











bq20z655-R1

ZHCS382A -AUGUST 2011-REVISED AUGUST 2015

# 采用 Impedance Track™ 且符合 SBS 1.1 标准的 bq20z655-R1 电量监测 计和保护

### 1 特性

- 已为下一代产品申请专利 Impedance Track™技术 可准确测量锂离子和锂聚合物电池中的可用电量
  - 在电池生命周期内误差优于 1%
- 支持智能电池规范 SBS V1.1
- 可针对 2 节至 4 节锂离子与锂聚合物电池进行灵活 配置
- 具有超低功耗模式的强大 8 位精简指令集 (RISC) CPU
- 全面的可编程保护 功能
  - 电压、电流和温度
- 符合 JEITA 标准
- 可处理更复杂充电模式的附加灵活性
- 使用寿命的数据记录
- 驱动电池组供电条件下的 3、4 或 5 断码液晶显示 屏与 LED
- 支持 SHA-1 认证
- 同一封装中的完整电池保护与电量计解决方案

### 2 应用

- 医疗和测试设备
- 便携式仪表和工业设备
- 可充电电池组

### 3 说明

符合 SBS 标准的 bq20z655-R1 电量监测计和保护 IC 采用获得专利的 Impedance Track™ 技术,是专为电池组或系统内安装而设计的单个 IC 解决方案。

bq20z655-R1 借助其集成的高性能模拟外设,可测量 锂离子或锂聚合物电池的可用电量并保持精确的记录。bq20z655-R1 可监控容量变化、电池阻抗、开路电压 和电池组的其他关键参数,并通过串行通信总线将信息 报告给系统主机控制器。bq20z655-R1 借助集成模拟 前端 (AFE) 短路和过载保护,在最大限度减小智能电池电路中的外部组件数量、成本和尺寸的同时,可最大程度地提升功能性和安全性。

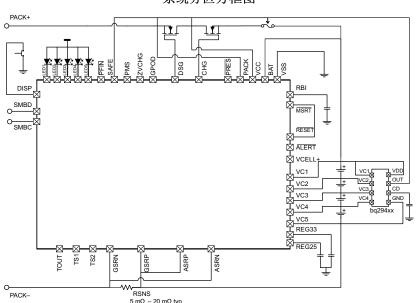
所实施的 Impedance Track™ 电量监测技术可持续地分析电池阻抗,从而可实现极高精度的电量监测。这样一来,通过使用每个周期的每个阶段内的放电速率、温度以及电池老化情况,可精确计算出剩余容量。

### 器件信息(1)

器件型号	封装	封装尺寸 (标称值)
bq20z655-R1	DBT (44)	4.40mm × 11.29mm

(1) 如需了解所有可用封装,请参阅数据表末尾的可订购产品附录。

### 系统分区方框图



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注: 之前版本的页码可能与当前版本有所不同。

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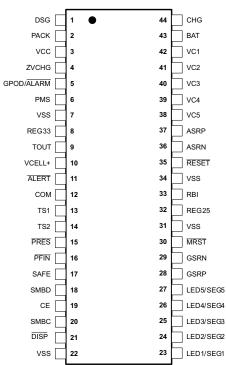
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### 5 Pin Configuration and Functions





#### **Pin Functions**

PIN		I/O <sup>(1)</sup>	DECODINE
NO.	NAME	1/0	DESCRIPTION
1	DSG	0	High side N-chan discharge FET gate drive
2	PACK	IA, P	Battery pack input voltage sense input. It also serves as device wakeup when device is in shutdown mode.
3	VCC	Р	Positive device supply input. Connect to the center connection of the CHG FET and DSG FET to ensure device supply either from battery stack or battery pack input
4	ZVCHG	0	P-chan pre-charge FET gate drive
5	GPOD	OD	High voltage general purpose open drain output. Can be configured to be used in pre-charge condition
6	PMS	1	Pre-charge mode setting input. Connect to PACK to enable 0-V precharge using charge FET connected at CHG pin. Connect to VSS to disable 0-V precharge using charge FET connected at CHG pin.
7	VSS	Р	Negative supply voltage input. Connect all VSS pins together for operation of device
8	REG33	Р	3.3-V regulator output. Connect at least a 2.2-μF capacitor to REG33 and VSS
9	TOUT	Р	Thermistor bias supply output
10	VCELL+	_	Internal cell voltage multiplexer and amplifier output. Connect a 0.1-μF capacitor to VCELL+ and VSS
11	ALERT	OD	Alert output. In case of short circuit condition, overload condition and watchdog time out this pin will be triggered.
12	COM/TP	_	Output / open drain: LCD common connection
13	TS1	IA	1 <sup>st</sup> Thermistor voltage input connection to monitor temperature
14	TS2	IA	2 <sup>nd</sup> Thermistor voltage input connection to monitor temperature

(1) I = Input, IA = Analog input, I/O = Input/output, I/OD = Input/Open-drain output, O = Output, OA = Analog output, P = Power



### Pin Functions (continued)

	PIN	40	
NO.	NAME	I/O <sup>(1)</sup>	DESCRIPTION
15	PRES	ı	Active low input to sense system insertion. Typically requires additional ESD protection.
16	PFIN	I	Active low input to detect secondary protector status, and to allow the bq20z655-R1 to report the status of the 2 <sup>nd</sup> level protection input.
17	SAFE	OD	Active high output to enforce additional level of safety protection; for example, fuse blow.
18	SMBD	I/OD	SMBus data open-drain bidirectional pin used to transfer address and data to and from the bq20z655-R1
19	CE	_	A logical high on this pin only affects the normal operation on the charge FET when the battery is in charge/relax mode. For a logic low, the normal bq20z655-R1 firmware controls the charge FET.
20	SMBC	I/OD	SMBus clock open-drain bidirectional pin used to clock the data transfer to and from the bq20z655-R1
21	DISP	I	Input: In LED mode, this is the display enable input.
22	VSS	Р	Negative supply voltage input. Connect all VSS pins together for operation of device
23	LED1/SEG1	1	Output / open drain: LED 1 current sink. LCD segment 1
24	LED2/SEG2	I	Output / open drain: LED 2 current sink. LCD segment 2
25	LED3/SEG3	I	Output / open drain: LED 3 current sink. LCD segment 3
26	LED4/SEG4	1	Output / open drain: LED 4 current sink. LCD segment 4
27	LED5/SEG5	1	Output / open drain: LED 5 current sink. LCD segment 5
28	GSRP	IA	Coulomb counter differential input. Connect to one side of the sense resistor
29	GSRN	IA	Coulomb counter differential input. Connect to one side of the sense resistor
30	MRST	I	Master reset input that forces the device into reset when held low. Must be held high for normal operation. Connect to RESET for correct operation of device
31	VSS	Р	Negative supply voltage input. Connect all VSS pins together for operation of device
32	REG25	Р	2.5-V regulator output. Connect at least a 1-mF capacitor to REG25 and VSS
33	RBI	Р	RAM / Register backup input. Connect a capacitor to this pin and VSS to protect loss of RAM/Register data in case of short circuit condition.
34	VSS	Р	Negative supply voltage input. Connect all VSS pins together for operation of device
35	RESET	0	Reset output. Connect to MSRT.
36	ASRN	IA	Short circuit and overload detection differential input. Connect to sense resistor
37	ASRP	IA	Short circuit and overload detection differential input. Connect to sense resistor
38	VC5	IA, P	Cell voltage sense input and cell balancing input for the negative voltage of the bottom cell in cell stack.
39	VC4	IA, P	Cell voltage sense input and cell balancing input for the positive voltage of the bottom cell and the negative voltage of the second lowest cell in cell stack.
40	VC3	IA, P	Cell voltage sense input and cell balancing input for the positive voltage of the second lowest cell in cell stack and the negative voltage of the second highest cell in 4 cell applications.
41	VC2	IA, P	Cell voltage sense input and cell balancing input for the positive voltage of the second highest cell and the negative voltage of the highest cell in 4 cell applications. Connect to VC3 in 2 cell stack applications.
42	VC1	IA, P	Cell voltage sense input and cell balancing input for the positive voltage of the highest cell in cell stack in 4 cell applications. Connect to VC2 in 2- or 3-stack applications.
43	BAT	I, P	Battery stack voltage sense input.
44	CHG	0	High side N-channel charge FET gate drive



### 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

			MIN	MAX	UNIT
		BAT, VCC	-0.3	34	
$V_{SS}$		PACK, PMS	-0.3	34	
	Supply voltage	VC(n)-VC(n+1); n = 1, 2, 3, 4	-0.3	8.5	V
		VC1, VC2, VC3, VC4	-0.3	34	
		VC5	-0.3	1	
V <sub>IN</sub>	Input voltage	PFIN, SMBD, SMBC. LED1, LED2, LED3, LED4, LED5, DISP	-0.3	6	
		TS1, TS2, SAFE, VCELL+, PRES, ALERT	-0.3	V <sub>(REG25)</sub> + 0.3	V
		MRST, GSRN, GSRP, RBI	-0.3	V <sub>(REG25)</sub> + 0.3	V
		ASRN, ASRP	-1	1	
		DSG, CHG, GPOD	-0.3	34	
		ZVCHG	-0.3	V <sub>(BAT)</sub>	
$V_{\text{OUT}}$	Output voltage	TOUT, ALERT, REG33	-0.3	6	V
		RESET	-0.3	7	
		REG25	-0.3	2.75	
I <sub>SS</sub>	Maximum combined sink current for input pins	PRES, PFIN, SMBD, SMBC, LED1, LED2, LED3, LED4, LED5		50	mA
T <sub>A</sub>	Operating free-air temperatur	-40	85	°C	
T <sub>F</sub>	Functional temperature		-40	100	°C
T <sub>stg</sub>	Storage temperature		-65	150	°C

