













### ISO721, ISO721M, ISO722, ISO722M

SLLS629L - JANUARY 2006-REVISED OCTOBER 2015

# **ISO72x Single Channel High-Speed Digital Isolators**

#### **Features**

- 100 and 150-Mbps Signaling Rate Options
- Low Propagation Delay
- Low Pulse Skew
- Low-Power Sleep Mode
- High Electromagnetic Immunity
- Low Input-Current Requirement
- Failsafe Output
- Drop-In Replacement for Most Opto and Magnetic **Isolators**
- Operates from 3.3 V and 5 V Supplies
- -40°C to 125°C Operating Temperature Range
- 50 kV/µs Transient Immunity, Typical
- Safety and Regulatory Approvals
  - VDE Basic Insulation with 4000-V<sub>PK</sub> V<sub>IOTM</sub>, 560  $V_{PK} V_{IORM}$
  - 2500 V<sub>RMS</sub> Isolation per UL 1577
  - CSA Approved for Component Acceptance Notice 5A and IEC 60950-1

# 2 Applications

- Industrial Fieldbus
  - Modbus
  - Profibus
  - DeviceNet<sup>™</sup> Data Buses
  - Smart Distributed Systems (SDS™)
- Computer Peripheral Interface
- Servo Control Interface
- **Data Acquisition**

#### 3 Description

The ISO721, ISO721M, ISO722, and ISO722M are digital isolators with a logic input and output buffer separated by a silicon dioxide (SiO<sub>2</sub>) insulation barrier. This barrier provides galvanic isolation of up to 4000 V<sub>PK</sub> per VDE. Used in conjunction with isolated power supplies, these devices prevent noise currents on a data bus or other circuits from entering the local ground, and interfering with or damaging sensitive circuitry.

A binary input signal is conditioned, translated to a balanced signal, then differentiated by the capacitive isolation barrier. Across the isolation barrier, a differential comparator receives the logic transition information, then sets or resets a flip-flop and the output circuit accordingly. A periodic update pulse is sent across the barrier to ensure the proper dc level of the output.

If this dc-refresh pulse is not received for more than 4 us, the input is assumed to be unpowered or not being actively driven, and the failsafe circuit drives the output to a logic-high state.

The symmetry of the dielectric and capacitor within the integrated circuitry provides for close capacitive matching, and allows fast transient voltage changes between the input and output grounds without corrupting the output. The small capacitance and resulting time constant provide for fast operation with signaling rates from 0 Mbps (DC) to 100 Mbps for the ISO721 and the ISO722 devices, and 0 Mbps to 150 Mbps with the ISO721M and the ISO722M devices.

These devices require two supply voltages of 3.3 V, 5 V, or any combination. All inputs are 5-V tolerant when supplied from a 3.3-V supply and all outputs are 4 mA CMOS.

The ISO722 and ISO722M devices include an activelow output enable that when driven to a high logic level, places the output in a high-impedance state and turns off internal bias circuitry to conserve power.

Both the ISO721 and ISO722 devices have TTL input thresholds and a noise filter at the input that prevent transient pulses of up to 2 ns in duration from being passed to the output of the device.

The ISO721M and ISO722M devices have CMOS V<sub>CC</sub> / 2 input thresholds, but do not have the noisefilter and the additional propagation delay. These features of the ISO721M device also provide for reduced-jitter operation.

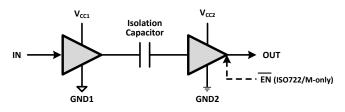
The ISO721, ISO721M, ISO722, and ISO722M devices are characterized for operation over the ambient temperature range of -40°C to 125°C.

#### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)		
ISO721	SOP (8)	9.50mm x 6.57mm		
ISO721				
ISO721M	COIC (8)	4.00		
ISO722	SOIC (8)	4.90mm x 3.91mm		
ISO722M				

(1) For all available packages, see the orderable addendum at the end of the datasheet.

#### Simplified Schematic





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### 4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

C	hanges from Revision K (February 2012) to Revision L	Page
•	Moved Power Dissipation metric into new table, called Power Dissipation	8
•	Added header row above "V <sub>IORM</sub> " row with the text "DIN V VDE V 0884-10 (VDE V 0884-10):2006-12" in the Insulation Characteristics table	19
•	Added "UL 1577" header row over "V <sub>ISO</sub> " row in the Insulation Characteristics table	19
•	Moved "V <sub>ISO</sub> " row to the bottom of the Insulation Characteristics table.	19
•	Deleted "per UL" in "Isolation voltage" in the Insulation Characteristics table.	19
•	Changed the D-8 MIN value of L(101) from "4.8" to "4" in the Package Insulation Characteristics table	20
•	Changed the D-8 MIN value of L(102) from "4.3" to "4" in the Package Insulation Characteristics table	20
•	Changed Test Condition "DIN IEC 60112/VDE 0303 Part 1" to "DIN EN 60112 (VDE 0303-11); IEC 60112" in the Package Insulation Characteristics table.	
•	Deleted bottom row of the Package Insulation Characteristics table.	20

