

## Precision Hall-Effect Latches for Consumer and Industrial Applications

### FEATURES AND BENEFITS

- Symmetrical latch switchpoints
- Superior ruggedness and fault tolerance
- Reverse-polarity and transient protection
- Operation from  $-40^{\circ}\text{C}$  to  $175^{\circ}\text{C}$  junction temperature
- Output short-circuit and overvoltage protection
- Superior temperature stability
- Resistant to physical stress
- High EMC immunity,  $\pm 12$  kV HBM ESD
- Operation from unregulated supplies, 2.8 to 24 V
- Chopper stabilization
- Solid-state reliability
- Industry-standard packages and pinouts

### PACKAGES:

*Not to scale*



### DESCRIPTION

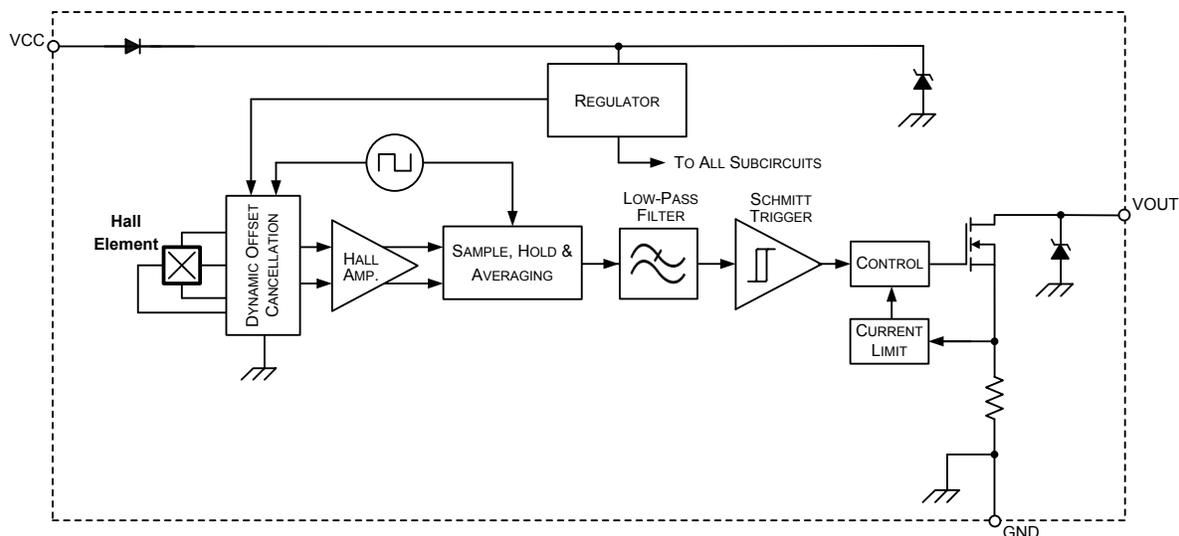
The APS13290 and APS13291 are three-wire, planar Hall-effect sensor integrated circuits (ICs) especially suited for operation over extended temperature ranges (up to  $125^{\circ}\text{C}$ ).

This family of precision Hall-effect latch ICs are ideal for industrial and consumer applications and feature performance enhancements permitting high-temperature operation up to  $175^{\circ}\text{C}$  junction temperatures. In addition, the APS13290/1 include a number of features designed specifically to maximize system robustness, such as reverse-polarity protection, output current limiter, overvoltage, and EMC protection.

The single silicon chip includes: a voltage regulator, a Hall plate, small signal amplifier, chopper stabilization, Schmitt trigger, and a short-circuit-protected open-drain output. A south pole of sufficient strength turns the output on; a north pole of sufficient strength is necessary to turn the output off. The devices include on-board transient protection for all pins, permitting operation directly from a vehicle battery or regulator with supply voltages from 2.8 to 24 V.

Two package styles provide a choice of through-hole or surface mounting. Package type LH is a modified 3-pin SOT23W surface-mount package, while UA is a three-pin ultramini SIP for through-hole mounting. Both packages are lead (Pb) free and RoHS compliant, with 100% matte-tin-plated leadframes.

### Functional Block Diagram



## SELECTION GUIDE

Part Number	Packing <sup>[1]</sup>	Mounting	Magnetic Switchpoints <sup>[2]</sup>		Ambient Temperature, $T_A$
			Operate $B_{OP}$ (G)	Release $B_{RP}$ (G)	
APS13290KLHALX	13-in. reel, 10000 pieces/reel	3-pin SOT23W surface mount	5 to 40	-5 to -40	-40°C to 125°C
APS13290KLHALT <sup>[3]</sup>	7-in. reel, 3000 pieces/reel	3-pin SOT23W surface mount			
APS13290KUAA	Bulk, 500 pieces/bag	3-pin SIP through hole	25 to 80	-25 to -80	
APS13291KLHALX	13-in. reel, 10000 pieces/reel	3-pin SOT23W surface mount			
APS13291KLHALT <sup>[3]</sup>	7-in. reel, 3000 pieces/reel	3-pin SOT23W surface mount			
APS13291KUAA	Bulk, 500 pieces/bag	3-pin SIP through hole			

<sup>[1]</sup> Contact Allegro for additional packing options.

<sup>[2]</sup> Algebraic convention used: (+) south polarity, (-) north polarity.

<sup>[3]</sup> Available through authorized Allegro distributors only.