<sup>(1)</sup> Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 6.2 ESD Ratings

			VALUE	UNIT
V <sub>(ESD)</sub>	Electrostatic	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	±2000	\/
	discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	±500	V

<sup>(1)</sup> JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

### 6.3 Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted)

			MIN	NOM MAX	UNIT
V <sub>SS</sub>	Supply voltage	VCC, BAT	4.5	25	V
V <sub>(STARTUP)</sub>	Minimum start-up voltage	VCC, BAT, PACK	5.5		V
		VC(n)-VC(n+1); n = 1,2,3,4	0	5	V
		VC1, VC2, VC3, VC4	0	V <sub>SS</sub>	V
$V_{IN}$	Input voltage	VC5	0	0.5	V
		ASRN, ASRP	-0.5	0.5	V
		PACK, PMS	0	25	V
V <sub>(GPOD)</sub>	Output voltage	GPOD	0	25	V
I <sub>(GPOD)</sub>	Drain current <sup>(1)</sup>	GPOD		1	mA
C <sub>(REG25)</sub>	2.5-V LDO capacitor	REG25	1		μF
C <sub>(REG33)</sub>	3.3-V LDO capacitor	REG33	2.2		μF
C <sub>(VCELL+)</sub>	Cell voltage output capacitor	VCELL+	0.1		μF
R <sub>(PACK)</sub>	PACK input block resistor <sup>(2)</sup>	PACK	1		kΩ

<sup>(1)</sup> Use an external resistor to limit the current to GPOD to 1 mA in high voltage application.

<sup>(2)</sup> JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

<sup>(2)</sup> Use an external resistor to limit the inrush current PACK pin required.



### 6.4 Thermal Information

		bq20z655-R1	
	THERMAL METRIC <sup>(1)</sup>	DBT (TSSOP)	UNIT
		44 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	60.9	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	15.3	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	30.2	°C/W
ΨЈТ	Junction-to-top characterization parameter	0.3	°C/W
ΨЈВ	Junction-to-board characterization parameter	27.2	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	N/A	°C/W

<sup>(1)</sup> For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report, SPRA953.

### 6.5 Electrical Characteristics

Over operating free-air temperature range,  $T_A = -40^{\circ}C$  to 85°C,  $V_{(REG25)} = 2.41$  V to 2.59 V,  $V_{(BAT)} = 14$  V,  $C_{(REG25)} = 1$   $\mu$ F,  $C_{(REG33)} = 2.2$   $\mu$ F; typical values at  $T_A = 25^{\circ}C$  (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
SUPPLY CURF	RENT					
I <sub>(NORMAL)</sub>	Firmware running			550		μA
I <sub>(SLEEP)</sub>	Sleep mode	CHG FET on; DSG FET on		124		μA
		CHG FET off; DSG FET on		90		μΑ
		CHG FET off; DSG FET off		52		μA
I <sub>(SHUTDOWN)</sub>	Shutdown mode			0.1	1	μA
SHUTDOWN W	VAKE; T <sub>A</sub> = 25°C (unless otherw	ise noted)				
I <sub>(PACK)</sub>	Shutdown exit at V <sub>STARTUP</sub> threshold				1	μΑ
SRx WAKE FR	OM SLEEP; T <sub>A</sub> = 25°C (unless o	otherwise noted)	•		'	
V <sub>(WAKE)</sub>	Positive or negative wake threshold with 1-mV, 2.25-mV, 4.5-mV and 9-mV programmable options		1.25		10	mV
		V <sub>(WAKE)</sub> = 1 mV; I <sub>(WAKE)</sub> = 0, RSNS1 = 0, RSNS0 = 1;	-0.7		0.7	
V	Accuracy of V <sub>(WAKE)</sub>	$V_{(WAKE)} = 2.25 \text{ mV}; I_{(WAKE)} = 1, RSNS1 = 0, RSNS0 = 1; I_{(WAKE)} = 0, RSNS1 = 1, RSNS0 = 0;$	-0.8		0.8	ma\/
V <sub>(WAKE_ACR)</sub>		$ \begin{aligned} & V_{(WAKE)} = 4.5 \text{ mV; } I_{(WAKE)} = 1, \text{ RSNS1} = 1, \text{ RSNS0} = 1; \\ I_{(WAKE)} = 0, \text{ RSNS1} = 1, \text{ RSNS0} = 0; \end{aligned} $	-1		1	mV
		$V_{(WAKE)} = 9 \text{ mV}; I_{(WAKE)} = 1, RSNS1 = 1, RSNS0 = 1;$	-1.4		1.4	
V <sub>(WAKE_TCO)</sub>	Temperature drift of $V_{(WAKE)}$ accuracy			0.5		%/°C
$t_{(WAKE)}$	Time from application of current and wake of bq20z655-R1			1	10	ms
WATCHDOG T	IMER				*	
t <sub>WDTINT</sub>	Watchdog start-up detect time		250	500	1000	ms
t <sub>WDWT</sub>	Watchdog detect time		50	100	150	μs
2.5V LDO; I <sub>(REC</sub>	<sub>333OUT)</sub> = 0 mA; T <sub>A</sub> = 25°C (unles	s otherwise noted)				
V <sub>(REG25)</sub>	Regulator output voltage	4.5 < VCC or BAT < 25 V; $I_{(REG25OUT)} \le 16$ mA; $T_A = -40$ °C to 100°C	2.41	2.5	2.59	V
$\Delta V_{(REG25TEMP)}$	Regulator output change with temperature	$I_{(REG25OUT)} = 2 \text{ mA}; T_A = -40^{\circ}\text{C to } 100^{\circ}\text{C}$		±0.2%		
$\Delta V_{(REG25LINE)}$	Line regulation	5.4 < VCC or BAT < 25 V; I <sub>(REG25OUT)</sub> = 2 mA		3	10	mV
41/	Load regulation	0.2 mA ≤ I <sub>(REG25OUT)</sub> ≤ 2 mA		7	25	mV
$\Delta V_{(REG25LOAD)}$	Load regulation	0.2 mA ≤ I <sub>(REG25OUT)</sub> ≤ 16 mA		25	50	IIIV
I <sub>(REG25MAX)</sub>	Current limit	drawing current until REG25 = 2 V to 0 V	5	40	75	mA
3.3V LDO: I/PE/	<sub>325OUT)</sub> = 0 mA; T <sub>A</sub> = 25°C (unles	s otherwise noted)	1		l	



