TPS77601-EP, TPS77615-EP, TPS77618-EP, TPS77625-EP TPS77628-EP, TPS77633-EP WITH PG OUTPUT FAST-TRANSIENT-RESPONSE 500-mA LOW-DROPOUT VOLTAGE REGULATORS SGLS176A - AUGUST 2003 - REVISED MARCH 2007

- Controlled Baseline
 - One Assembly/Test Site, One Fabrication Site
- Extended Temperature Performance of -40°C to 125°C
- Enhanced Diminishing Manufacturing Sources (DMS) Support
- Enhanced Product Change Notification
- Qualification Pedigree[†]
- Open Drain Power Good
- 500-mA Low-Dropout Voltage Regulator
- Available in 1.5-V, 1.8-V, 2.5-V, 2.8-V, 3.3-V Fixed Output and Adjustable Versions
- Dropout Voltage to 169 mV (Typ) at 500 mA (TPS77633)

[†] Component qualification in accordance with JEDEC and industry standards to ensure reliable operation over an extended temperature range. This includes, but is not limited to, Highly Accelerated Stress Test (HAST) or biased 85/85, temperature cycle, autoclave or unbiased HAST, electromigration, bond intermetallic life, and mold compound life. Such qualification testing should not be viewed as justifying use of this component beyond specified performance and environmental limits.

- Ultralow 85 µA Typical Quiescent Current
- Fast Transient Response
- 2% Tolerance Over Specified Conditions for Fixed-Output Versions
- 20-Pin TSSOP PowerPADTM (PWP) Package
- Thermal Shutdown Protection

PWP PACKAGE (TOP VIEW)									
GND/HSINK	1	20	GND/HSINK						
GND/HSINK	2	19	GND/HSINK						
GND [3	18	NC						
NC [4	17	NC						
EN [5	16	PG						
IN [6	15	FB/NC						
IN [7	14	OUT						
NC [8	13	OUT						
GND/HSINK	9	12	GND/HSINK						
GND/HSINK	10	11	GND/HSINK						

NC - No internal connection

description

The TPS776xx devices are designed to have a fast transient response and be stable with a $10-\mu$ F low ESR capacitors. This combination provides high performance at a reasonable cost.

Because the PMOS device behaves as a low-value resistor, the dropout voltage is very low (typically 169 mV at an output current of 500 mA for the TPS77633) and is directly proportional to the output current. Additionally, since the PMOS pass element is a voltage-driven device, the quiescent current is very low and independent of output loading (typically 85 μ A over the full range of output current, 0 mA to 500 mA). These two key specifications yield a significant improvement in operating life for battery-powered systems. This LDO family also features a sleep mode; applying a TTL high signal to \overline{EN} (enable) shuts down the regulator, reducing the quiescent current to 1 μ A at T_J = 25°C.

Power good (PG) of the TPS776xx is an active high output, which can be used to implement a power-on reset or a low-battery indicator.

The TPS776xx are offered in 1.5-V, 1.8-V, 2.5-V, 2.8 V, and 3.3-V fixed-voltage versions and in an adjustable version (programmable over the range of 1.2 V to 5.5 V for TPS77601 option). Output voltage tolerance is specified as a maximum of 2% over line, load, and temperature ranges. The TPS776xx family is available in 20 pin TSSOP package.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

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ORDERING INFORMATION[†]

TA	OUTPUT VOLTAGE (V TYP)	PACK	AGE	ORDERABLE PART NUMBER	TOP-SIDE MARKING
	3.3	TSSOP – PW	Tape and reel	TPS77633QPWPREP	77633QE
	2.8	TSSOP – PW	Tape and reel	TPS77628QPWPREP§	77628QE
	2.5	TSSOP – PW	Tape and reel	TPS77625QPWPREP	77625QE
-40°C to 125°C	1.8	TSSOP – PW	Tape and reel	TPS77618QPWPREP	77618QE
	1.5	TSSOP – PW	Tape and reel	TPS77615QPWPREP	77615QE
	Adjustable [‡] 1.2 V to 5.5 V	TSSOP – PW	Tape and reel	TPS77601QPWPREP	77601QE

[†] Package drawings, standard packing quantities, thermal data, symbolization, and PCB design guidelines are available at www.ti.com/sc/package.