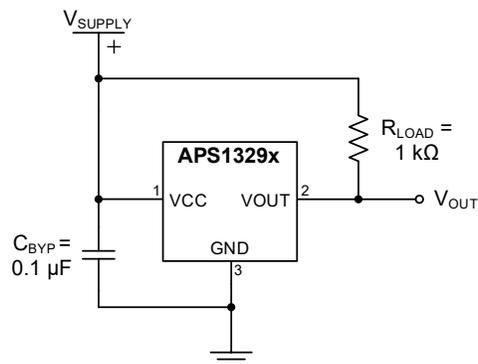


Figure 1: Typical Application Circuit

## ABSOLUTE MAXIMUM RATINGS

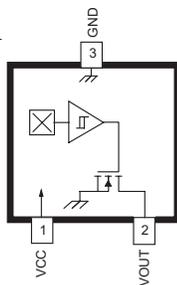
Characteristic	Symbol	Notes	Rating	Units
Forward Supply Voltage <sup>[1]</sup>	V <sub>CC</sub>		30	V
Reverse Supply Voltage <sup>[1]</sup>	V <sub>RCC</sub>		-18	V
Output Off Voltage <sup>[1]</sup>	V <sub>OUT</sub>		30	V
Output Current <sup>[2]</sup>	I <sub>OUT</sub>		60	mA
Reverse Output Current	I <sub>ROUT</sub>		-50	mA
Magnetic Flux Density <sup>[3]</sup>	B		Unlimited	-
Operating Temperature Range	T <sub>A</sub>	Range K	-40 to 125	°C
Maximum Junction Temperature	T <sub>J</sub> (max)		165	°C
		For 500 hours	175	°C
Storage Temperature	T <sub>stg</sub>		-65 to 170	°C
ESD Voltage	V <sub>ESD</sub> (HBM)	Human Body Model according to AEC-Q100-002	±12	kV
	V <sub>ESD</sub> (CDM)	Charged Device Model according to AEC-Q100-011	±1	kV

<sup>[1]</sup> This rating does not apply to extremely short voltage transients such as load dump and/or ESD. Those events have individual ratings, specific to the respective transient voltage event.

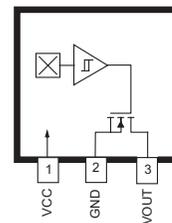
<sup>[2]</sup> Through short-circuit current limiting device.

<sup>[3]</sup> Guaranteed by design.

## PINOUT DIAGRAMS AND TERMINAL LIST TABLE



Package LH



Package UA

### Terminal List

Name	Description	Number	
		Package LH	Package UA
VCC	Connects power supply to chip	1	1
VOUT	Output from circuit	2	3
GND	Ground	3	2

**ELECTRICAL CHARACTERISTICS:** Valid over full operating voltage, ambient temperature range  $T_A = -40^\circ\text{C}$  to  $125^\circ\text{C}$ , and with  $C_{BYP} = 0.1 \mu\text{F}$  (unless otherwise specified)

Characteristics	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit [2]
<b>ELECTRICAL CHARACTERISTICS</b>						
Forward Supply Voltage	$V_{CC}$	Operating, $T_J < 175^\circ\text{C}$	2.8	–	24	V
Supply Current	$I_{CC}$		1	2	3	mA
Output Leakage Current	$I_{OUTOFF}$	$V_{OUTOFF} = 24 \text{ V}$ , $B < B_{RP}$	–	–	10	$\mu\text{A}$
Output Saturation Voltage	$V_{OUT(SAT)}$	$I_{OUT} = 20 \text{ mA}$ , $B > B_{OP}$	–	200	500	mV
Output Off Voltage	$V_{OUTOFF}$	$B < B_{RP}$	–	–	24	V
Power-On Time [3]	$t_{ON}$	$V_{CC} \geq V_{CC(\text{min})}$ , $B < B_{RP(\text{min})} - 10 \text{ G}$ , $B > B_{OP(\text{max})} + 10 \text{ G}$	–	–	25	$\mu\text{s}$
Power-On State, Output [3]	POS	$V_{CC} \geq V_{CC(\text{min})}$ , $t < t_{ON}$	Low			–
Chopping Frequency	$f_C$		–	800	–	kHz
Output Rise Time [4]	$t_r$	$R_{LOAD} = 1 \text{ k}\Omega$ , $C_L = 20 \text{ pF}$	–	0.2	2	$\mu\text{s}$
Output Fall Time [4]	$t_f$	$R_{LOAD} = 1 \text{ k}\Omega$ , $C_L = 20 \text{ pF}$	–	0.1	2	$\mu\text{s}$
<b>TRANSIENT PROTECTION CHARACTERISTICS</b>						
Output Short-Circuit Current Limit	$I_{OM}$		30	–	60	mA
Output Zener Clamp Voltage	$V_{Z\text{output}}$	$I_{OUT} = 3 \text{ mA}$ , $T_A = 25^\circ\text{C}$ , Output Off	30	–	–	V
Reverse Battery Current	$I_{RCC}$	$V_{RCC} = -18 \text{ V}$ , $T_A = 25^\circ\text{C}$	–	–	–5	mA
Supply Zener Clamp Voltage	$V_Z$	$I_{CC} = I_{CC(\text{max})} + 3 \text{ mA}$ , $T_A = 25^\circ\text{C}$	30	–	–	V
<b>MAGNETIC CHARACTERISTICS</b>						
Operate Point	$B_{OP}$	APS13290	5	20	40	G
		APS13291	25	50	80	G
Release Point	$B_{RP}$	APS13290	–40	–20	–5	G
		APS13291	–80	–50	–25	G
Hysteresis	$B_{HYS}$	APS13290	10	40	80	G
		APS13291	50	100	160	G
Symmetry	$B_{SYM}$	$B_{OP} + B_{RP}$	–27.5	–	27.5	G
Magnetic Offset	$B_{OFF}$	$(B_{OP} + B_{RP}) / 2$	–13.75	–	13.75	G

[1] Typical data are at  $T_A = 25^\circ\text{C}$  and  $V_{CC} = 12 \text{ V}$ .