### **Electrical Characteristics (continued)**

Over operating free-air temperature range,  $T_A = -40^{\circ}C$  to  $85^{\circ}C$ ,  $V_{(REG25)} = 2.41$  V to 2.59 V,  $V_{(BAT)} = 14$  V,  $C_{(REG25)} = 1$   $\mu$ F,  $C_{(REG33)} = 2.2$   $\mu$ F; typical values at  $T_A = 25^{\circ}C$  (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>(REG33)</sub>	Regulator output voltage		3	3.3	3.6	V
$\Delta V_{(REG33TEMP)}$	Regulator output change with temperature	$I_{(REG33OUT)} = 2 \text{ mA}; T_A = -40^{\circ}\text{C to } 100^{\circ}\text{C}$		±0.2%		
ΔV <sub>(REG33LINE)</sub>	Line regulation	5.4 < VCC or BAT < 25 V; I <sub>(REG33OUT)</sub> = 2 mA		3	10	mV
A\/	Load regulation	0.2 mA ≤ I <sub>(REG33OUT)</sub> ≤ 2 mA		7	17	mV
$\Delta V_{(REG33LOAD)}$	Load regulation	0.2 mA ≤ I <sub>(REG33OUT)</sub> ≤ 25 mA		40	100	IIIV
ı	Current limit	drawing current until REG33 = 3 V	25	100	145	mA
(REG33MAX)	Current minit	short REG33 to VSS, REG33 = 0 V	12		65	IIIA
THERMISTOR	DRIVE					
$V_{(TOUT)}$	Output voltage	$I_{(TOUT)} = 0 \text{ mA}; T_A = 25^{\circ}\text{C}$		V <sub>(REG25)</sub>		V
R <sub>DS(on)</sub>	TOUT pass element resistance	$I_{(TOUT)}=1$ mA; $R_{DS(on)}=(V_{(REG25)}$ - $V_{(TOUT)}$ )/ 1 mA; $T_A=-40^{\circ}C$ to $100^{\circ}C$		50	100	Ω
LED OUTPUTS						
V <sub>OL</sub>	Output low voltage	LED1, LED2, LED3, LED4, LED5			0.4	V
VCELL+ HIGH	VOLTAGE TRANSLATION					
		$VC(n) - VC(n+1) = 0 V; T_A = -40^{\circ}C \text{ to } 100^{\circ}C$	0.95	0.975	1	
V <sub>(VCELL+OUT)</sub>		VC(n) - VC(n+1) = 4.5 V; T <sub>A</sub> = -40°C to 100°C	0.275	0.3	0.375	
V <sub>(VCELL+REF)</sub>		Internal AFE reference voltage ; T <sub>A</sub> = -40°C to 100°C	0.965	0.975	0.985	
V <sub>(VCELL+PACK)</sub>	Translation output	Voltage at PACK pin; T <sub>A</sub> = -40°C to 100°C	0.98 x V <sub>(PACK)</sub> /18	V <sub>(PACK)</sub> /1 8	1.02 x V <sub>(PACK)</sub> /1 8	V
V <sub>(VCELL+BAT)</sub>	-	Voltage at BAT pin; T <sub>A</sub> = -40°C to 100°C	0.98 × V <sub>(BAT)</sub> /18	V <sub>(BAT)</sub> /18	1.02 × V <sub>(BAT)</sub> /18	
CMMR	Common mode rejection ratio	VCELL+	40			dB
		K= {VCELL+ output (VC5=0 V; VC4=4.5 V) - VCELL+ output (VC5=0 V; VC4=0 V)}/4.5	0.147	0.15	0.153	
K	Cell scale factor	K= {VCELL+ output (VC2=13.5 V; VC1=18 V) - VCELL+ output (VC5=13.5 V; VC1=13.5 V)}/4.5	0.147	0.15	0.153	
I <sub>(VCELL+OUT)</sub>	Drive Current to VCELL+ capacitor	VC(n) - VC(n+1) = 0 V; VCELL+ = 0 V; T <sub>A</sub> = -40°C to 100°C	12	18		μΑ
V <sub>(VCELL+O)</sub>	CELL offset error	CELL output (VC2 = VC1 = 18 V) - CELL output (VC2 = VC1 = 0 V)	-18	-1	18	mV
I <sub>VCnL</sub>	VC(n) pin leakage current	VC1, VC2, VC3, VC4, VC5 = 3 V	-1	0.01	1	μΑ
CELL BALANC	CING					
R <sub>BAL</sub>	internal cell balancing FET resistance	$R_{DS(on)}$ for internal FET switch at $V_{DS} = 2 \text{ V}$ ; $T_A = 25^{\circ}\text{C}$	200	400	600	Ω
HARDWARE S	HORT CIRCUIT AND OVERLOA	D PROTECTION; T <sub>A</sub> = 25°C (unless otherwise noted)				
		V <sub>OL</sub> = 25 mV (minimum)	15	25	35	
$V_{(OL)}$	OL detection threshold voltage accuracy	V <sub>OL</sub> = 100 mV; RSNS = 0, 1	90	100	110	mV
		V <sub>OL</sub> = 205 mV (maximum)	185	205	225	
		V <sub>(SCC)</sub> = 50 mV (minimum)	30	50	70	-
V <sub>(SCC)</sub>	SCC detection threshold voltage accuracy	V <sub>(SCC)</sub> = 200 mV; RSNS = 0, 1	180	200	220	mV
		V <sub>(SCC)</sub> = 475 mV (maximum)	428	475	523	
		V <sub>(SCD)</sub> = -50 mV (minimum)	-30	-50	-70	
V <sub>(SCD)</sub>	SCD detection threshold voltage accuracy	V <sub>(SCD)</sub> = -200 mV; RSNS = 0, 1	-180	-200	-220	mV
	voltage accuracy	V <sub>(SCD)</sub> = -475 mV (maximum)	-428	-475	-523	
t <sub>da</sub>	Delay time accuracy			±15.25		μS
t <sub>pd</sub>	Protection circuit propagation delay			50		μS
EET DRIVE OI	RCUIT; T <sub>A</sub> = 25°C (unless other)	viso notod)	•			



### **Electrical Characteristics (continued)**

Over operating free-air temperature range,  $T_A = -40^{\circ}\text{C}$  to 85°C,  $V_{(REG25)} = 2.41 \text{ V}$  to 2.59 V,  $V_{(BAT)} = 14 \text{ V}$ ,  $C_{(REG25)} = 1 \mu\text{F}$ ,  $C_{(REG33)} = 2.2 \mu\text{F}$ ; typical values at  $T_A = 25^{\circ}\text{C}$  (unless otherwise noted)

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>(DSGON)</sub>	DSG pin output on voltage	$\begin{array}{l} V_{(DSGON)} = V_{(DSG)} \cdot V_{(PACK)}; \\ V_{(GS)} \ connected \ to \ 10 \ M\Omega; \ DSG \ and \ CHG \ on; \\ T_A = -40^{\circ}C \ to \ 100^{\circ}C \end{array}$		8	12	16	V
V <sub>(CHGON)</sub>	CHG pin output on voltage	$V_{(CHGON)} = V_{(CHG)}$ $V_{(GS)} = 10 \text{ M}\Omega; DS$	- V <sub>(BAT)</sub> ; SG and CHG on; T <sub>A</sub> = -40°C to 100°C	8	12	16	V
V <sub>(DSGOFF)</sub>	DSG pin output off voltage	$V_{(DSGOFF)} = V_{(DSG)}$	- V <sub>(PACK)</sub>			0.2	V
V <sub>(CHGOFF)</sub>	CHG pin output off voltage	$V_{(CHGOFF)} = V_{(CHG)}$	) - V <sub>(BAT)</sub>			0.2	V
4	Ding time	C 4700 pF	$V_{(CHG)}$ : $V_{(PACK)} \ge V_{(PACK)} + 4 V$		400	1000	
t <sub>r</sub>	Rise time	C <sub>L</sub> = 4700 pF	V <sub>(DSG)</sub> : V <sub>(BAT)</sub> ≥V <sub>(BAT)</sub> + 4 V		400	1000	μS
4	Fall time	C 4700 pF	$V_{(CHG)}$ : $V_{(PACK)} + V_{(CHGON)} \ge V_{(PACK)} + 1 V$		40	200	
t <sub>f</sub>	Fall time	C <sub>L</sub> = 4700 pF	$V_{(DSG)}$ : VC1 + $V_{(DSGON)} \ge$ VC1 + 1 V		40	200	μS
V <sub>(ZVCHG)</sub>	ZVCHG clamp voltage	BAT = 4.5 V		3.3	3.5	3.7	V
LOGIC; T <sub>A</sub> =	-40°C to 100°C (unless otherwise	e noted)					
D.	latera el mullon de cieta e e	ALERT		60	100	200	1.0
R <sub>(PULLUP)</sub>	Internal pullup resistance	RESET		1	3	6	kΩ
		ALERT				0.2	
$V_{OL}$	Logic low output voltage level	$\overline{RESET}$ ; $V_{(BAT)} = \overline{I}$	7 V; V <sub>(REG25)</sub> = 1.5 V; I <sub>(RESET)</sub> = 200 μA			0.4	V
		GPOD; I <sub>(GPOD)</sub> = 5	50 μA			0.6	
LOGIC SMB	C, SMBD, PFIN, PRES, SAFE, ALE	RT, DISP					
V <sub>IH</sub>	High-level input voltage			2			V
V <sub>IL</sub>	Low-level input voltage					0.8	V
V <sub>OH</sub>	Output voltage high (RC[0:7] bus)	I <sub>L</sub> = -0.5 mA		V <sub>REG25</sub> -0.			V
V <sub>OL</sub>	Low-level output voltage	PRES, PFIN, ALE	RT, DISP; I <sub>L</sub> = 7 mA;			0.4	V
Cı	Input capacitance				5		pF
I <sub>(SAFE)</sub>	SAFE source currents	SAFE active, SAF	FE = V <sub>(REG25)</sub> -0.6 V	-3			mA
I <sub>lkg(SAFE)</sub>	SAFE leakage current	SAFE inactive		-0.2		0.2	μA
I <sub>lkg</sub>	Input leakage current					1	μA
	s otherwise specified, the specific	ation limits are va	llid at all measurement speed modes.)				
	Input voltage range	TS1, TS2, using I	nternal V <sub>ref</sub>	-0.2		1	V
	Conversion time				31.5		ms
	Resolution (no missing codes)			16			bits
	Effective resolution			14	15		bits
	Integral nonlinearity					±0.03	%FSR <sup>(1)</sup>
	Offset error <sup>(2)</sup>				140	250	μV
	Offset error drift <sup>(2)</sup>	T <sub>A</sub> = 25°C to 85°C			2.5	18	μV/°C
	Full-scale error <sup>(3)</sup>	.,			±0.1%	±0.7%	
	Full-scale error drift				50		PPM/°C
	Effective input resistance (4)			8			ΜΩ
COULOMB	· · · · · · · · · · · · · · · · · · ·	1		1			
	Input voltage range			-0.20		0.20	V
	Conversion time	Single conversion		5.25	250	0.20	ms
	Effective resolution	Single conversion		15	200		bits
	211001170 10001011011	-0.1 V to 0.2 V		10	±0.007	±0.034	Dito
	Integral nonlinearity	-0.1 V to 0.2 V			±0.007	±0.004	%FSR
	Offset error <sup>(5)</sup>	$T_A = 25^{\circ}\text{C to } 85^{\circ}\text{C}$			10		μV
	Oliger elloi	1A - 20 0 10 00 C	,		10		μν

- (1) Full-scale reference
- (2) Post-calibration performance and no I/O changes during conversion with SRN as the ground reference.
- (3) Uncalibrated performance. This gain error can be eliminated with external calibration.
- (4) The A/D input is a switched-capacitor input. Because the input is switched, the effective input resistance is a measure of the average resistance.
- (5) Post-calibration performance



