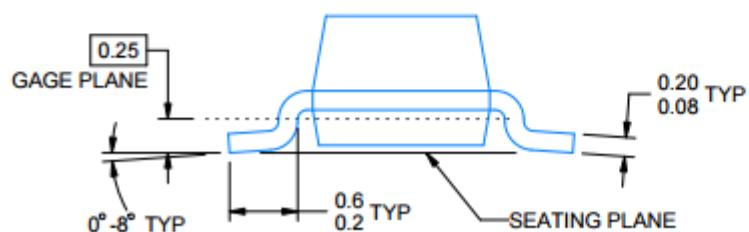
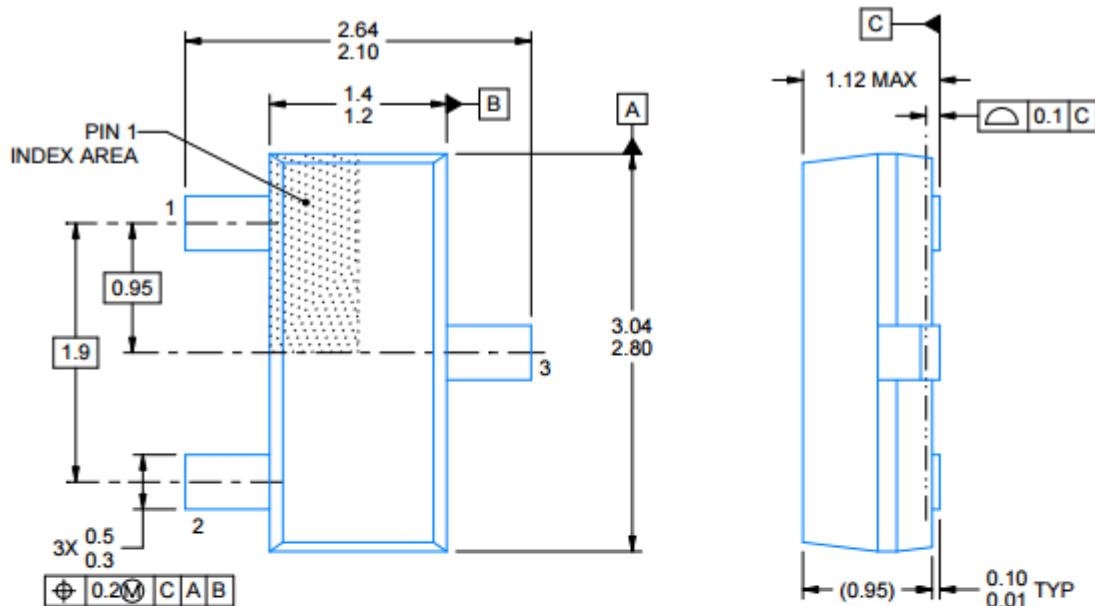


Figure 26. Crow Bar

## Typical Application (continued)





以上信息仅供参考，如需帮助联系客服人员。谢谢 XINLUDA