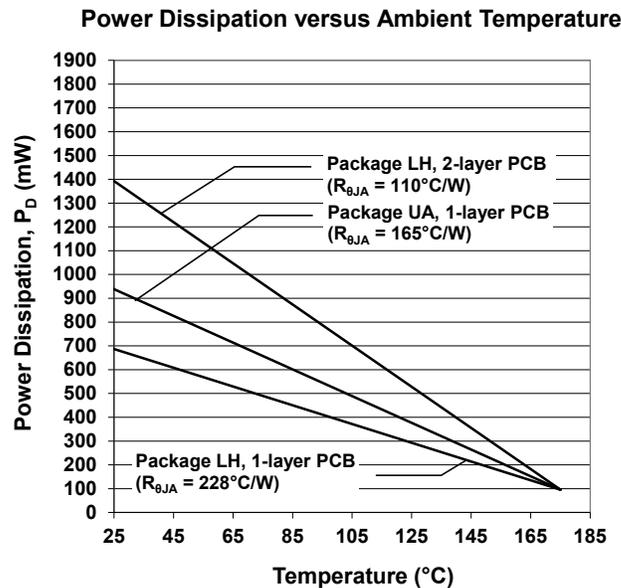
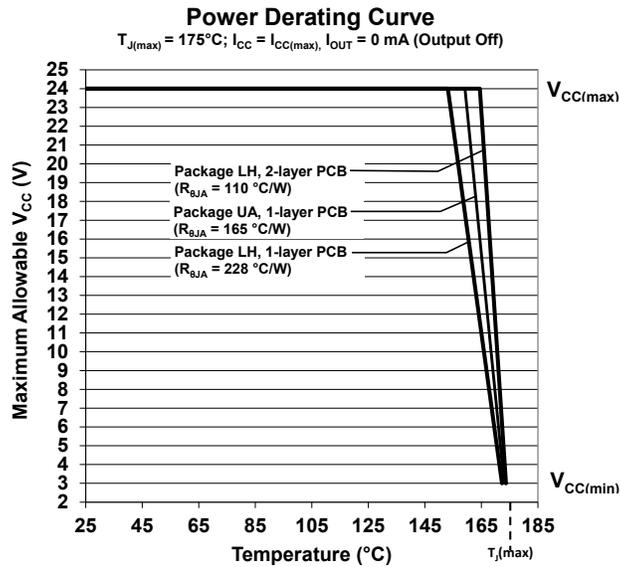
[2] 1 G (gauss) = 0.1 mT (millitesla).

[3] Guaranteed by device design and characterization.

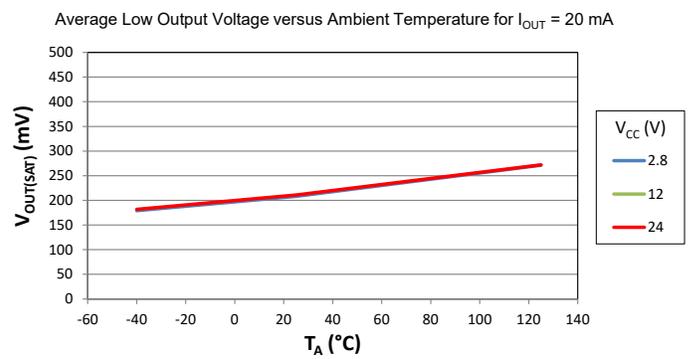
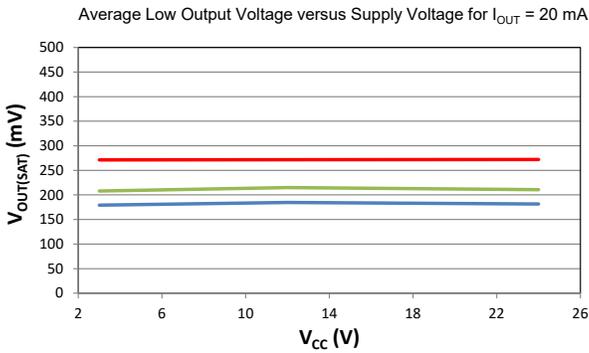
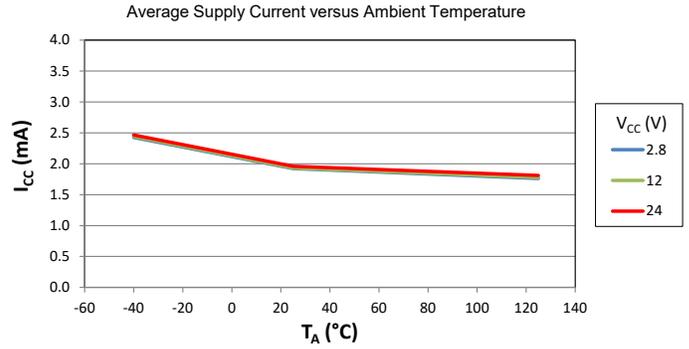
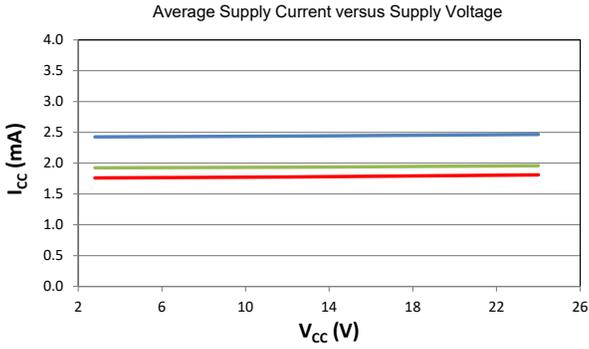
[4]  $C_L$  = oscilloscope probe capacitance.