[‡]The TPS77601 is programmable using an external resistor divider (see application information).

§ TPS77628 is Product Preview.



[†] See application information section for capacitor selection details.



TPS77601-EP, TPS77615-EP, TPS77618-EP, TPS77625-EP TPS77628-EP, TPS77633-EP WITH PG OUTPUT FAST-TRANSIENT-RESPONSE 500-mA LOW-DROPOUT VOLTAGE REGULATORS SGLS176A - AUGUST 2003 - REVISED MARCH 2007

Figure 1. Typical Application Configuration for Fixed Output Options

functional block diagram—adjustable version



functional block diagram—fixed-voltage version





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Terminal Functions

TSSOP Package

TER	TERMINAL								
NAME	NO.	I/O	DESCRIPTION						
EN	5	I	Enable input						
FB/NC	15	I	Feedback input voltage for adjustable device (no connect for fixed options)						
GND	3		Regulator ground						
GND/HSINK	1, 2, 9, 10, 11, 12, 19, 20		Ground/heatsink						
IN	6, 7	I	Input voltage						
NC	4, 8, 17, 18		No connect						
OUT	13, 14	0	Regulated output voltage						
PG	16	0	PG output						



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absolute maximum ratings over operating free-air temperature range (unless otherwise noted)[†]

Input voltage range [‡] , V _I	
Maximum PG voltage	
Peak output current	Internally limited
Output voltage, V _O (OUT, FB)	
Continuous total power dissipation	See dissipation rating tables
Operating virtual junction temperature range, T ₁	–40°C to 125°C
Storage temperature range, T _{stg}	–65°C to 150°C
ESD rating, HBM	

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

[‡]All voltage values are with respect to network terminal ground.

DISSIPATION RATING TABLE – FREE-AIR TEMPERATURES									
PACKAGE	AIR FLOW (CFM)	T _A < 25°C POWER RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 70°C POWER RATING	T _A = 85°C POWER RATING				
PWP§	0	2.9 W	23.5 mW/°C	1.9 W	1.5 W				
PWP3	300	4.3 W	34.6 mW/°C	2.8 W	2.2 W				
PWP¶	0	3 W	23.8 mW/°C	1.9 W	1.5 W				
PWP1	300	7.2 W	57.9 mW/°C	4.6 W	3.8 W				

§ This parameter is measured with the recommended copper heat sink pattern on a 1-layer PCB, 5-in × 5-in PCB, 1 oz. copper, 2-in \times 2-in coverage (4 in²).

This parameter is measured with the recommended copper heat sink pattern on a 8-layer PCB, 1.5-in × 2-in PCB, 1 oz. copper with layers 1, 2, 4, 5, 7, and 8 at 5% coverage (0.9 in²) and layers 3 and 6 at 100% coverage (6 in²). For more information, refer to TI technical brief SLMA002.

recommended operating conditions

	MIN	MAX	UNIT
Input voltage, VI#	2.7	10	V
Output voltage range, VO	1.2	5.5	V
Output current, IO (see Note 1)	0	500	mA
Operating virtual junction temperature, T _J (see Note 1)	-40	125	°C

To calculate the minimum input voltage for your maximum output current, use the following equation: V_{I(min)} = V_{O(max)} + V_{DO(max load)}. NOTE 1: Continuous current and operating junction temperature are limited by internal protection circuitry, but it is not recommended that the device

operate under conditions beyond those specified in this table for extended periods of time.