### **Electrical Characteristics (continued)**

Over operating free-air temperature range,  $T_A = -40$ °C to 85°C,  $V_{(REG25)} = 2.41$  V to 2.59 V,  $V_{(BAT)} = 14$  V,  $C_{(REG25)} = 1$   $\mu$ F,  $C_{(REG33)} = 2.2 \mu F$ ; typical values at  $T_A = 25$ °C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Offset error drift			0.4	0.7	μV/°C
	Full-scale error <sup>(6)</sup> (7)			±0.35%		
	Full-scale error drift			150		PPM/°C
	Effective input resistance <sup>(8)</sup>	T <sub>A</sub> = 25°C to 85°C	2.5			МΩ
INTERNAL	TEMPERATURE SENSOR					
V <sub>(TEMP)</sub>	Temperature sensor voltage	9)		-2		mV/°C
	REFERENCE					
	Output voltage		1.215	1.225	1.230	V
	Output voltage drift			65		PPM/°C
HIGH-FREG	QUENCY OSCILLATOR					
f <sub>(OSC)</sub>	Operating frequency			4.194		MHz
,	(10) (11)		-3%	0.25%	3%	
f <sub>(EIO)</sub>	Frequency error (10) (11)	T <sub>A</sub> = 20°C to 70°C	-2%	0.25%	2%	
t <sub>(SXO)</sub>	Start-up time <sup>(12)</sup>			2.5	5	ms
LOW-FREQ	UENCY OSCILLATOR					
f <sub>(LOSC)</sub>	Operating frequency			32.768		kHz
ı	Frequency error <sup>(11)</sup> (13)		-2.5%	0.25%	2.5%	
f <sub>(LEIO)</sub>		T <sub>A</sub> = 20°C to 70°C	-1.5%	0.25%	1.5%	
t <sub>(LSXO)</sub>	Start-up time <sup>(12)</sup>				500	μs

- Reference voltage for the coulomb counter is typically  $V_{ref}/3.969$  at  $V_{(REG25)} = 2.5 \text{ V}$ ,  $T_A = 25^{\circ}\text{C}$ .
- Uncalibrated performance. This gain error can be eliminated with external calibration.
- (8) The CC input is a switched capacitor input. Because the input is switched, the effective input resistance is a measure of the average resistance.
- -53.7 LSB/°C
- (10) The frequency error is measured from 4.194 MHz.
- (11) The frequency drift is included and measured from the trimmed frequency at  $V_{(REG25)} = 2.5 \text{ V}$ ,  $T_A = 25^{\circ}\text{C}$ . (12) The start-up time is defined as the time it takes for the oscillator output frequency to be  $\pm 3\%$ .
- (13) The frequency error is measured from 32.768 kHz.

#### 6.6 Power-on Reset

Over operating free-air temperature range (unless otherwise noted),  $T_A = -40^{\circ}\text{C}$  to 85°C,  $V_{(REG25)} = 2.41 \text{ V}$  to 2.59 V,  $V_{(BAT)} = 14 \text{ V}$ ,  $C_{(REG25)} = 1 \text{ }\mu\text{F}$ ,  $C_{(REG33)} = 2.2 \text{ }\mu\text{F}$ ; typical values at  $T_A = 25^{\circ}\text{C}$  (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
VIT-	Negative-going voltage input		1.7	1.8	1.9	V
VHYS	Power-on reset hysteresis		5	125	200	mV
t <sub>RST</sub>	RESET active low time	Active low time after power up or watchdog reset	100	250	560	μs



### 6.7 Data Flash Characteristics Over Recommended Operating Temperature and Supply Voltage

Typical values at  $T_A = 25$ °C and  $V_{(REG25)} = 2.5$  V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Data retention		10			Years
	Flash programming write-cycles		20k			Cycles
t <sub>(ROWPROG)</sub>	Row programming time	See <sup>(1)</sup>			2	ms
t <sub>(MASSERASE)</sub>	Mass-erase time				200	ms
t <sub>(PAGEERASE)</sub>	Page-erase time				20	ms
I <sub>(DDPROG)</sub>	Flash-write supply current			5	10	mA
I <sub>(DDERASE)</sub>	Flash-erase supply current			5	10	mA
RAM/REGIS	TER BACKUP					
1	RB data-retention input current	$V_{(RBI)} > V_{(RBI)MIN}$ , $V_{REG25} < V_{IT-}$ , $T_A = 85$ °C		1000	2500	nA
I <sub>(RB)</sub>	No data-retention input current	$V_{(RBI)} > V_{(RBI)MIN}$ , $V_{REG25} < V_{IT-}$ , $T_A = 25$ °C		90	220	ПA
$V_{(RB)}$	RB data-retention input voltage (1)		1.7			V

<sup>(1)</sup> Specified by design. Not production tested.

### 6.8 SMBus Timing Requirements

 $T_A = -40$ °C to 85°C Typical Values at  $T_A = 25$ °C and  $V_{REG25} = 2.5$  V (Unless Otherwise Noted)

			MIN	NOM	MAX	UNIT
f <sub>(SMB)</sub>	SMBus operating frequency	Slave mode, SMBC 50% duty cycle	10		100	kHz
f <sub>(MAS)</sub>	SMBus master clock frequency	Master mode, No clock low slave extend		51.2		kHz
t <sub>(BUF)</sub>	Bus free time between start and stop (see Figure 1)		4.7			μs
t <sub>(HD:STA)</sub>	Hold time after (repeated) start (see Figure 1)		4			μs
t <sub>(SU:STA)</sub>	Repeated start setup time (see Figure 1)		4.7			μs
t <sub>(SU:STO)</sub>	Stop setup time (see Figure 1)		4			μs
	Data hald time (and Figure 4)	Receive mode	0			ns
t(HD:DAT)	Data hold time (see Figure 1)	Transmit mode	300			
t <sub>(SU:DAT)</sub>	Data setup time (see Figure 1)		250			ns
t <sub>(TIMEOUT)</sub>	Error signal/detect (see Figure 1)	See (1)	25		35	μs
t <sub>(LOW)</sub>	Clock low period (see Figure 1)		4.7			μs
t <sub>(HIGH)</sub>	Clock high period (see Figure 1)	See (2)	4		50	μs
t <sub>(LOW:SEXT)</sub>	Cumulative clock low slave extend time	See (3)			25	ms
t <sub>(LOW:MEXT)</sub>	Cumulative clock low master extend time (see Figure 1)	See <sup>(4)</sup>			10	ms
t <sub>f</sub>	Clock/data fall time	See <sup>(5)</sup>			300	ns
t <sub>r</sub>	Clock/data rise time	See <sup>(6)</sup>			1000	ns

The bq20z655-R1 times out when any clock low exceeds  $t_{(TIMEOUT)}$ .  $t_{(HIGH)}$ , Max, is the minimum bus idle time. SMBC = SMBD = 1 for t > 50 ms causes reset of any transaction involving bq20z655-R1 that is in progress. This specification is valid when the NC\_SMB control bit remains in the default cleared state (CLK[0]=0).

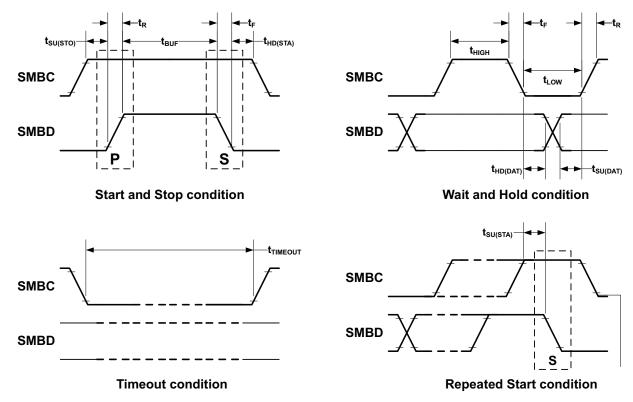
 $t_{(LOW:SEXT)}$  is the cumulative time a slave device is allowed to extend the clock cycles in one message from initial start to the stop.

t(LOW:MEXT) is the cumulative time a master device is allowed to extend the clock cycles in one message from initial start to the stop.

Rise time  $t_r = VILMAX - 0.15$ ) to (VIHMIN + 0.15)

Fall time  $t_f = 0.9 V_{DD}$  to (VILMAX – 0.15)





A. SCLKACK is the acknowledge-related clock pulse generated by the master.

Figure 1. SMBus Timing Diagram

### 6.9 Typical Characteristics

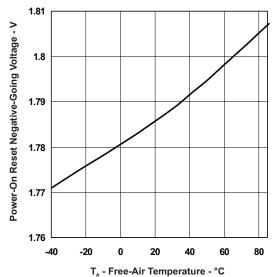


Figure 2. Power On Reset Behavior vs Free-Air Temperature

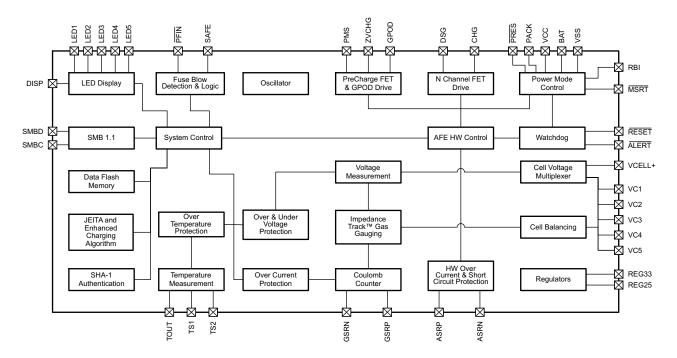


### 7 Detailed Description

#### 7.1 Overview

The bq20z655-R1 incorporating patented Impedance Track™ technology is a single IC solution designed for battery-pack or in-system installation. This SBS-compliant gas gauge and protection IC implemented with Impedance Track™ gas gauging technology continuously analyzes the battery impedance, resulting in superior gas-gauging accuracy.