**THERMAL CHARACTERISTICS:** May require derating at maximum conditions; see application information

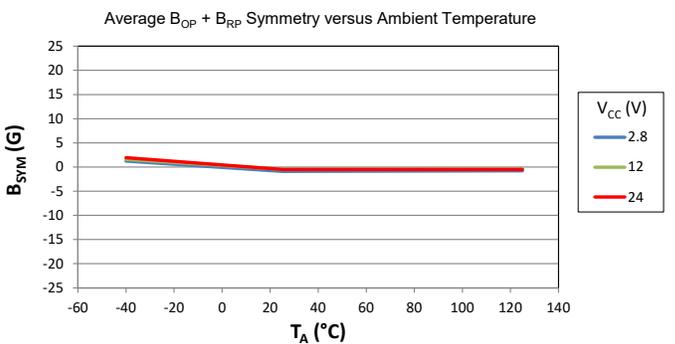
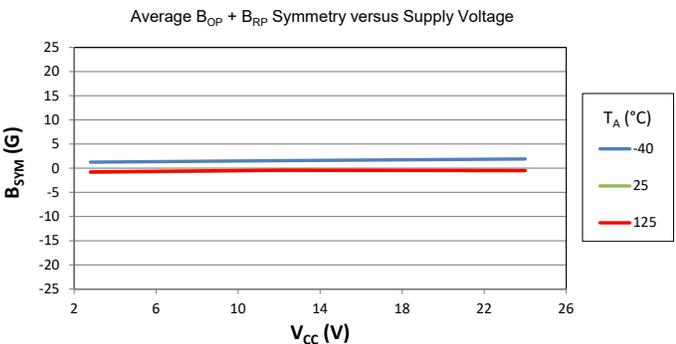
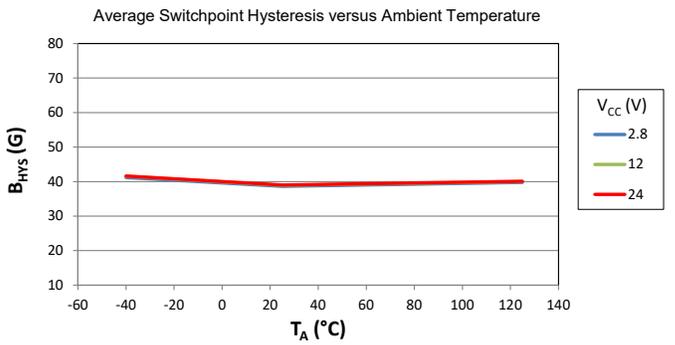
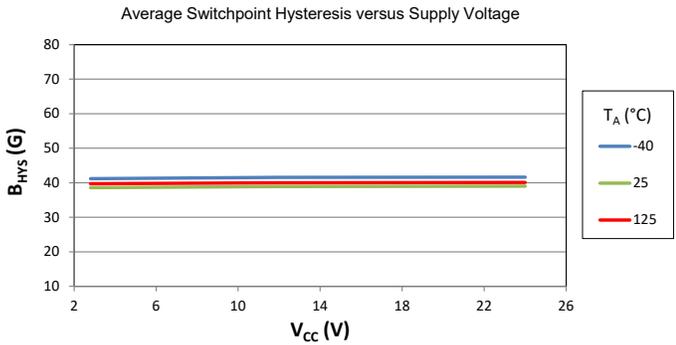
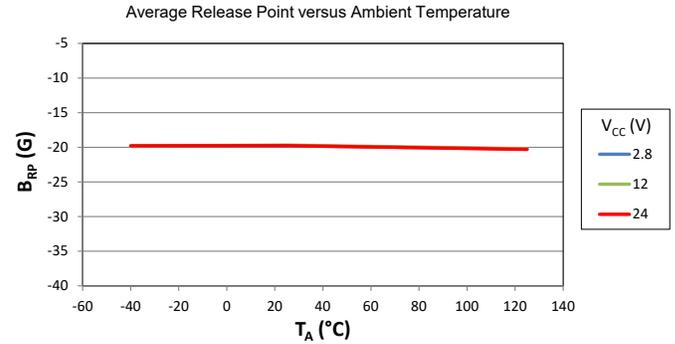
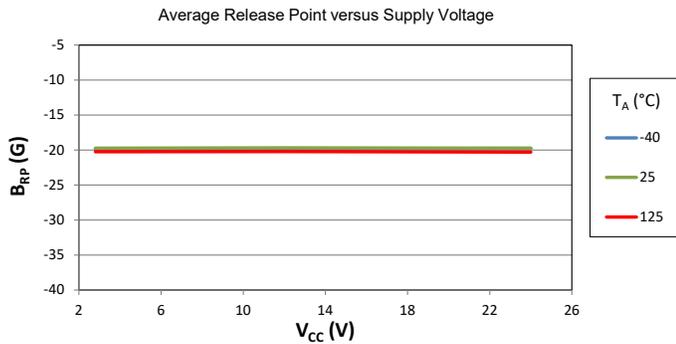
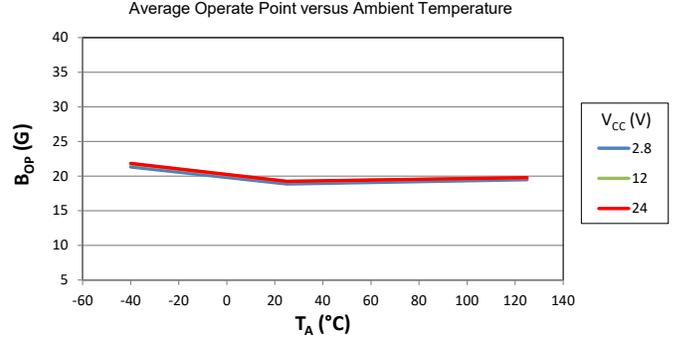
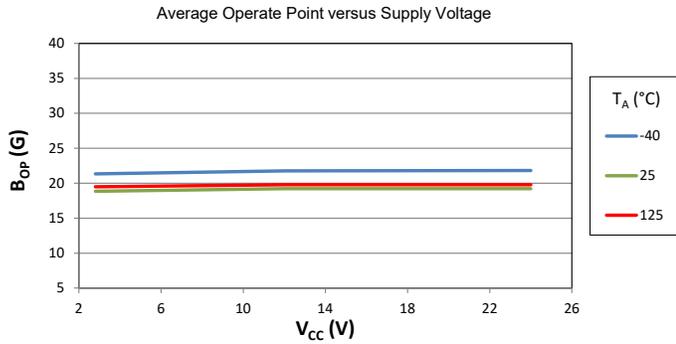
Characteristic	Symbol	Test Conditions	Value	Units
Package Thermal Resistance	$R_{\theta JA}$	Package LH, 1-layer PCB with copper limited to solder pads	228	$^{\circ}\text{C}/\text{W}$
		Package LH, 2-layer PCB with 0.463 in <sup>2</sup> of copper area each side connected by thermal vias	110	$^{\circ}\text{C}/\text{W}$
		Package UA, 1-layer PCB with copper limited to solder pads	165	$^{\circ}\text{C}/\text{W}$