#### Changes from Revision J (July 2010) to Revision K

Page

•	Added ESD Rating table, Feature Description section, Device Functional Modes, Application and Implementation
	section, Power Supply Recommendations section, Layout section, Device and Documentation Support section, and
	Mechanical, Packaging, and Orderable Information
•	Changed the Title From: 3.3-V / 5-V High-Speed Digital Isolators To: ISO72x Single Channel High-Speed Digital Isolators

Product Folder Links: ISO721 ISO721M ISO722 ISO722M





•	Changed VDE text From: "DIN EN 60747-5-5 (VDE 0884-5)" To: "DIN V VDE V 0884-10 (VDE V 0884-10):2006-12" in the Regulatory Information table	
•	Changed CSA File number: 1698195 To: 220991 in the <i>Regulatory Information</i> table	
•	Changed the C <sub>TI</sub> MIN value From: ≥ 175 V To: 400 V in the <i>Package Insulation Characteristics</i> table	
•	Changed R <sub>IO</sub> Test Condition From: T <sub>A</sub> < 100°C To: T <sub>A</sub> = 25°C in <i>Package Insulation Characteristics</i>	
•	Moved the R <sub>IO</sub> values from the TYP column to the MIN column of <i>Package Insulation Characteristics</i>	
•	Changed the title of Figure 16 From: $\theta_{JC}$ Thermal Derating Curve per DIN EN 60747-5-5 To: $\theta_{JC}$ Thermal Derating Curve per VDE	
•	Changed Table 1, added row X, PD, X, Undetermined	
•	Changed Table 2, added row X, PD, X, Undetermined	21
•	Changed Figure 17	21
CI	hanges from Revision I (February 2010) to Revision J	Page
•	Changed Note 1 of the <i>Electrical Characteristics</i> , 5 V table	6
•	Changed Note 1 of the <i>Electrical Characteristics</i> , 5 V, 3.3 V table	<mark>7</mark>
•	Changed Note 1 of the <i>Electrical Characteristics</i> , 3.3 V, 5 V table	<mark>7</mark>
•	Changed Note 1 of the <i>Electrical Characteristics</i> , 3.3 V table	
•	Changed V to Vpeak in UNIT column of IEC Insulation Characteristics table	19
•	Added row for V <sub>ISO</sub> to Insulation Characteristics table	19
•	Changed the title From: Package Characteristics To: Package Insulation Characteristics	20
•	hanges from Revision H (June 2009) to Revision I  Changed Features From: 50 kV/s Transient Immunity, Typical To: 50 kV/µs Transient Immunity, Typical	Page
CI	hanges from Revision G (December 2008) to Revision H	Page
•	Changed the first paragraph of the <i>Description</i> From: "silicon oxide (SiO <sub>2</sub> )" To: "silicon dioxide (SiO <sub>2</sub> )"	1
•	Added the DUB 8 pin package to the Pin Configuration and Functions	4
•	Added package designators D-8 and DUB-8 to the table Descriptions/Test Conditions of the <i>Package Insulation Characteristics</i> table	20
_		
CI	hanges from Revision F (November 2008) to Revision G	Page
<u>.</u>	Changed the Features List From: 4000-V <sub>(peak)</sub> Isolation To: 4000-V <sub>(peak)</sub> Isolation, 560-V <sub>peak</sub> V <sub>IORM</sub>	1
CI	hanges from Revision E (May 2008) to Revision F	Page
•	Changed Figure 19 text From: 20 mm max. from V <sub>CC1</sub> To: 2 mm max. from V <sub>CC1</sub>	23
•	Changed Note in Table 3 From: The ISO72x pin 1 and pin 3 are internally connected together. Either or both may be used as $V_{\text{CC1}}$ . To: Pin 1 should be used as $V_{\text{CC1}}$ . Pin 3 may also be used as $V_{\text{CC1}}$ or left open as long as Pin 1 is connected to $V_{\text{CC1}}$	23
•	Changed Note in Table 3 From: The ISO721 and ISO721M pin 5 and pin 7 are internally connected together. Either or both may be used as GND2. To: Pin 5 should be used as GND2. Pin 7 may also be used as GND2 or left open as long as Pin 5 is connected to GND2.	



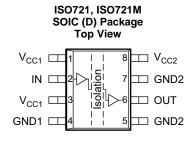
Changes from Revision D (February 2007) to Revision E  Changed changed the V <sub>CC</sub> MIN value From: 4.5 V To: 3 V in the Recommended Operating Conditions table  Added Note 1 to the Recommended Operating Conditions table  Added Note 1 to the Electrical Characteristics, 5 V table  Added Note 1 to the Electrical Characteristics, 5 V, 3.3 V table  Added Note 1 to the Electrical Characteristics, 3.3 V, 5 V table  Added Note 1 to the Electrical Characteristics, 3.3 V table	Page	
•	Changed changed the V <sub>CC</sub> MIN value From: 4.5 V To: 3 V in the Recommended Operating Conditions table	
•	Added Note 1 to the Recommended Operating Conditions table	
•	Added Note 1 to the <i>Electrical Characteristics</i> , 5 V table	
•	Added Note 1 to the <i>Electrical Characteristics</i> , 5 V, 3.3 V table	
•	Added Note 1 to the <i>Electrical Characteristics</i> , 3.3 V, 5 V table	
•	Added Note 1 to the <i>Electrical Characteristics</i> , 3.3 V table	8