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electrical characteristics over recommended operating free-air temperature range, V_I = V_{O(tvp)} + 1 V, I_O = 1 mA, EN = 0 V, C_o = 10 μ F (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
		$1.2 \text{ V} \le \text{V}_{O} \le 5.5 \text{ V}, \qquad \text{T}_{J} = 25^{\circ}\text{C}$		VO			
	TPS77601	$1.2 \text{ V} \le \text{V}_{O} \le 5.5 \text{ V}, \qquad \text{T}_{J} = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C}$	0.98VO		1.02VO		
		$T_J = 25^{\circ}C$, $2.7 V < V_{IN} < 10 V$		1.5		V	
	TPS77615	$T_J = -40^{\circ}C$ to 125°C, 2.7 V < V _{IN} < 10 V	1.470		1.530		
		$T_J = 25^{\circ}C$, $2.8 V < V_{IN} < 10 V$		1.8			
Output voltage (10 μA to 500 mA load)	TPS77618	$T_J = -40^{\circ}C$ to 125°C, 2.8 V < V _{IN} < 10 V	1.764		1.836		
(see Note 2)	TROTTOOL	$T_J = 25^{\circ}C$, $3.5 V < V_{IN} < 10 V$		2.5			
	TPS77625	$T_J = -40^{\circ}C$ to $125^{\circ}C$, $3.5 V < V_{IN} < 10 V$	2.450		2.550		
	-	$T_J = 25^{\circ}C$, $3.8 V < V_{IN} < 10 V$		2.8		V	
	TPS77628	$T_J = -40^{\circ}C \text{ to } 125^{\circ}C, 3.8 \text{ V} < \text{V}_{IN} < 10 \text{ V}$	2.744		2.856		
	TD077000	$T_J = 25^{\circ}C$, $4.3 V < V_{IN} < 10 V$		3.3			
	TPS77633	$T_J = -40^{\circ}C$ to $125^{\circ}C$, $4.3 \text{ V} < \text{V}_{IN} < 10 \text{ V}$	3.234		3.366		
Quiescent current (GND current)		$10 \ \mu A < I_O < 500 \ mA, T_J = 25^{\circ}C$		85			
$\overline{EN} = 0V$, (see Note 2)		$I_{O} = 500 \text{ mA},$ $T_{J} = -40^{\circ}\text{C} \text{ to } 125^{\circ}\text{C}$			125	μA	
Output voltage line regulation ($\Delta V_O/V_O$) (see Notes 2 and 3)		$V_O + 1 \ V < V_I \le 10 \ V, T_J = 25^\circ C$		0.01		%/V	
Load regulation				3		mV	
Output noise voltage (TPS77x18)		BW = 200 Hz to 100 kHz, I _C = 500 mA $C_0 = 10 \ \mu$ F, $T_J = 25^{\circ}$ C		53		μVrms	
Output current limit		$V_{O} = 0 V$		1.7	2.4	А	
Thermal shutdown junction temperature				150		°C	
		$\overline{EN} = V_{I}, $ $T_{J} = 25^{\circ}C, 2.7 V < V_{I} < 10 V$		1		μΑ	
Standby current		$ \overline{\text{EN}} = \text{V}_{\text{I}}, \qquad \text{T}_{\text{J}} = -40^{\circ}\text{C to } 125^{\circ}\text{C} \\ 2.7 \text{ V} < \text{V}_{\text{I}} < 10 \text{ V} $			10	μA	
FB input current	TPS77601	FB = 1.5 V		2		nA	
High level enable input voltage			1.7			V	
Low level enable input voltage					0.9	V	
Power supply ripple rejection (see Note 2))	f = 1 KHz, C _O = 10 μF, T _J = 25°C		60		dB	

NOTES: 2. Test condition for minimum IN operating voltage is 2.7 V or V_{O(typ)} + 1 V, whichever is greater. Test condition for maximum IN voltage is 10V.