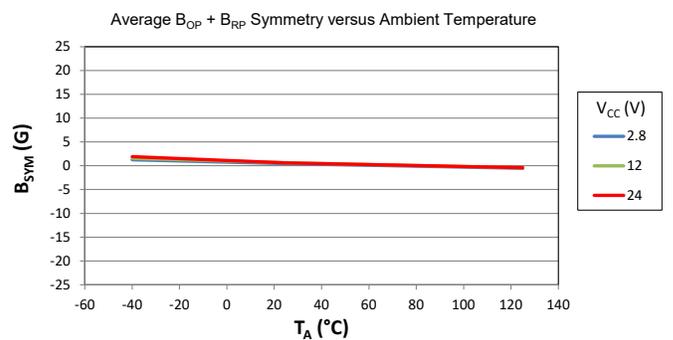
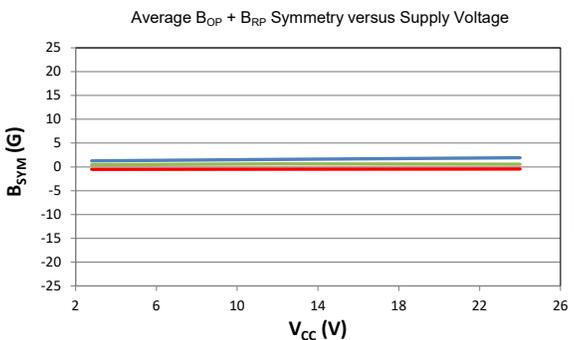
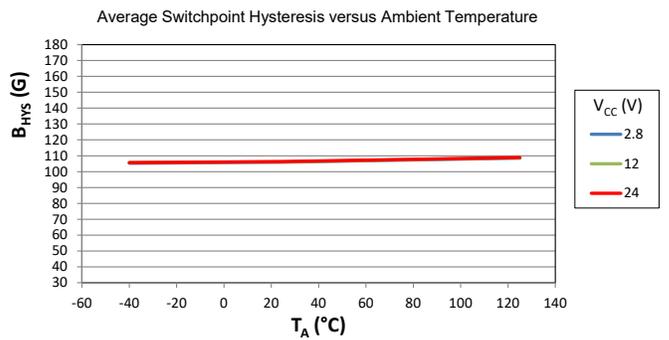
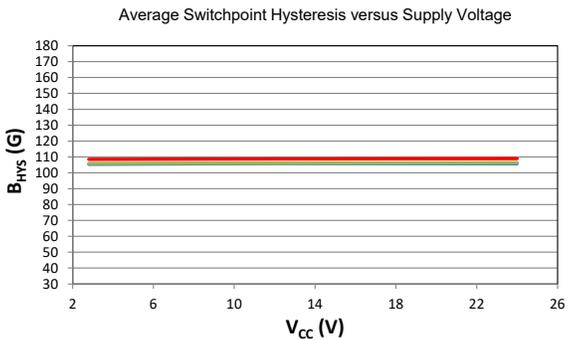
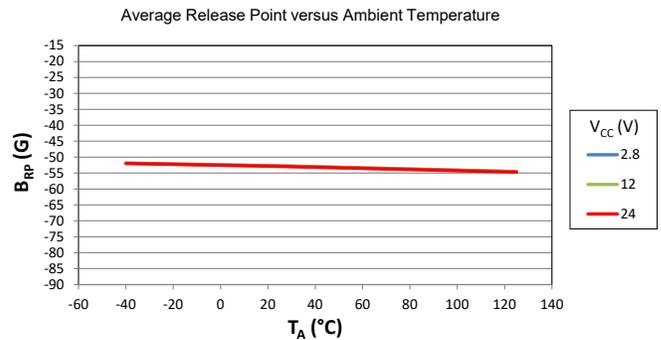
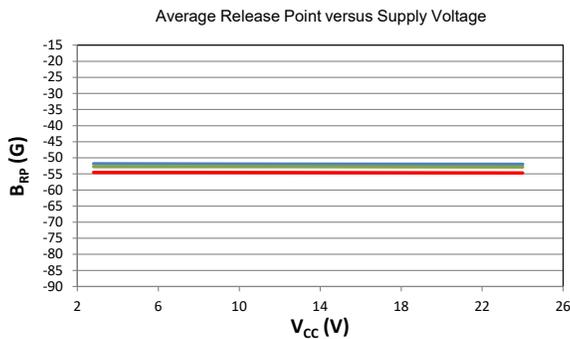
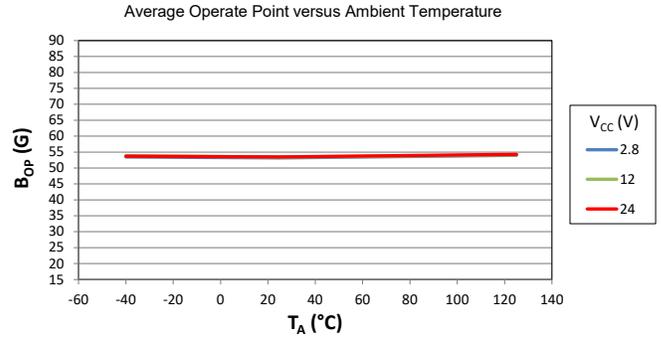
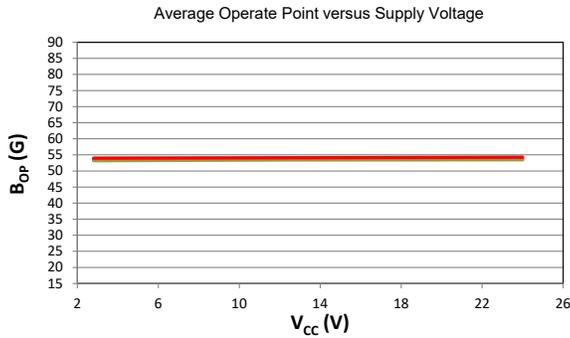
**CHARACTERISTIC PERFORMANCE DATA  
Electrical Characteristics**



CHARACTERISTIC PERFORMANCE DATA (continued)  
APS13290 Magnetic Characteristics



CHARACTERISTIC PERFORMANCE DATA (continued)  
APS13291 Magnetic Characteristics

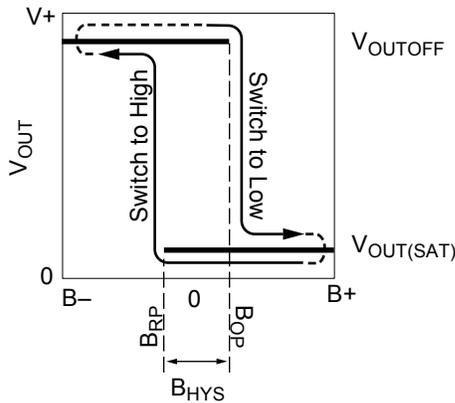


**FUNCTIONAL DESCRIPTION**

**OPERATION**

The output of these devices switches low (turns on) when a magnetic field perpendicular to the Hall element exceeds the operate point threshold,  $B_{OP}$  (see Figure 2). After turn-on, the output voltage is  $V_{OUT(SAT)}$ . The output transistor is capable of continuously sinking up to 30 mA. When the magnetic field is reduced below the release point,  $B_{RP}$ , the device output goes high (turns off) to  $V_{OUT(OFF)}$ . The difference in the magnetic operate and release points is the hysteresis,  $B_{HYS}$ , of the device. This built-in hysteresis allows clean switching of the output even in the presence of external mechanical vibration and electrical noise.