### 7.2 Functional Block Diagram



### 7.3 Feature Description

### 7.3.1 Feature Set

#### 7.3.1.1 Primary (1st Level) Safety Features

The bq20z655-R1 supports a wide range of battery and system protection features that can easily be configured. The primary safety features include:

- Cell over/undervoltage protection
- · Charge and discharge overcurrent
- Short Circuit protection
- Charge and discharge overtemperature with independent alarms and thresholds for each thermistor
- AFE Watchdog

### 7.3.1.2 Secondary (2nd Level) Safety Features

The secondary safety features of the bq20z655-R1 can be used to indicate more serious faults through the SAFE pin. This pin can be used to blow an in-line fuse to permanently disable the battery pack from charging or discharging. The secondary safety protection features include:

- Safety overvoltage
- Safety undervoltage
- 2nd level protection IC input
- Safety overcurrent in charge and discharge



#### **Feature Description (continued)**

- Safety over-temperature in charge and discharge with independent alarms and thresholds for each thermistor
- · Charge FET and zero-volt charge FET fault
- Discharge FET fault
- Cell imbalance detection (active and at rest)
- Open thermistor detection
- · Fuse blow detection
- AFE communication fault

### 7.3.1.3 Charge Control Features

The bq20z655-R1 charge control features include:

- Supports JEITA temperature ranges. Reports charging voltage and charging current according to the active temperature range.
- Handles more complex charging profiles. Allows for splitting the standard temperature range into two subranges and allows for varying the charging current according to the cell voltage.
- Reports the appropriate charging current needed for constant current charging and the appropriate charging voltage needed for constant voltage charging to a smart charger using SMBus broadcasts.
- Determines the chemical state of charge of each battery cell using Impedance Track™ and can reduce the
  charge difference of the battery cells in fully charged state of the battery pack gradually using cell balancing
  algorithm during charging. This prevents fully charged cells from overcharging and causing excessive
  degradation and also increases the usable pack energy by preventing premature charge termination
- · Supports pre-charging and zero-volt charging
- Supports charge inhibit and charge suspend if battery pack temperature is out of temperature range
- · Reports charging fault and also indicate charge status through charge and discharge alarms.

### 7.3.1.4 Gas Gauging

The bq20z655-R1 uses the Impedance Track™ Technology to measure and calculate the available charge in battery cells. The achievable accuracy is better than 1% error over the lifetime of the battery and there is no full charge discharge learning cycle required.

See the *Theory and Implementation of Impedance Track Battery Fuel-Gauging Algorithm* application note (SLUA364) for further details.

### 7.3.1.5 Lifetime Data Logging Features

The bq20z655-R1 offers lifetime data logging, where important measurements are stored for warranty and analysis purposes. The data monitored include:

- Lifetime maximum temperature
- Lifetime maximum temperature count
- Lifetime maximum temperature duration
- Lifetime minimum temperature
- Lifetime maximum battery cell voltage
- Lifetime maximum battery cell voltage count
- Lifetime maximum battery cell voltage duration
- Lifetime minimum battery cell voltage
- Lifetime maximum battery pack voltage
- Lifetime minimum battery pack voltage
- Lifetime maximum charge current
- Lifetime maximum discharge current
- Lifetime maximum charge power
- Lifetime maximum discharge power
- Lifetime maximum average discharge current
- Lifetime maximum average discharge power



### **Feature Description (continued)**

Lifetime average temperature

#### 7.3.1.6 Authentication

The bg20z655-R1 supports authentication by the host using SHA-1.

### 7.3.2 Battery Parameter Measurements

The bq20z655-R1 uses an integrating delta-sigma analog-to-digital converter (ADC) for current measurement, and a second delta-sigma ADC for individual cell and battery voltage, and temperature measurement.

### 7.3.2.1 Charge and Discharge Counting

The integrating delta-sigma ADC measures the charge/discharge flow of the battery by measuring the voltage drop across a small-value sense resistor between the SR1 and SR2 pins. The integrating ADC measures bipolar signals from -0.25 V to 0.25 V. The bq20z655-R1 detects charge activity when  $V_{SR} = V_{(SRP)}$ - $V_{(SRN)}$  is positive and discharge activity when  $V_{SR} = V_{(SRP)}$ - $V_{(SRN)}$  is negative. The bq20z655-R1 continuously integrates the signal over time, using an internal counter. The fundamental rate of the counter is 0.65 nVh.

### 7.3.2.2 Voltage

The bq20z655-R1 updates the individual series cell voltages at one second intervals. The internal ADC of the bq20z655-R1 measures the voltage, scales and calibrates it appropriately. This data is also used to calculate the impedance of the cell for the Impedance Track™ gas-gauging.

#### 7.3.2.3 Current

The bq20z655-R1 uses the SRP and SRN inputs to measure and calculate the battery charge and discharge current using a 5-m $\Omega$  to 20-m $\Omega$  typical sense resistor.

#### 7.3.2.4 Wake Function

The bq20z655-R1 can exit sleep mode, if enabled, by the presence of a programmable level of current signal across SRP and SRN.

#### 7.3.2.5 Auto Calibration

The bq20z655-R1 provides an auto-calibration feature to cancel the voltage offset error across SRN and SRP for maximum charge measurement accuracy. The bq20z655-R1 performs auto-calibration when the SMBus lines stay low continuously for a minimum of a programmable amount of time.

### 7.3.2.6 Temperature

The bq20z655-R1 has an internal temperature sensor and 2 external temperature sensor inputs, TS1 and TS2, used in conjunction with two identical NTC thermistors (default are Semitec 103AT) to sense the battery environmental temperature. The bq20z655-R1 can be configured to use the internal temperature sensor or up to 2 external temperature sensors.

#### 7.4 Device Functional Modes

### 7.4.1 Power Modes

The bq20z655-R1 supports three different power modes to reduce power consumption:

- In Normal Mode, the bq20z655-R1 performs measurements, calculations, protection decisions and data updates in 1 second intervals. Between these intervals, the bq20z655-R1 is in a reduced power stage.
- In Sleep Mode, the bq20z655-R1 performs measurements, calculations, protection decisions and data update
  in adjustable time intervals. Between these intervals, the bq20z655-R1 is in a reduced power stage. The
  bq20z655-R1 has a wake function that enables exit from Sleep mode, when current flow or failure is detected.
- In Shutdown Mode, the bq20z655-R1 is completely disabled.



### 7.5 Programming

### 7.5.1 Configuration

#### 7.5.1.1 Oscillator Function

The bq20z655-R1 fully integrates the system oscillators therefore, no external components are required for this feature.

### 7.5.1.2 System Present Operation

The bq20z655-R1 periodically verifies the PRES pin and detects that the battery is present in the system through a low state on a PRES input. When this occurs, the bq20z655-R1 enters normal operating mode. When the pack is removed from the system and the PRES input is high, the bq20z655-R1 enters the battery-removed state, disabling the charge, discharge, and ZVCHG FETs. The PRES input is ignored and can be left floating when non-removal mode is set in the data flash.

#### 7.5.2 Communications

The bq20z655-R1 uses SMBus v1.1 with Master Mode and package error checking (PEC) options per the SBS specification.

#### 7.5.2.1 SMBus On and Off State

The bq20z655-R1 detects an SMBus off state when SMBC and SMBD are logic-low for ≥ 2 seconds. Clearing this state requires either SMBC or SMBD to transition high. Within 1 ms, the communication bus is available.



### **Programming (continued)**

### 7.5.3 SBS Commands

**Table 1. SBS Commands** 

SBS				SIZE IN	MIN	MAX	DEFAULT	
CMD	MODE	NAME	FORMAT	BYTES	VALUE	VALUE	VALUE	UNIT
0x00	R/W	ManufacturerAccess	Hex	2	0x0000	Oxffff	_	_
0x01	R/W	RemainingCapacityAlarm	Integer	2	0	700 or 1000	300 or 432	mAh or 10 mWh
0x02	R/W	RemainingTimeAlarm	Unsigned integer	2	0	30	10	min
0x03	R/W	BatteryMode	Hex	2	0x0000	Oxffff	_	_
0x04	R/W	AtRate	Integer	2	-32,768	32,767		mA or 10 mW
0x05	R	AtRateTimeToFull	Unsigned integer	2	0	65,535	_	min
0x06	R	AtRateTimeToEmpty	Unsigned integer	2	0	65,535	_	min
0x07	R	AtRateOK	Unsigned integer	2	0	65,535		_
0x08	R	Temperature	Unsigned integer	2	0	65,535	_	0.1°K
0x09	R	Voltage	Unsigned integer	2	0	20,000	_	mV
0x0a	R	Current	Integer	2	-32,768	32767	_	mA
0x0b	R	AverageCurrent	Integer	2	-32,768	32,767	_	mA
0x0c	R	MaxError	Unsigned integer	1	0	100	_	%
0x0d	R	RelativeStateOfCharge	Unsigned integer	1	0	100	_	%
0x0e	R	AbsoluteStateOfCharge	Unsigned integer	1	0	100+	_	%
0x0f	R/W	RemainingCapacity	Unsigned integer	2	0	65,535	_	mAh or 10 mWh
0x10	R	FullChargeCapacity	Unsigned integer	2	0	65,535	_	mAh or 10 mWh
0x11	R	RunTimeToEmpty	Unsigned integer	2	0	65,534	_	min
0x12	R	AverageTimeToEmpty	Unsigned integer	2	0	65,534	_	min
0x13	R	AverageTimeToFull	Unsigned integer	2	0	65,534	_	min
0x14	R	ChargingCurrent	Unsigned integer	2	0	65,534	_	mA
0x15	R	ChargingVoltage	Unsigned integer	2	0	65,534	_	mV
0x16	R	BatteryStatus	Hex	2	0x0000	0xdbff	_	_
0x17	R/W	CycleCount	Unsigned integer	2	0	65,535	0	_
0x18	R/W	DesignCapacity	Integer	2	0	32,767	4400 or 6336	mAh or 10 mWh
0x19	R/W	DesignVoltage	Integer	2	7000	18,000	14,400	mV
0x1a	R/W	SpecificationInfo	Hex	2	0x0000	0xffff	0x0031	_
0x1b	R/W	ManufactureDate	Unsigned integer	2	0	65,535	0	_
0x1c	R/W	SerialNumber	Hex	2	0x0000	0xffff	0x0000	_
0x20	R/W	ManufacturerName	String	20+1	_		Texas Instruments	_