3. If V_O \leq 1.8 V then V_{Imin} = 2.7 V, V_{Imax} = 10 V:

Line Reg. (mV) =
$$(\%/V) \times \frac{V_O(V_{Imax} - 2.7 V)}{100} \times 1000$$

If $V_O \ge 2.5$ V then $V_{Imin} = V_O + 1$ V, $V_{Imax} = 10$ V:

Line Reg. (mV) =
$$(\%/V) \times \frac{V_O(V_{Imax} - (V_O + 1V))}{100} \times 1000$$



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electrical characteristics over recommended operating free-air temperature range, $V_I = V_{O(typ)} + 1 V$, $I_O = 1 mA$, $\overline{EN} = 0 V$, $C_O = 10 \mu F$ (unless otherwise noted) (continued)

PARAMETER			TEST	MIN	TYP	MAX	UNIT	
	Minimum input voltage for valid	PG	I _{O(PG)} = 300 μA			1.1		V
	Trip threshold voltage		VO decreasing		92		98	%VO
PG	Hysteresis voltage		Measured at VO			0.5		%VO
	Output low voltage		VI = 2.7 V,	$I_{O(PG)} = 1 \text{ mA}$		0.15	0.4	V
	Leakage current		V(PG) = 5 V			1	μΑ	
	· · · ·		<u>EN</u> = 0 V		-1	0	1	
Input curre	ent (EN)		$\overline{EN} = V_{I}$		-1		1	μA
			I _O = 500 mA,	T _J = 25°C		285		
Dropout voltage (see Note 4)		TPS77628	I _O = 500 mA,	T _J = −40°C to 125°C			410	
		TPS77633	IO = 500 mA,	TJ = 25°C	169			mV
			l _O = 500 mA,	T _J = −40°C to 125°C			287	

NOTE 4: IN voltage equals VO(typ) - 100 mV; TPS77615, TPS77618, and TPS77625 dropout voltage limited by input voltage range limitations (i.e., TPS77633 input voltage needs to drop to 3.2 V for purpose of this test).

PICAL CHARACTERISTICS

Table of Graphs

			FIGURE
N/	vs Output current		2, 3, 4
VO	Output voltage	vs Free-air temperature	5, 6, 7
	Ground current	vs Free-air temperature	8
	Power supply ripple rejection	vs Frequency	9
	Output spectral noise density	vs Frequency	10
Z ₀	Output impedance	vs Frequency	11
V	Deserved and to an	vs Input voltage	12
VDO	Dropout voltage	vs Free-air temperature	13
	Input voltage (min)	vs Output voltage	14
	Line transient response		15, 17
	Load transient response		16, 18
VO	Output voltage	vs Time	19
	Equivalent series resistance (ESR)	vs Output current	21 – 24



TY-

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Figure 20. Test Circuit for Typical Regions of Stability (Figures 21 through 24) (Fixed Output Options)



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TYPICAL CHARACTERISTICS



† Equivalent series resistance (ESR) refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to Co.



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APPLICATION INFORMATION

The TPS776xx family includes five fixed-output voltage regulators (1.5 V, 1.8 V, 2.5 V, 2.8 V, and 3.3 V), and an adjustable regulator, the TPS77601 (adjustable from 1.2 V to 5.5 V).

device operation

The TPS776xx feature very low quiescent current, which remains virtually constant even with varying loads. Conventional LDO regulators use a pnp pass element, the base current of which is directly proportional to the load current through the regulator ($I_B = I_C/\beta$). The TPS776xx use a PMOS transistor to pass current; because the gate of the PMOS is voltage driven, operating current is low and invariable over the full load range.

Another pitfall associated with the pnp-pass element is its tendency to saturate when the device goes into dropout. The resulting drop in β forces an increase in I_B to maintain the load. During power up, this translates to large start-up currents. Systems with limited supply current may fail to start up. In battery-powered systems, it means rapid battery discharge when the voltage decays below the minimum required for regulation. The TPS776xx quiescent currents remain low even when the regulator drops out, eliminating both problems.

The TPS776xx family also features a shutdown mode that places the output in the high-impedance state (essentially equal to the feedback-divider resistance) and reduces quiescent current to 2 μ A. If the shutdown feature is not used, EN should be tied to ground.

minimum load requirements

The TPS776xx family is stable even at zero load; no minimum load is required for operation.