Removal of the magnetic field will leave the device output latched on if the last crossed switchpoint is  $B_{OP}$ , or latched off if the last crossed switchpoint is  $B_{RP}$ .



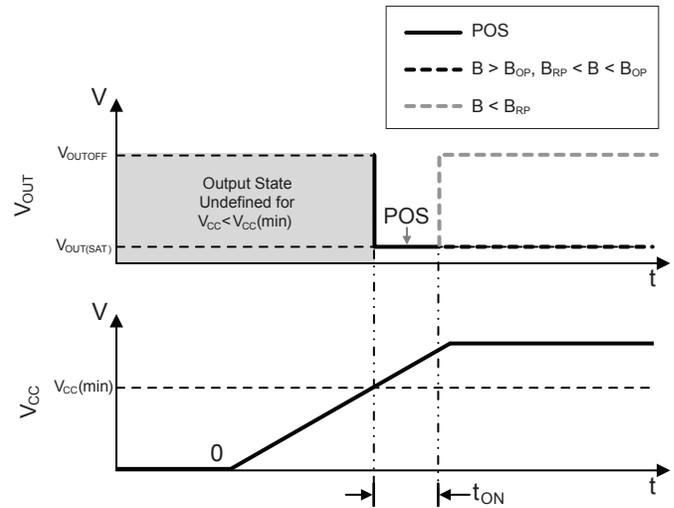
**Figure 2: Switching Behavior of Latches**

On the horizontal axis, the B+ direction indicates increasing south polarity magnetic field strength, and the B- direction indicates increasing north polarity field strength.

**POWER-ON BEHAVIOR**

Device power-on occurs once  $t_{ON}$  has elapsed. During the time prior to  $t_{ON}$ , and after  $V_{CC} \geq V_{CC(min)}$ , the output state is  $V_{OUT(SAT)}$  (Low). After  $t_{ON}$  has elapsed, the output will correspond with the applied magnetic field for  $B > B_{OP}$  or  $B < B_{RP}$ . See Figure 3 for an example.

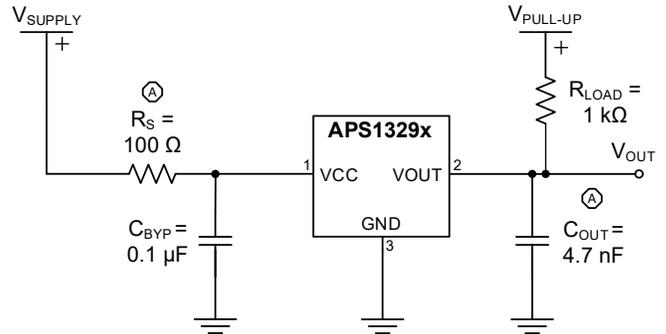
Powering-on the device in the hysteresis range (less than  $B_{OP}$  and higher than  $B_{RP}$ ) will give an output state of  $V_{OUT(SAT)}$ . The correct state is attained after the first excursion beyond  $B_{OP}$  or  $B_{RP}$ .



**Figure 3: Power-On Timing Diagram**

**APPLICATIONS**

It is strongly recommended that an external bypass capacitor be connected (in close proximity to the Hall element) between the supply and ground of the device to guarantee correct performance under harsh environmental conditions and to reduce noise from internal circuitry. As is shown in Figure 1: Typical Application Circuit, a 0.1  $\mu\text{F}$  capacitor is typical. In applications where maximum robustness is required, such as long-life industrial motors, additional measures may be taken. In Figure 4: Enhanced Protection Circuit, a resistor in series with the VCC pin and a capacitor on the VOUT pin enhance the EMC immunity of the device. It is up to the user to fully qualify the Allegro sensor IC in their end system to ensure they achieve their system requirements.



Ⓐ  $R_S$  and  $C_{OUT}$  are recommended for maximum robustness in an automotive environment.

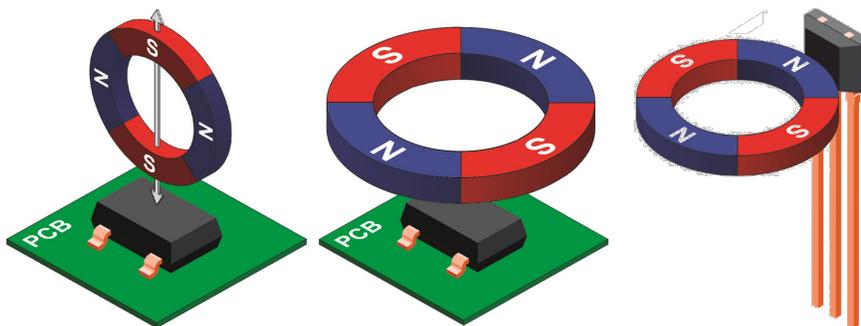
**Figure 4: Enhanced Protection Circuit**

These devices are sensitive in the direction perpendicular to the branded face, as depicted in Figure 5. For further information, extensive applications information on magnets and Hall-effect sensors is available in:

- *Hall-Effect IC Applications Guide, AN27701,*
- *Hall-Effect Devices: Guidelines for Designing Subassemblies Using Hall-Effect Devices AN27703.1*
- *Soldering Methods for Allegro’s Products – SMD and Through-Hole, AN26009*

All are provided on the Allegro website:

[www.allegromicro.com](http://www.allegromicro.com)