### **Programming (continued)**

### **Table 1. SBS Commands (continued)**

SBS CMD	MODE	NAME	FORMAT	SIZE IN BYTES	MIN VALUE	MAX VALUE	DEFAULT VALUE	UNIT
0x21	R/W	DeviceName	String	20+1	_	_	bq20z655-R1	_
0x22	R/W	DeviceChemistry	String	4+1	_	-	LION	_
0x23	R	ManufacturerData	String	14+1	_	_	_	_
0x2f	R/W	Authenticate	String	20+1	_	_	_	_
0x3c	R	CellVoltage4	Unsigned integer	2	0	65,535	_	mV
0x3d	R	CellVoltage3	Unsigned integer	2	0	65,535	_	mV
0x3e	R	CellVoltage2	Unsigned integer	2	0	65,535	_	mV
0x3f	R	CellVoltage1	Unsigned integer	2	0	65,535		mV

### **Table 2. Extended SBS Commands**

SBS CMD	MODE	NAME	FORMAT	SIZE IN BYTES	MIN VALUE	MAX VALUE	DEFAULT VALUE	UNIT
0x45	R	AFEData	String	11+1	_	_	_	_
0x46	R/W	FETControl	Hex	2	0x00	0xff	_	_
0x4f	R	StateOfHealth	Hex	2	0x0000	0xffff	_	%
0x51	R	SafetyStatus	Hex	2	0x0000	0xffff	_	_
0x52	R	PFAlert	Hex	2	0x0000	0xffff	_	_
0x53	R	PFStatus	Hex	2	0x0000	0xffff	_	_
0x54	R	OperationStatus	Hex	2	0x0000	0xffff	_	_
0x55	R	ChargingStatus	Hex	2	0x0000	0xffff	_	_
0x57	R	ResetData	Hex	2	0x0000	0xffff	_	_
0x58	R	WDResetData	Unsigned integer	2	0	65,535	_	_
0x5a	R	PackVoltage	Unsigned integer	2	0	65,535	_	mV
0x5d	R	AverageVoltage	Unsigned integer	2	0	65,535	_	mV
0x5e	R	TS1Temperature	Integer	2	-400	1200	_	0.1°C
0x5f	R	TS2Temperature	Integer	2	-400	1200	_	0.1°C
0x60	R/W	UnSealKey	Hex	4	0x00000000	0xfffffff	_	_
0x61	R/W	FullAccessKey	Hex	4	0x00000000	0xfffffff	_	_
0x62	R/W	PFKey	Hex	4	0x00000000	0xfffffff	_	_
0x63	R/W	AuthenKey3	Hex	4	0x00000000	0xfffffff	_	_
0x64	R/W	AuthenKey2	Hex	4	0x00000000	0xfffffff	_	_
0x65	R/W	AuthenKey1	Hex	4	0x00000000	0xfffffff	_	_
0x66	R/W	AuthenKey0	Hex	4	0x00000000	0xfffffff	_	_
0x68	R	SafetyAlert2	Hex	2	0x0000	0x000f	_	_
0x69	R	SafetyStatus2	Hex	2	0x0000	0x000f	_	_
0x6a	R	PFAlert2	Hex	2	0x0000	0x000f	_	_
0x6b	R	PFStatus2	Hex	2	0x0000	0x000f	_	_
0x6c	R	ManufBlock1	String	20	_	_	_	_
0x6d	R	ManufBlock2	String	20	_	_	_	_
0x6e	R	ManufBlock3	String	20	_		_	_



### Table 2. Extended SBS Commands (continued)

SBS CMD	MODE	NAME	FORMAT	SIZE IN BYTES	MIN VALUE	MAX VALUE	DEFAULT VALUE	UNIT
0x6f	R	ManufBlock4	String	20	_	_	_	
0x70	R/W	ManufacturerInfo	String	31+1	_	_	_	_
0x71	R/W	SenseResistor	Unsigned integer	2	0	65,535	_	μΩ
0x72	R	TempRange	Hex	2	_	_	_	_
0x73	R	LifetimeData1	String	32+1	_	_	_	_
0x74	R	LifetimeData2	String	8+1	_	_	_	
0x77	R/W	DataFlashSubClassID	Hex	2	0x0000	Oxffff	_	
0x78	R/W	DataFlashSubClassPage1	Hex	32	_	_	_	_
0x79	R/W	DataFlashSubClassPage2	Hex	32	_	_	_	_
0x7a	R/W	DataFlashSubClassPage3	Hex	32	_	_	_	_
0x7b	R/W	DataFlashSubClassPage4	Hex	32	_	_	_	_
0x7c	R/W	DataFlashSubClassPage5	Hex	32	_	_	_	_
0x7d	R/W	DataFlashSubClassPage6	Hex	32	_	_	_	_
0x7e	R/W	DataFlashSubClassPage7	Hex	32	_	_	_	_
0x7f	R/W	DataFlashSubClassPage8	Hex	32	_	_	_	_



### 8 Application and Implementation

#### NOTE

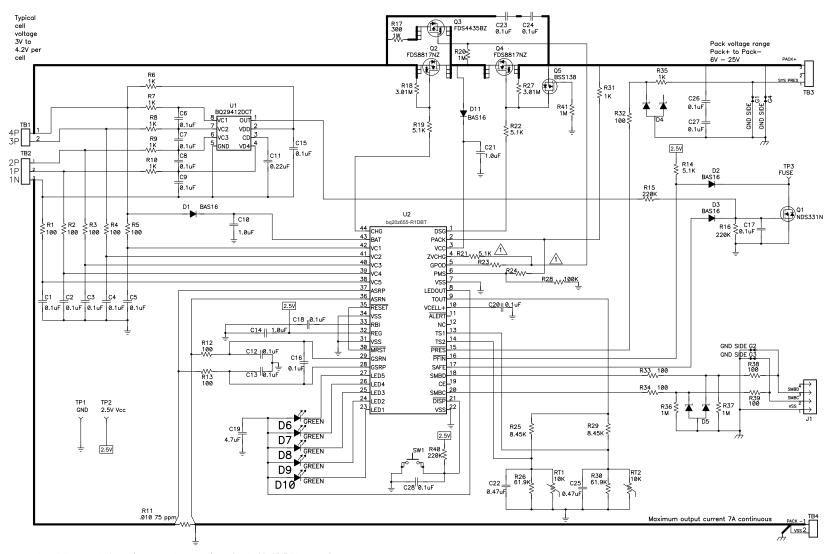
Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 8.1 Application Information

The bq20z655-R1 is a gas gauge with primary protection support, and that can be used with a 2-series to 4-series Li-lon/Li Polymer battery pack. To implement and design a comprehensive set of parameters for a specific battery pack, users need the BQEV graphical user-interface tool installed on a PC during development. The firmware installed on the BQEV tool has default values for this product, which are summarized in the bq20z655 Technical Reference Manual (SLUU493). Using the tool, BQEV these default values can be changed to cater to specific application requirements during development once the system parameters, such as fault trigger thresholds for protection, enable/disable of certain features for operation, configuration of cells, chemistry that best matches the cell used, and more are known. This data is referred to as the *golden image*.

## TEXAS INSTRUMENTS

### 8.2 Typical Application



Install these parts for various pre-charge options. See bq20z65 EVM user guide.  $\hfill \triangle$  Default configuration has R23 & R24 NOT installed.

Figure 3. Application Schematic



#### 8.2.1 Design Requirements

Table 3 shows the default settings for the main parameters. Use the BQEV tool to update the settings to meet the specific application or battery pack configuration requirements.

**Table 3. Design Parameters** 

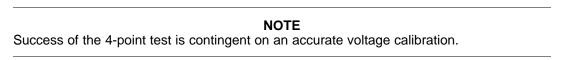
PARAMETER	EXAMPLE VALUE
Cell configuration	4s1p (4 series with 1 parallel)
Design capacity	4400 mAH
Device chemistry	0100 (LION)
Cell overvoltage at standard temperature	4300 mV
Cell undervoltage	2200 mV
Cell Shutdown voltage	1750 mV
Overcurrent in CHARGE mode	6000 mA
Overcurrent in DISCHARGE mode	−6000 mA
Short circuit in CHARGE mode	0.1 V/Rsense across SRP, SRN
Short circuit in DISCHARGE mode	0.1 V/Rsense across SRP, SRN
Safety overvoltage	4500 mV
Cell balancing	Disabled
Internal and external temperature sensor	External temperature sensor is used
Undertemperature charging	0°C
Undertemperature discharging	0°C
BROADCAST mode	Disabled
Battery Trip Point (BTP) with active high interrupt	Disabled

#### 8.2.2 Detailed Design Procedure

#### 8.2.2.1 Choosing the Correct Chemistry

For the Impedance Track™ algorithm to work properly, the exact chemistry of the lithium cells needs to be known and the correct .SENC file needs to be loaded.