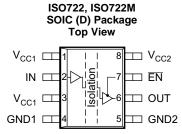


# 5 Device Comparison Table

PRODUCT	SIGNALING RATE	OUTPUT ENABLED	INPUT THRESHOLDS	NOISE FILTER
ISO721	100 Mbps	NO	TTL	YES
ISO721M	150 Mbps	NO	CMOS	NO
ISO722	100 Mbps	YES	TTL	YES
ISO722M	150 Mbps	YES	CMOS	NO

# 6 Pin Configuration and Functions





**ISO721** 

**Pin Functions** 

	PIN			
NAME	ISO721x NO.	ISO722x NO.	I/O	DESCRIPTION
V <sub>CC1</sub>	1, 3	1, 3	-	Power supply, V <sub>CC1</sub>
V <sub>CC2</sub>	8	8	1	Power supply, V <sub>CC2</sub>
IN	2	2	I	Input
OUT	6	6	0	Output
EN	-	7	I	Output enable. OUT is enabled when $\overline{EN}$ is low or disconnected and disabled when $\overline{EN}$ is high.
GND1	4	4	-	Ground connection for V <sub>CC1</sub>
GND2	5, 7	5	-	Ground connection for V <sub>CC2</sub>



# 7 Specifications

# 7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

			MIN	MAX	UNIT
$V_{CC}$	Supply voltage	V <sub>CC1</sub> , V <sub>CC2</sub>	-0.5	6	٧
VI	Input voltage	IN, OUT, or EN	-0.5	$V_{CC} + 0.5^{(2)}$	V
Io	Output current	,		±15	mA
TJ	Maximum junction temperature			170	°C
T <sub>stg</sub>	Storage temperature		-65	150	°C

<sup>(1)</sup> Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

# 7.2 ESD Ratings

			VALUE	UNIT
		Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	±2000	
V <sub>(ESD)</sub>	Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	±1000	V

<sup>(1)</sup> JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

# 7.3 Recommended Operating Conditions

			MIN	TYP	MAX	UNIT
V <sub>CC</sub>	Supply voltage <sup>(1)</sup> , V <sub>CC1</sub> , V <sub>CC2</sub>		3		5.5	V
I <sub>OH</sub>	Outroit summent				4	mA
I <sub>OL</sub>	Output current		-4			mA
t <sub>ui</sub>	Input nulse duration	ISO72x	10			
	Input pulse duration	ISO72xM	6.67			ns
4 / 4	Signaling Rate	ISO72x	0		100	Mhana
1 / t <sub>ui</sub>		ISO72xM	0		150	Mbps
V <sub>IH</sub>	High-level input voltage (IN, EN)	10070	2		5.5	V
$V_{IL}$	Low-level input voltage (IN, EN)	ISO72x	0		0.8	V
V <sub>IH</sub>	High-level input voltage (IN, EN)	10070.14	0.7 V <sub>CC</sub>		$V_{CC}$	V
$V_{IL}$	Low-level input voltage (IN, EN)	IOS72xM	0		0.3 V <sub>CC</sub>	V
T <sub>A</sub>	Ambient temperature		-40	25	125	°C
$T_J$	Junction temperature	See Thermal Information			150	°C
Н	External magnetic field intensity per certification	IEC 61000-4-8 and IEC 61000-4-9			1000	A/m

(1) For the 5-V operation,  $V_{CC1}$  or  $V_{CC2}$  is specified from 4.5 V to 5.5 V. For the 3-V operation,  $V_{CC1}$  or  $V_{CC2}$  is specified from 3 V to 3.6 V.

<sup>(2)</sup> Maximum voltage must not exceed 6 V.

<sup>(2)</sup> JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



#### 7.4 Thermal Information

				ISO721	ISO72x	
	THERMAL METRIC <sup>(1)</sup>				D	UNIT
		8 PINS	8 PINS			
$R_{\theta JA}$	Junction-to-ambient thermal resistance	High-K Board		86.6	114.7	°C/W
		Low-K Board		N/A	263	
R <sub>0JC(top)</sub>	Junction-to-case (top) thermal resistance			70.3	63	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance			50.2	54.8	°C/W
ΨЈТ	Junction-to-top characterization paramete	r		34.3	18.9	°C/W
ΨЈВ	JB Junction-to-board characterization parameter		49.8	54.3	°C/W	
R <sub>0JC(bot)</sub>	Junction-to-case (bottom) thermal resistar	nce		N/A	N/A	°C/W

<sup>(1)</sup> For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.

# 7.5 Electrical Characteristics, 5 V

 $V_{CC1}$  and  $V_{CC2}$  at 5 V  $\pm$  10% (over recommended operating conditions unless otherwise noted.)