**Figure 5: Sensing Configurations**

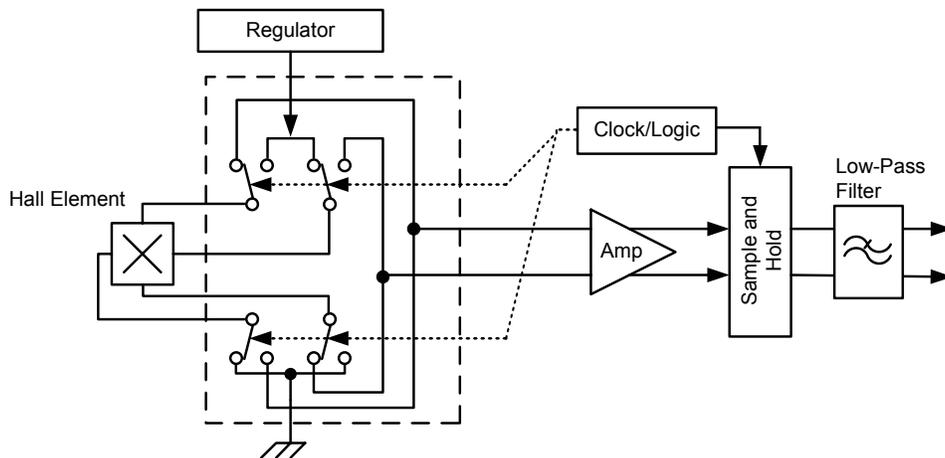
## CHOPPER STABILIZATION

A limiting factor for switchpoint accuracy when using Hall-effect technology is the small signal voltage developed across the Hall plate. This voltage is proportionally small relative to the offset that can be produced at the output of the Hall sensor. This makes it difficult to process the signal and maintain an accurate, reliable output over the specified temperature and voltage range. Chopper stabilization is a proven approach used to minimize Hall offset.

The Allegro technique, dynamic quadrature offset cancellation, removes key sources of the output drift induced by temperature and package stress. This offset reduction technique is based on a signal modulation-demodulation process. Figure 6 illustrates how it is implemented.

The undesired offset signal is separated from the magnetically induced signal in the frequency domain through modulation.

The subsequent demodulation acts as a modulation process for the offset, causing the magnetically induced signal to recover its original spectrum at baseband while the DC offset becomes a high-frequency signal. Then, using a low-pass filter, the signal passes while the modulated DC offset is suppressed. Allegro's innovative chopper stabilization technique uses a high-frequency clock. The high-frequency operation allows a greater sampling rate that produces higher accuracy, reduced jitter, and faster signal processing. Additionally, filtering is more effective and results in a lower noise analog signal at the sensor output. Devices such as the APS13290 and APS13291 that use this approach have an extremely stable quiescent Hall output voltage, are immune to thermal stress, and have precise recoverability after temperature cycling. This technique is made possible through the use of a BiCMOS process which allows the use of low offset and low noise amplifiers in combination with high-density logic and sample-and-hold circuits.



**Figure 6: Model of Chopper Stabilization  
(Dynamic Offset Cancellation)**

## POWER DERATING

The device must be operated below the maximum junction temperature of the device,  $T_J(\text{max})$ . Under certain combinations of peak conditions, reliable operation may require derating supplied power or improving the heat dissipation properties of the application. This section presents a procedure for correlating factors affecting operating  $T_J$ . (Thermal data is also available on the Allegro MicroSystems website.)

The Package Thermal Resistance,  $R_{\theta JA}$ , is a figure of merit summarizing the ability of the application and the device to dissipate heat from the junction (die), through all paths to the ambient air. Its primary component is the Effective Thermal Conductivity,  $K$ , of the printed circuit board, including adjacent devices and traces. Radiation from the die through the device case,  $R_{\theta JC}$ , is a relatively small component of  $R_{\theta JA}$ . Ambient air temperature,  $T_A$ , and air motion are significant external factors, damped by overmolding.

The resulting power dissipation capability directly reflects upon the ability of the device to withstand extreme operating conditions. The junction temperature mission profile specified in the Absolute Maximum Ratings table designates a total operating life capability based on qualification for the most extreme conditions, where  $T_J$  may reach 175°C.

The silicon IC is heated internally when current is flowing into the VCC terminal. When the output is on, current sinking into the VOUT terminal generates additional heat. This may increase the junction temperature,  $T_J$ , above the surrounding ambient temperature. The APS13290 and APS13291 are permitted to operate up to  $T_J = 175^\circ\text{C}$ . As mentioned above, an operating device will increase  $T_J$  according to equations 1, 2, and 3 below. This allows an estimation of the maximum ambient operating temperature.

$$P_D = V_{IN} \times I_{IN} \quad (1)$$

$$\Delta T = P_D \times R_{\theta JA} \quad (2)$$

$$T_J = T_A + \Delta T \quad (3)$$

For example, given common conditions such as:  $T_A = 25^\circ\text{C}$ ,  $V_{CC} = 12\text{ V}$ ,  $I_{CC} = 2\text{ mA}$ ,  $V_{OUT} = 200\text{ mV}$ ,  $I_{OUT} = 20\text{ mA}$  (output on), and  $R_{\theta JA} = 165^\circ\text{C/W}$ , then:

$$P_D = (V_{CC} \times I_{CC}) + (V_{OUT} \times I_{OUT}) = \\ (12\text{ V} \times 2\text{ mA}) + (200\text{ mV} \times 20\text{ mA}) = \\ 24\text{ mW} + 4\text{ mW} = 28\text{ mW}$$

$$\Delta T = P_D \times R_{\theta JA} = 28\text{ mW} \times 165^\circ\text{C/W} = 4.6^\circ\text{C}$$

$$T_J = T_A + \Delta T = 25^\circ\text{C} + 4.6^\circ\text{C} = 29.6^\circ\text{C}$$

A worst-case estimate,  $P_D(\text{max})$ , represents the maximum allowable power level ( $V_{CC}(\text{max})$ ,  $I_{CC}(\text{max})$ ), without exceeding  $T_J(\text{max})$ , at a selected  $R_{\theta JA}$ .