If you are using the bqEASY design wizard, it asks you to choose the correct chemistry from a list of manufacturers and model numbers, or test for a compatible chemistry using a 4-point test.



The process for updating the .SENC file is outlined in detail in the application report Updating Firmware With The bq20zxx and EVM.

### 8.2.2.2 High-Current Path

The high-current path begins at the PACK+ terminal of the battery pack. As charge current travels through the pack, it finds its way through protection FETs, a chemical fuse, the lithium-ion cells and cell connections, and the sense resistor, and then returns to the PACK- terminal. In addition, some components are placed across the PACK+ and PACK- terminals to reduce effects from electrostatic discharge.

### 8.2.2.3 Protection FETs

Select the N-channel charge and discharge FETs for a given application. Most portable battery applications are a good match for the CSD17308Q3. The TI CSD17308Q3 is a 47-A, 30-V device with Rds(on) of 8.2 m $\Omega$  when the gate drive voltage is 8 V.

If a precharge FET is used, R1 is calculated to limit the precharge current to the desired rate. Be sure to account for the power dissipation of the series resistor. The precharge current is limited to (VCHARGER – VBAT)/R1 and maximum power dissipation is (Vcharger – Vbat)2/R1.



The gates of all protection FETs are pulled to the source with a high-value resistor between the gate and source to ensure they are turned off if the gate drive is open.

Capacitors C1 and C2 help protect the FETs during an ESD event. Using two devices ensures normal operation if one becomes shorted. To have good ESD protection, the copper trace inductance of the capacitor leads must be designed to be as short and wide as possible. Ensure that the voltage rating of both C1 and C2 are adequate to hold off the applied voltage if one of the capacitors becomes shorted.

#### 8.2.2.4 Lithium-Ion Cell Connections

The important part to remember about the cell connections is that high current flows through the top and bottom connections; therefore, the voltage sense leads at these points must be made with a Kelvin connection to avoid any errors due to a drop in the high-current copper trace. The location marked 4P in indicates the Kelvin connection of the most positive battery node.

#### 8.2.2.5 Sense Resistor

As with the cell connections, the quality of the Kelvin connections at the sense resistor is critical. The sense resistor must have a temperature coefficient no greater than 50 ppm to minimize current measurement drift with temperature. Choose the value of the sense resistor to correspond to the available overcurrent and short circuit ranges of the bq20z655. Select the smallest value possible to minimize the negative voltage generated on the VSS nodes during a short circuit.

#### 8.2.2.6 ESD Mitigation

A pair of series 0.1-µF ceramic capacitors is placed across the PACK+ and PACK- terminals to help in the mitigation of external electrostatic discharges. The two devices in series ensure continued operation of the pack if one of the capacitors becomes shorted. Optionally, a tranzorb such as the SMBJ2A can be placed across the terminals to further improve ESD immunity.

#### 8.2.2.7 System Present

The System Present signal is used to inform the gas gauge whether the pack is installed into or removed from the system. In the host system, this pin is grounded. The PRES pin of the bq20z655 is occasionally sampled to test for system present. To save power, an internal pullup is provided by the gas gauge during a brief 4- $\mu$ s sampling pulse once per second. A resistor can be used to pull the signal low and the resistance must be 20 k $\Omega$  or lower to insure that the test pulse is lower than the VIL limit. The pullup current source is typically 10  $\mu$ A to 20  $\mu$ A.

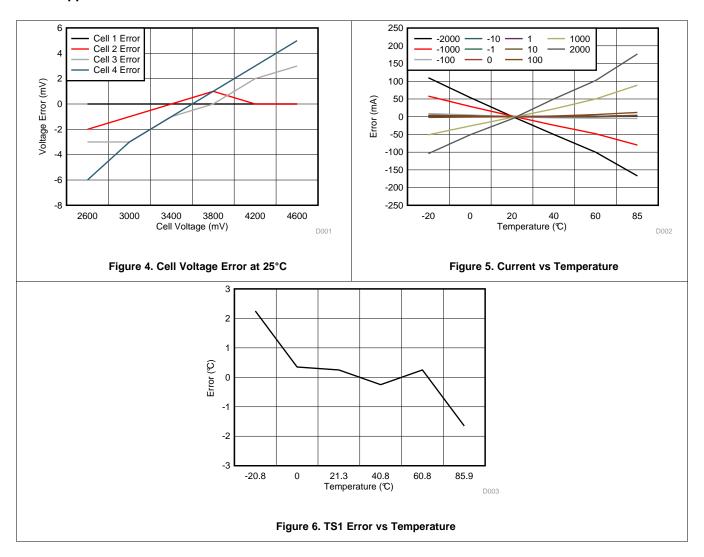
Because the System Present signal is part of the pack connector interface to the outside world, it must be protected from external electrostatic discharge events. An integrated ESD protection on the PRES device pin reduces the external protection requirement to just R29 for an 8-kV ESD contact rating. However, if it is possible that the System Present signal may short to PACK+, then a resistor, diode combo must be included for high-voltage protection.

### 8.2.2.8 SMBus Communication

The SMBus clock and data pins have integrated high-voltage ESD protection circuits, however, adding a Zener diode and series resistor provides more robust ESD performance.



#### 8.2.3 Application Curves



### 9 Power Supply Recommendations

The device manages its supply voltage dynamically according to the operation conditions. Normally, the BAT input is the primary power source to the device. The BAT pin should be connected to the positive termination of the battery stack. The input voltage for the BAT pin ranges from 4.5 V to 25 V. The VCC pin is the secondary power input, which activates when the BAT voltage falls below minimum Vcc. This allows the device to source power from a charger (if present) connected to the PACK pin. The VCC pin should be connected to the common drain of the CHG and DSG FETs. The charger input should be connected to the PACK pin.



### 10 Layout

### 10.1 Layout Guidelines

A battery fuel gauge circuit board is a challenging environment due to the fundamental incompatibility of high-current traces and ultra-low current semiconductor devices. The best way to protect against unwanted trace-to-trace coupling is with a component placement, such as that shown in Figure 11, where the high-current section is on the opposite side of the board from the electronic devices. Clearly this is not possible in many situations due to mechanical constraints. Still, every attempt should be made to route high-current traces away from signal traces, which enter the directly. IC references and registers can be disturbed and in rare cases damaged due to magnetic and capacitive coupling from the high-current path. During surge current and ESD events, the high-current traces appear inductive and can couple unwanted noise into sensitive nodes of the gas gauge electronics, as illustrated in Figure 12.

Kelvin voltage sensing is extremely important to accurately measure current and top and bottom cell voltages. Place all filter components as close as possible to the device. Route the traces from the sense resistor in parallel to the filter circuit. Adding a ground plane around the filter network can add additional noise immunity. Figure 7 and Figure 8 demonstrates correct kelvin current sensing.

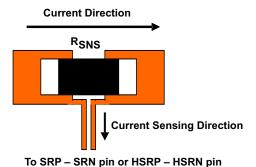


Figure 7. Sensing Resistor PCB Layout

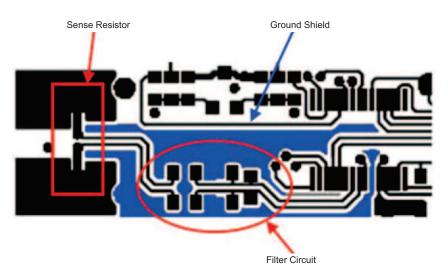


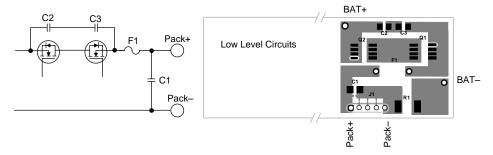
Figure 8. Sense Resistor, Ground Shield, and Filter Circuit Layout

#### 10.1.1 Protector FET Bypass and Pack Terminal Bypass Capacitors

The general principle is to use wide copper traces to lower the inductance of the bypass capacitor circuit. In Figure 9, an example layout demonstrates this technique.



### **Layout Guidelines (continued)**



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Figure 9. Use Wide Copper Traces to Lower the Inductance of Bypass Capacitors C1, C2, and C3

### 10.1.2 ESD Spark Gap

Protect SMBus Clock, Data, and other communication lines from ESD with a spark gap at the connector. The pattern in Figure 10 recommended, with 0.2-mm spacing between the points.