	PARAM	ETER	TEST CO	NDITIONS	MIN	TYP	MAX	UNIT
	\/ aumphy augrent	Quiescent	\\ \\ a=0\\\ .	a lood		0.5	1	m Λ
I <sub>CC1</sub>	V <sub>CC1</sub> supply current	25 Mbps	$V_I = V_{CC}$ or 0 V, no load			2	4	mA
		ISO722/722M Sleep Mode	V V 070V	EN at V <sub>CC</sub>			200	μA
I <sub>CC2</sub>	V <sub>CC2</sub> supply current	Quiescent	$V_I = V_{CC}$ or 0 V, No load	EN at 0 V or ISO721/721M		8	12	mA
		25 Mbps	V <sub>I</sub> = V <sub>CC</sub> or 0 V, no load			10	14	
V High lavel autout valtage		I <sub>OH</sub> = -4 mA, See Figure 10		V <sub>CC</sub> - 0.8	4.6		V	
V <sub>OH</sub>	V <sub>OH</sub> High-level output voltage		$I_{OH} = -20 \mu A, Se$	e Figure 10	V <sub>CC</sub> - 0.1	5		V
.,	I am laval antent mate	_	I <sub>OL</sub> = 4 mA, See Figure 10			0.2	0.4	V
V <sub>OL</sub>	Low-level output voltag	е	$I_{OL} = 20 \mu A$ , See	Figure 10		0	0.1	V
V <sub>I(HYS)</sub>	Input voltage hysteresis	3				150		mV
I <sub>IH</sub>	High-level input current		EN, IN at 2 V				10	
I <sub>IL</sub>	I <sub>IL</sub> Low-level input current		EN, IN at 0.8 V		-10			μA
I <sub>OZ</sub>	High-impedance output current	ISO722, ISO722M	EN, IN at V <sub>CC</sub>				1	μΑ
C <sub>I</sub>	C <sub>I</sub> Input capacitance to ground		IN at $V_{CC}$ , $V_I = 0$ .	$4 \sin (4 \times 10^6 \pi t)$		1		pF
CMTI	Common-mode transie	nt immunity	$V_I = V_{CC}$ or 0 V, S	See Figure 14	25	50		kV/μs

Product Folder Links: ISO721 ISO721M ISO722 ISO722M



# 7.6 Electrical Characteristics, 5 V, 3.3 V

 $V_{CC1}$  at 5 V  $\pm$  10%,  $V_{CC2}$  at 3.3 V  $\pm$  10% (over recommended operating conditions unless otherwise noted.)

	PARAMET	ER	TEST CON	IDITIONS	MIN	TYP	MAX	UNIT
	V supply surrent	Quiescent	V V 050V 00	lood		0.5	1	m Λ
I <sub>CC1</sub>	V <sub>CC1</sub> supply current	25 Mbps	$V_I = V_{CC}$ or 0 V, no	loau		2	4	mA
		ISO722/722M Sleep mode	$V_I = V_{CC}$ or 0 V,	EN at V <sub>CC</sub>			150	μΑ
I <sub>CC2</sub>	V <sub>CC2</sub> supply current	Quiescent	No load	EN at 0 V or ISO721/721M		4	6.5	mA
		25 Mbps	$V_I = V_{CC}$ or 0 V, no	load		5	7.5	
V	V High level extent valtage		I <sub>OH</sub> = -4 mA, See Figure 10		V <sub>CC</sub> – 0.4	3		V
V <sub>OH</sub>	High-level output volta	ge	$I_{OH} = -20 \mu A$ , See	Figure 10	V <sub>CC</sub> – 0.1	3.3		<b>v</b>
\/	Low lovel output volto		I <sub>OL</sub> = 4 mA, See Figure 10			0.2	0.4	V
V <sub>OL</sub>	Low-level output voltaç	ge	I <sub>OL</sub> = 20 μA, See Figure 10			0	0.1	V
V <sub>I(HYS)</sub>	Input voltage hysteres	is				150		mV
I <sub>IH</sub>	High-level input curren	t	EN, IN at 2 V				10	μΑ
I <sub>IL</sub>	Low-level input current	t	EN, IN at 0.8 V	EN, IN at 0.8 V				μΑ
l <sub>OZ</sub>	High-impedance output current	ISO722, ISO722M	EN, IN at V <sub>CC</sub>				1	μΑ
C <sub>I</sub>	Input capacitance to g	round	IN at $V_{CC}$ , $V_I = 0.4 \sin (4 \times 10^6 \pi t)$			1		pF
CMTI	Common-mode transie	ent immunity	$V_I = V_{CC}$ or 0 V, Se	e Figure 14	25	40		kV/μs

# 7.7 Electrical Characteristics, 3.3 V, 5 V

 $V_{\text{CC1}}$  at 3.3 V  $\pm$  10%,  $V_{\text{CC2}}$  at 5 V  $\pm$  10% (over recommended operating conditions unless otherwise noted.)