FB—pin connection (adjustable version only)

The FB pin is an input pin to sense the output voltage and close the loop for the adjustable option . The output voltage is sensed through a resistor divider network to close the loop as it is shown in Figure 26. Normally, this connection should be as short as possible; however, the connection can be made near a critical circuit to improve performance at that point. Internally, FB connects to a high-impedance wide-bandwidth amplifier and noise pickup feeds through to the regulator output. Routing the FB connection to minimize/avoid noise pickup is essential.

external capacitor requirements

An input capacitor is not usually required; however, a ceramic bypass capacitor (0.047 μ F or larger) improves load transient response and noise rejection if the TPS776xx is located more than a few inches from the power supply. A higher-capacitance electrolytic capacitor may be necessary if large (hundreds of milliamps) load transients with fast rise times are anticipated.

Like all low dropout regulators, the TPS776xx require an output capacitor connected between OUT and GND to stabilize the internal control loop. The minimum recommended capacitance value is 10 μ F and the ESR (equivalent series resistance) must be between 50 m Ω and 1.5 Ω . Capacitor values 10 μ F or larger are acceptable, provided the ESR is less than 1.5 Ω . Solid tantalum electrolytic, aluminum electrolytic, and multilayer ceramic capacitors are all suitable, provided they meet the requirements described previously.



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APPLICATION INFORMATION

external capacitor requirements (continued)



Figure 25. Typical Application Circuit (Fixed Versions)

programming the TPS77601 adjustable LDO regulator

The output voltage of the TPS77601 adjustable regulator is programmed using an external resistor divider as shown in Figure 26. The output voltage is calculated using:

$$V_{O} = V_{ref} \times \left(1 + \frac{R1}{R2}\right)$$
(1)

Where:

V_{ref} = 1.1834 V typ (the internal reference voltage)

Resistors R1 and R2 should be chosen for approximately 10-µA divider current. Lower value resistors can be used but offer no inherent advantage and waste more power. Higher values should be avoided as leakage currents at FB increase the output voltage error. The recommended design procedure is to choose R2 = 110 k Ω to set the divider current at approximately 10 μ A and then calculate R1 using:

$$R1 = \left(\frac{V_{O}}{V_{ref}} - 1\right) \times R2$$
(2)



OUT	OUTPUT VOLTAGE							
PROGRAMMING GUIDE								
TOUT								

OUTPUT VOLTAGE	R1	R2	UNIT
2.5 V	121	110	kΩ
3.3 V	196	110	kΩ
3.6 V	226	110	kΩ
4.75 V	332	110	kΩ

Figure 26. TPS77601 Adjustable LDO Regulator Programming



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APPLICATION INFORMATION

power-good indicator

The TPS776xx features a power-good (PG) output that can be used to monitor the status of the regulator. The internal comparator monitors the output voltage: when the output drops to between 92% and 98% of its nominal regulated value, the PG output transistor turns on, taking the signal low. The open-drain output requires a pullup resistor. If not used, it can be left floating. PG can be used to drive power-on reset circuitry or used as a low-battery indicator.

regulator protection

The TPS776xx PMOS-pass transistors have a built-in back diode that conducts reverse currents when the input voltage drops below the output voltage (e.g., during power down). Current is conducted from the output to the input and is not internally limited. When extended reverse voltage is anticipated, external limiting may be appropriate.

The TPS776xx also feature internal current limiting and thermal protection. During normal operation, the TPS776xx limit output current to approximately 1.7 A. When current limiting engages, the output voltage scales back linearly until the overcurrent condition ends. While current limiting is designed to prevent gross device failure, care should be taken not to exceed the power dissipation ratings of the package. If the temperature of the device exceeds 150°C(typ), thermal-protection circuitry shuts it down. Once the device has cooled below 130°C(typ), regulator operation resumes.

power dissipation and junction temperature

Specified regulator operation is assured to a junction temperature of 125°C; the maximum junction temperature should be restricted to 125°C under normal operating conditions. This restriction limits the power dissipation the regulator can handle in any given application. To ensure the junction temperature is within acceptable limits, calculate the maximum allowable dissipation, $P_{D(max)}$, and the actual dissipation, P_{D} , which must be less than or equal to $P_{D(max)}$.

The maximum-power-dissipation limit is determined using the following equation:

$$P_{D(max)} = \frac{T_J max - T_A}{R_{\theta JA}}$$

Where:

T_Jmax is the maximum allowable junction temperature.