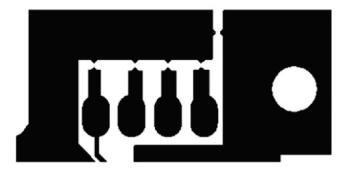


Figure 10. Recommended Spark-Gap Pattern Helps Protect Communication Lines from ESD



### 10.2 Layout Example

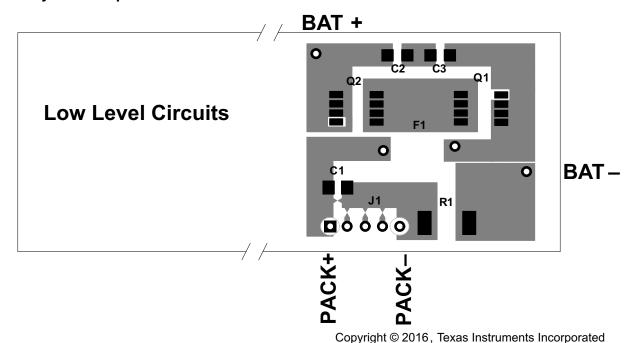


Figure 11. Separating High- and Low-Current Sections Provides an Advantage in Noise Immunity

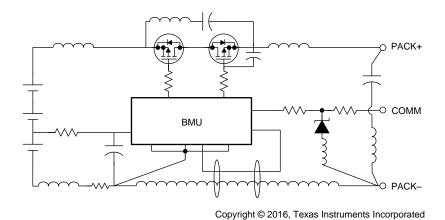


Figure 12. Avoid Close Spacing Between High-Current and Low-Level Signal Lines



### 11 器件和文档支持

### 11.1 文档支持

#### 11.1.1 相关文档

请参阅如下相关文档:

- 《bg20z655 技术参考手册》, SLUU493
- 《采用 Impedance Track 技术的 bg20z655EVM 和 bg34z651EVM SBS 1.1 评估模块》,SLUU697
- 《bg20zxx 电量监测计系列快速入门指南》, SLUA421

有关与 bq20zXX 系列相关的其他应用手册,请访问 bq20z80 页面www.ti.com.cn/上)。

### 11.2 社区资源

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**Design Support** *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

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### 11.4 静电放电警告



这些装置包含有限的内置 ESD 保护。 存储或装卸时,应将导线一起截短或将装置放置于导电泡棉中,以防止 MOS 门极遭受静电损伤。

#### 11.5 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

### 12 机械、封装和可订购信息

以下页面包括机械、封装和可订购信息。这些信息是指定器件的最新可用数据。这些数据发生变化时,我们可能不会另行通知或修订此文档。如欲获取此产品说明书的浏览器版本,请参见左侧的导航栏。



### PACKAGE OPTION ADDENDUM

11-Dec-2020

#### PACKAGING INFORMATION

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Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
BQ20Z655DBT-R1	ACTIVE	TSSOP	DBT	44	40	RoHS & Green	NIPDAU	Level-2-250C-1 YEAR	-40 to 85	BQ20Z655	Samples
BQ20Z655DBTR-R1	ACTIVE	TSSOP	DBT	44	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	BQ20Z655	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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11-Dec-2020

### PACKAGE MATERIALS INFORMATION

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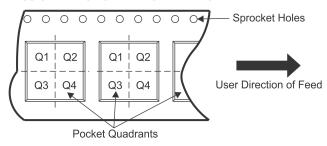
### TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

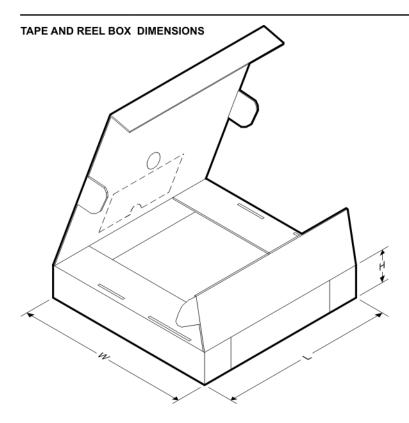
### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



### \*All dimensions are nominal

Device	Package Type	Package Drawing			Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
BQ20Z655DBTR-R1	TSSOP	DBT	44	2000	330.0	24.4	6.8	11.7	1.6	12.0	24.0	Q1

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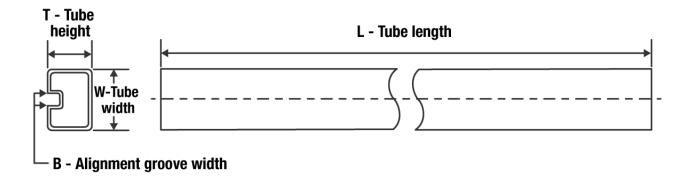
#### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
BQ20Z655DBTR-R1	TSSOP	DBT	44	2000	350.0	350.0	43.0

### **PACKAGE MATERIALS INFORMATION**

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### **TUBE**

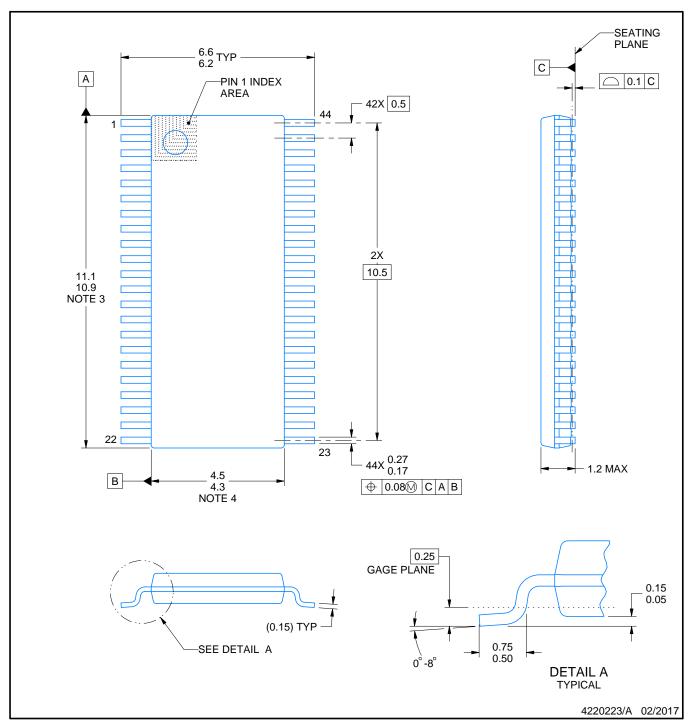


#### \*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
BQ20Z655DBT-R1	DBT	TSSOP	44	40	530	10.2	3600	3.5



SMALL OUTLINE PACKAGE



### NOTES:

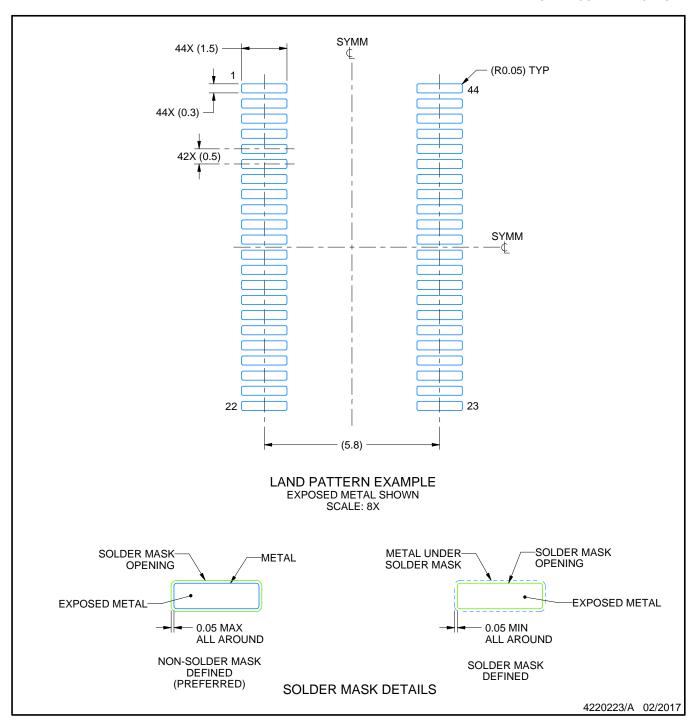
- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.



SMALL OUTLINE PACKAGE

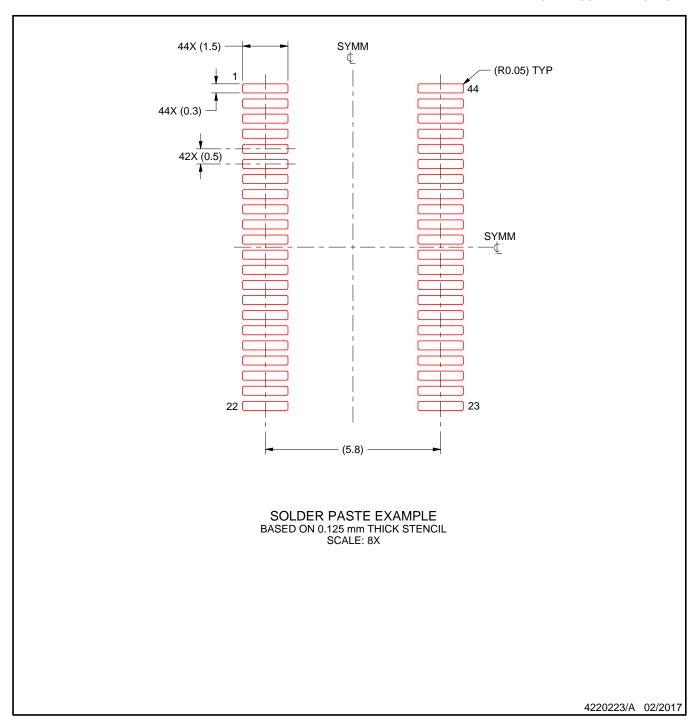


NOTES: (continued)

- 5. Publication IPC-7351 may have alternate designs.
- 6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE PACKAGE



NOTES: (continued)

- Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 8. Board assembly site may have different recommendations for stencil design.



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