	PARAMET	ER	TEST CONDIT	TIONS	MIN	TYP	MAX	UNIT
	\/ aumphy aumment	Quiescent	\/ \/ or 0\/ no los			0.3	0.5	A
I <sub>CC1</sub>	V <sub>CC1</sub> supply current	25 Mbps	$V_I = V_{CC}$ or 0 V, no loa	ıu		1	2	mA
		ISO722/722M Sleep mode	V = V or 0 V	EN at V <sub>CC</sub>			200	μΑ
I <sub>CC2</sub>	V <sub>CC2</sub> supply current	Quiescent	110 1000	EN at 0 V or ISO721/721 M		8	12	mA
		25 Mbps	V <sub>I</sub> = V <sub>CC</sub> or 0 V, No load			10	14	
\/	V IPak lavel extent valle ve		I <sub>OH</sub> = -4 mA, See Figure 10		V <sub>CC</sub> - 0.8	4.6		V
V <sub>OH</sub>	High-level output voltage	ge	$I_{OH} = -20 \mu A$ , See Figure	ure 10	V <sub>CC</sub> - 0.1	5		· ·
\/	Low lovel output voltee	••	I <sub>OL</sub> = 4 mA, See Figure 10			0.2	0.4	V
V <sub>OL</sub>	Low-level output voltag	je	I <sub>OL</sub> = 20 μA, See Figure 10			0	0.1	
V <sub>I(HYS)</sub>	Input voltage hysteresi	S				150		mV
I <sub>IH</sub>	High-level input curren	t	EN, IN at 2 V				10	μΑ
I <sub>IL</sub>	Low-level input current		EN, IN at 0.8 V		-10			μΑ
l <sub>OZ</sub>	High-impedance output current	ISO722, ISO722M	EN, IN at V <sub>CC</sub>				1	μΑ
C <sub>I</sub>	Input capacitance to gr	ound	IN at $V_{CC}$ , $V_I = 0.4 \sin$	$(4 \times 10^6 \pi t)$		1		pF
CMTI	Common-mode transie	nt immunity	V <sub>I</sub> = V <sub>CC</sub> or 0 V, See F	igure 14	25	40		kV/μs



# 7.8 Electrical Characteristics, 3.3 V

 $V_{CC1}$  and  $V_{CC2}$  at 3.3 V  $\pm$  10% (over recommended operating conditions unless otherwise noted.)

001	PARAME	TER	TEST CONI	DITIONS	MIN	TYP	MAX	UNIT
	V gupply gurrent	Quiescent	\/ - \/ or 0 \/ no	lood		0.3	0.5	mA
I <sub>CC1</sub>	V <sub>CC1</sub> supply current	25 Mbps	$V_I = V_{CC}$ or 0 V, no	loau		1	2	ША
		ISO722/722M Sleep Mode		EN at V <sub>CC</sub>			150	μΑ
I <sub>CC2</sub>	V <sub>CC2</sub> supply current	Quiescent	$V_{I} = V_{CC}$ or 0 V, No load	EN at 0 V or ISO721/721 M		4	6.5	mA
		25 Mbps	$V_I = V_{CC}$ or 0 V, no	load		5	7.5	
V	High-level output volta	90	$I_{OH} = -4 \text{ mA}$ , See F	I <sub>OH</sub> = -4 mA, See Figure 10		3		V
V <sub>OH</sub>	High-level output volta	ge	$I_{OH} = -20 \mu A$ , See Figure 10		V <sub>CC</sub> - 0.1	3.3		
V	Low lovel output volto	70	I <sub>OL</sub> = 4 mA, See Figure 10			0.2	0.4	V
V <sub>OL</sub>	Low-level output voltaç	ge	$I_{OL} = 20 \mu A$ , See Fig	gure 10		0	0.1	V
$V_{I(HYS)}$	Input voltage hysteres	is				150		mV
I <sub>IH</sub>	High-level input curren	t	EN, IN at 2 V				10	μΑ
I <sub>IL</sub>	Low-level input current	t	EN, IN at 0.8 V		-10			μΑ
I <sub>OZ</sub>	High-impedance output current	ISO722, ISO722M	EN, IN at V <sub>CC</sub>				1	μΑ
C <sub>I</sub>	Input capacitance to g	round	IN at $V_{CC}$ , $V_I = 0.4$ s	in $(4 \times 10^6 \pi t)$		1		pF
CMTI	Common-mode transie	ent immunity	$V_I = V_{CC}$ or 0 V, See	Figure 14	25	40		kV/μs

# 7.9 Power Dissipation

over operating free-air temperature range (unless otherwise noted)

	PARAMETER		TEST CONDITIONS	ISO721 DUB 8 PINS	ISO72x D 8 PINS	UNIT
		ISO72x	VCC1 = VCC2 = 5.5 V, TJ = 150°C, CL = 15 pF, Input a 100-Mbps 50% duty-cycle square wave		159	mW
$P_{D}$	Power Dissipation	ISO72xM	VCC1 = VCC2 = 5.5 V, TJ = 150°C, CL = 15 pF, Input a 100-Mbps 50% duty-cycle square wave		195	mW
		ISO721	VCC1 = VCC2 = 5.5 V, TJ = 150°C, CL = 15 pF, Input a 100-Mbps 50% duty-cycle square wave	159		mW

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# 7.10 Switching Characteristics, 5 V

 $V_{CC1}$  and  $V_{CC2}$  at 5 V ± 10% (over recommended operating conditions unless otherwise noted.)

	PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay, low-to-high-level	output			13	17	24	ns
t <sub>PHL</sub>	Propagation delay, high-to-low-level	output	ISO72x		13	17	24	ns
t <sub>sk(p)</sub>	Pulse skew  t <sub>PHL</sub> - t <sub>PLH</sub>			EN at 0 V,		0.5	2	ns
t <sub>PLH</sub>	Propagation delay, low-to-high-level	output		See Figure 10	8	10	16	ns
t <sub>PHL</sub>	Propagation delay, high-to-low-level	output	ISO72xM		8	10	16	ns
t <sub>sk(p)</sub>	Pulse skew  t <sub>PHL</sub> - t <sub>PLH</sub>					0.5	1	ns
t <sub>sk(pp)</sub> (1)	Part-to-part skew					0	3	ns
t <sub>r</sub>	Output signal rise time			EN at 0 V,		1		
t <sub>f</sub>	Output signal fall time			See Figure 10		1		ns
t <sub>pHZ</sub>	Sleep-mode propagation delay, high-level-to-high-mpedance output			See Figure 11	6	8	15	ns
t <sub>pZH</sub>	Sleep-mode propagation delay, high-impedance-to-high-level output		ISO722	Occ riguio 11	3.5	4	8	μs
$t_{pLZ}$	Sleep-mode propagation delay, low-level-to-high-impedance output		ISO722M	O Firms 40	5.5	8	15	ns
t <sub>pZL</sub>	Sleep-mode propagation delay, high-impedance-to-low-level output			See Figure 12	4	5	8	μs
t <sub>fs</sub>	Failsafe output delay time from input	power loss		See Figure 13		3		μs
			100-Mbps I	NRZ data input, See Figure 15		2		
	Dook to nock over nottern iitter	ISO72x	100-Mbps unrestricted bit run length data input, See Figure 15			3		
t <sub>jit(PP)</sub>	Peak-to-peak eye-pattern jitter		150-Mbps NRZ data input, See Figure 15			1		ns
		ISO72xM	150-Mbps uinput, See	unrestricted bit run length data Figure 15		2		

<sup>(1)</sup>  $t_{sk(PP)}$  is the magnitude of the difference in propagation delay times between any specified terminals of two devices when both devices operate with the same supply voltages, at the same temperature, and have identical packages and test circuits.

Product Folder Links: ISO721 ISO721M ISO722 ISO722M



# 7.11 Switching Characteristics, 5 V, 3.3 V

 $V_{CC1}$  at 5 V  $\pm$  10%,  $V_{CC2}$  at 3.3 V  $\pm$  10% (over recommended operating conditions unless otherwise noted.)

	PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay, low-to-high-level	output			15	19	30	ns
t <sub>PHL</sub>	Propagation delay , high-to-low-level	output	ISO72x		15	19	30	ns
t <sub>sk(p)</sub>	Pulse skew  t <sub>PHL</sub> - t <sub>PLH</sub>			EN at 0 V,		0.5	3	ns
t <sub>PLH</sub>	Propagation delay, low-to-high-level	output		See Figure 10	10	12	20	ns
t <sub>PHL</sub>	Propagation delay, high-to-low-level	output	ISO72xM		10	12	20	ns
t <sub>sk(p)</sub>	Pulse skew  t <sub>PHL</sub> - t <sub>PLH</sub>					0.5	1	ns
t <sub>sk(pp)</sub> (1)	Part-to-part skew					0	5	ns
t <sub>r</sub>	Output signal rise time			EN at 0 V,		2		ns
t <sub>f</sub>	Output signal fall time			See Figure 10		2		ns
$t_{pHZ}$	Sleep-mode propagation delay, high-level-to-high-mpedance output			See Figure 11	7	11	25	ns
t <sub>pZH</sub>	Sleep-mode propagation delay, high-impedance-to-high-level output		ISO722	occ riguio 11	4.5	6	8	μs
t <sub>pLZ</sub>	Sleep-mode propagation delay, low-level-to-high-impedance output		ISO722M	O 5' 40	7	13	25	ns
t <sub>pZL</sub>	Sleep-mode propagation delay, high-impedance-to-low-level output			See Figure 12	4.5	6	8	μs
t <sub>fs</sub>	Failsafe output delay time from input	power loss		See Figure 13		3		μs
			100-Mbps I	NRZ data input, See Figure 15		2		1
	Deals to most, our mothers into	ISO72x	100-Mbps unrestricted bit run length data input, See Figure 15			3		۱ ـــ
t <sub>jit(PP)</sub>	Peak-to-peak eye-pattern jitter		150-Mbps NRZ data input, See Figure 15			1		ns
		ISO72xM	150-Mbps input, See	unrestricted bit run length data Figure 15		2		ĺ

<sup>(1)</sup>  $t_{sk(PP)}$  is the magnitude of the difference in propagation delay times between any specified terminals of two devices when both devices operate with the same supply voltages, at the same temperature, and have identical packages and test circuits.