 $R_{\theta JA}$ is the thermal resistance junction-to-ambient for the package, i.e., 32.6°C/W for the 20-terminal PWP with no airflow.

T_A is the ambient temperature.

The regulator dissipation is calculated using:

$$\mathsf{P}_{\mathsf{D}} = \left(\mathsf{V}_{\mathsf{I}} - \mathsf{V}_{\mathsf{O}}\right) \times \mathsf{I}_{\mathsf{O}}$$

Power dissipation resulting from quiescent current is negligible. Excessive power dissipation will trigger the thermal protection circuit.





10-Dec-2020

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
	(1)		Drainig		u .,	(2)	(6)	(3)		(4/5)	
TPS77601QPWPREP	ACTIVE	HTSSOP	PWP	20	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	77601QE	Samples
TPS77615QPWPREP	ACTIVE	HTSSOP	PWP	20	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	77615QE	Samples
TPS77618QPWPREP	ACTIVE	HTSSOP	PWP	20	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	77618QE	Samples
TPS77625QPWPREP	ACTIVE	HTSSOP	PWP	20	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	77625QE	Samples
TPS77633QPWPREP	ACTIVE	HTSSOP	PWP	20	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	77633QE	Samples
V62/03631-07XE	ACTIVE	HTSSOP	PWP	20	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	77601QE	Samples
V62/03631-08XE	ACTIVE	HTSSOP	PWP	20	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	77615QE	Samples
V62/03631-09XE	ACTIVE	HTSSOP	PWP	20	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	77618QE	Samples
V62/03631-10XE	ACTIVE	HTSSOP	PWP	20	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	77625QE	Samples
V62/03631-12XE	ACTIVE	HTSSOP	PWP	20	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	77633QE	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.



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PACKAGE OPTION ADDENDUM

10-Dec-2020

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS77601QPWPREP	HTSSOP	PWP	20	2000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1
TPS77615QPWPREP	HTSSOP	PWP	20	2000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1
TPS77618QPWPREP	HTSSOP	PWP	20	2000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1
TPS77625QPWPREP	HTSSOP	PWP	20	2000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1
TPS77633QPWPREP	HTSSOP	PWP	20	2000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1

TEXAS INSTRUMENTS

www.ti.com

PACKAGE MATERIALS INFORMATION

12-Feb-2019



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS77601QPWPREP	HTSSOP	PWP	20	2000	350.0	350.0	43.0
TPS77615QPWPREP	HTSSOP	PWP	20	2000	350.0	350.0	43.0
TPS77618QPWPREP	HTSSOP	PWP	20	2000	350.0	350.0	43.0
TPS77625QPWPREP	HTSSOP	PWP	20	2000	350.0	350.0	43.0
TPS77633QPWPREP	HTSSOP	PWP	20	2000	350.0	350.0	43.0

PWP (R-PDSO-G20)

PowerPAD[™] PLASTIC SMALL OUTLINE



All linear dimensions are in millimeters. NOTES: Α.

- Β. This drawing is subject to change without notice.
- Body dimensions do not include mold flash or protrusions. Mold flash and protrusion shall not exceed 0.15 per side. C.
- This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad D.
- Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com http://www.ti.com. E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions. E. Falls within JEDEC MO-153

PowerPAD is a trademark of Texas Instruments.



PWP (R-PDSO-G20) PowerPAD[™] SMALL PLASTIC OUTLINE

THERMAL INFORMATION

This PowerPAD[™] package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



PowerPAD is a trademark of Texas Instruments





NOTES:

- This drawing is subject to change without notice. Β.
- Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad. C.
- This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad D. Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <http://www.ti.com>. Publication IPC-7351 is recommended for alternate designs.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.
- F. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



PWP (R-PDSO-G20) PowerPAD[™] SMALL PLASTIC OUTLINE

THERMAL INFORMATION

This PowerPAD[™] package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

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The exposed thermal pad dimensions for this package are shown in the following illustration.



PowerPAD is a trademark of Texas Instruments



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