For example, given the conditions  $R_{\theta JA} = 228^\circ\text{C/W}$ ,  $T_J(\text{max}) = 175^\circ\text{C}$ ,  $V_{CC}(\text{max}) = 24\text{ V}$ ,  $I_{CC}(\text{max}) = 3.25\text{ mA}$ ,  $V_{OUT} = 500\text{ mV}$ , and  $I_{OUT} = 30\text{ mA}$  (output on), the maximum allowable operating ambient temperature can be determined.

The power dissipation required for the output is shown below:

$$P_D(V_{OUT}) = V_{OUT} \times I_{OUT} = 500\text{ mV} \times 30\text{ mA} = 15\text{ mW}$$

The power dissipation required for the IC supply is shown below:

$$P_D(V_{CC}) = V_{CC} \times I_{CC} = 24\text{ V} \times 3.25\text{ mA} = 78\text{ mW}$$

Next, by inverting using equation 2:

$$\Delta T = P_D \times R_{\theta JA} = [P_D(V_{OUT}) + P_D(V_{CC})] \times 228^\circ\text{C/W} = \\ (15\text{ mW} + 78\text{ mW}) \times 228^\circ\text{C/W} = \\ 93\text{ mW} \times 228^\circ\text{C/W} = 21.2^\circ\text{C}$$

Finally, by inverting equation 3 with respect to voltage:

$$T_A(\text{est}) = T_J(\text{max}) - \Delta T = 175^\circ\text{C} - 21.2^\circ\text{C} = 153.8^\circ\text{C}$$

In the above case, there is sufficient power dissipation capability to operate up to  $T_A(\text{est})$ . The example indicates that  $T_A(\text{max})$  can be as high as 153.8°C without exceeding  $T_J(\text{max})$ . However, the  $T_A(\text{max})$  rating of the devices is 125°C; the APS13290 and APS13291 performance is not guaranteed above  $T_A = 125^\circ\text{C}$ .



**Package UA, 3-Pin SIP**

**For Reference Only – Not for Tooling Use**

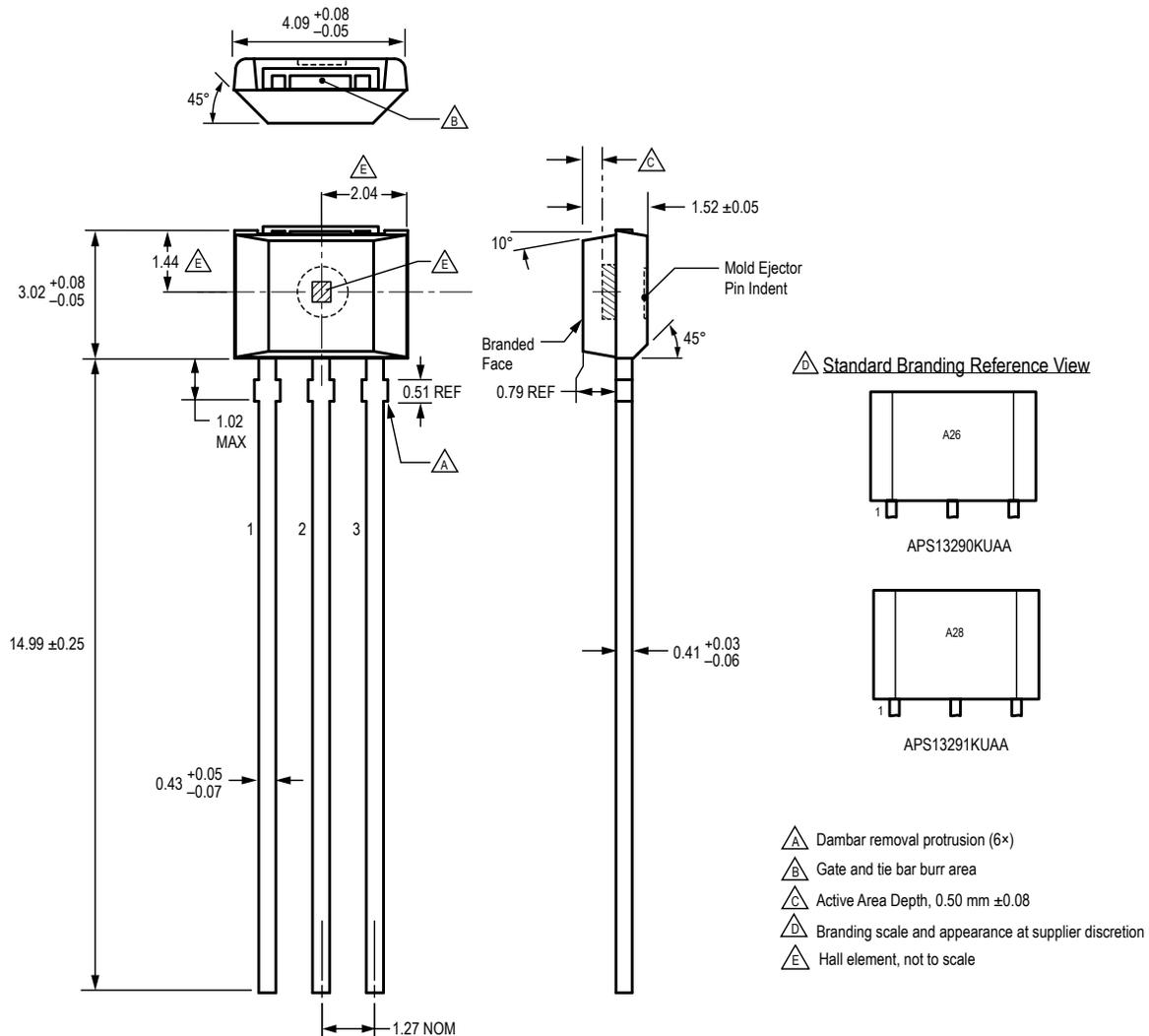
(Reference DWG-0000406, Rev. 1)

NOT TO SCALE

Dimensions in millimeters

Dimensions exclusive of mold flash, gate burrs, and dambar protrusions

Exact case and lead configuration at supplier discretion within limits shown



**Revision History**

Number	Date	Description
–	March 6, 2018	Initial release
1	February 11, 2019	Minor editorial updates
2	February 20, 2020	Minor editorial updates

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