# 7.12 Switching Characteristics, 3.3 V, 5 V

 $V_{CC1}$  at 3.3 V  $\pm$  10%,  $V_{CC2}$  at 5 V  $\pm$  10% (over recommended operating conditions unless otherwise noted.)

	PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay, low-to-high-level ou	tput			15	17	30	ns
t <sub>PHL</sub>	Propagation delay , high-to-low-level or	utput	ISO72x		15	17	30	ns
t <sub>sk(p)</sub>	Pulse skew  t <sub>PHL</sub> - t <sub>PLH</sub>			EN at 0 V,		0.5	2	ns
t <sub>PLH</sub>	Propagation delay, low-to-high-level ou	ıtput		See Figure 10	10	12	21	ns
t <sub>PHL</sub>	Propagation delay, high-to-low-level ou	ıtput	ISO72xM		10	12	21	ns
t <sub>sk(p)</sub>	Pulse skew  t <sub>PHL</sub> - t <sub>PLH</sub>					0.5	1	ns
$t_{sk(pp)}^{(1)}$	Part-to-part skew					0	5	ns
t <sub>r</sub>	Output signal rise time			EN at 0 V,		1		ns
t <sub>f</sub>	Output signal fall time			See Figure 10		1		ns
t <sub>pHZ</sub>	Sleep-mode propagation delay, high-level-to-high-mpedance output			See Figure 11	7	9	15	ns
t <sub>pZH</sub>	Sleep-mode propagation delay, high-impedance-to-high-level output		ISO722		4.5	5	8	μs
t <sub>pLZ</sub>	Sleep-mode propagation delay, low-level-to-high-impedance output		ISO722M	O	7	9	15	ns
t <sub>pZL</sub>	Sleep-mode propagation delay, high-impedance-to-low-level output			See Figure 12	4.5	5	8	μs
t <sub>fs</sub>	Failsafe output delay time from input po	ower loss		See Figure 13		3		μs
			100-Mbps I	NRZ data input, See Figure 15		2		
	Back to made our matters "then	ISO72x	100-Mbps unrestricted bit run length data input, See Figure 15			3		l
t <sub>jit(PP)</sub>	Peak-to-peak eye-pattern jitter		150-Mbps NRZ data input, See Figure 15			1		ns
		ISO72xM	150-Mbps input, See	unrestricted bit run length data Figure 15		2		<u> </u>

<sup>(1)</sup>  $t_{sk(PP)}$  is the magnitude of the difference in propagation delay times between any specified terminals of two devices when both devices operate with the same supply voltages, at the same temperature, and have identical packages and test circuits.

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# 7.13 Switching Characteristics, 3.3 V

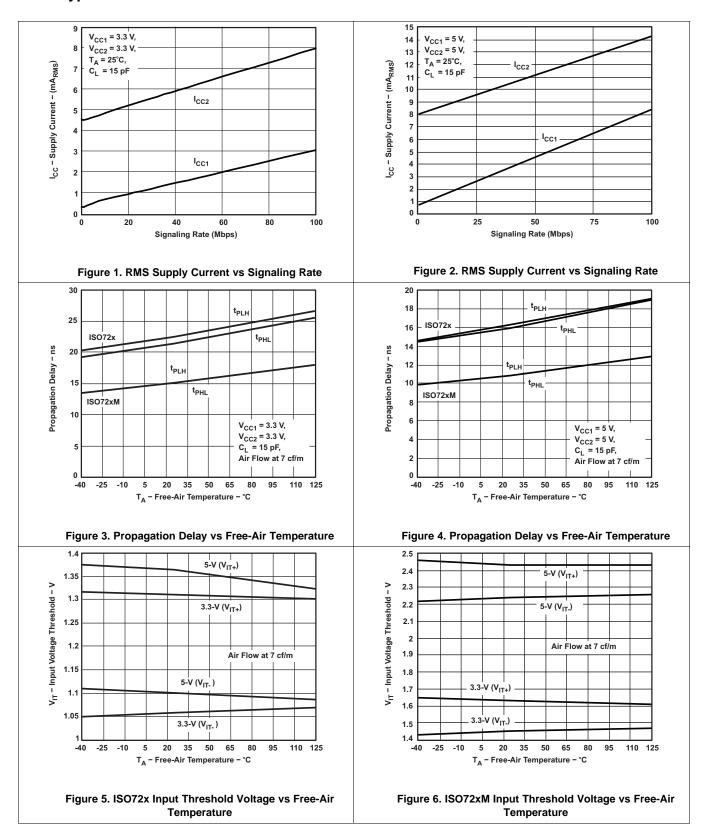
 $V_{CC1}$  and  $V_{CC2}$  at 3.3 V  $\pm$  10% (over recommended operating conditions unless otherwise noted.)

	PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	Propagation delay, low-to-high-level or	ıtput			17	20	34	ns
t <sub>PHL</sub>	Propagation delay , high-to-low-level o	utput	ISO72x		17	20	34	ns
t <sub>sk(p)</sub>	Pulse skew  t <sub>PHL</sub> - t <sub>PLH</sub>			EN at 0 V,		0.5	3	ns
t <sub>PLH</sub>	Propagation delay, low-to-high-level ou	utput		See Figure 10	10	12	25	ns
t <sub>PHL</sub>	Propagation delay, high-to-low-level ou	utput	ISO72xM		10	12	25	ns
t <sub>sk(p)</sub>	Pulse skew  t <sub>PHL</sub> - t <sub>PLH</sub>					0.5	1	ns
t <sub>sk(pp)</sub> <sup>(1)</sup>	Part-to-part skew					0	5	ns
t <sub>r</sub>	Output signal rise time			EN at 0 V,		2		20
t <sub>f</sub>	Output signal fall time			See Figure 10		2		ns
t <sub>pHZ</sub>	Sleep-mode propagation delay, high-level-to-high-mpedance output			See Figure 11	7	13	25	ns
t <sub>pZH</sub>	Sleep-mode propagation delay, high-impedance-to-high-level output		ISO722	occ riguic ri	5	6	8	μs
$t_{pLZ}$	Sleep-mode propagation delay, low-level-to-high-impedance output		ISO722M	2 - Firm 40	7	13	25	ns
t <sub>pZL</sub>	Sleep-mode propagation delay, high-impedance-to-low-level output			See Figure 12	5	6	8	μs
t <sub>fs</sub>	Failsafe output delay time from input p	ower loss		See Figure 13		3		μs
			100-Mbps I	NRZ data input, See Figure 15		2		
	Dools to mook our nettons "then	ISO72x	100-Mbps unrestricted bit run length data input, See Figure 15			3		
t <sub>jit(PP)</sub>	Peak-to-peak eye-pattern jitter		150-Mbps I	NRZ data input, See Figure 15		1		ns
		ISO72xM	150-Mbps input, See	unrestricted bit run length data Figure 15		2		

<sup>(1)</sup>  $t_{sk(PP)}$  is the magnitude of the difference in propagation delay times between any specified terminals of two devices when both devices operate with the same supply voltages, at the same temperature, and have identical packages and test circuits.

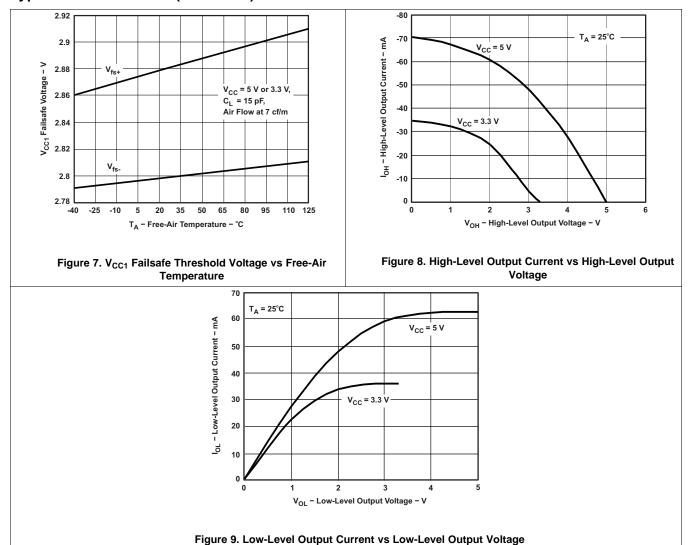
# TEXAS INSTRUMENTS

# 7.14 Typical Characteristics





# **Typical Characteristics (continued)**



## 8 Parameter Measurement Information

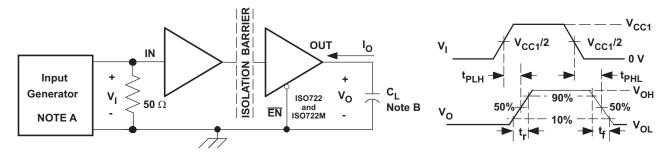


Figure 10. Switching Characteristic Test Circuit and Voltage Waveforms

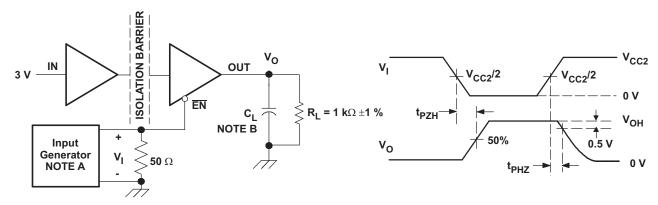


Figure 11. ISO722 Sleep-Mode High-Level Output Test Circuit and Voltage Waveforms

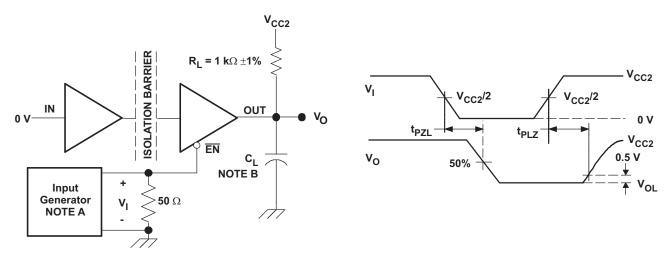


Figure 12. ISO722 Sleep-Mode Low-Level Output Test Circuit and Voltage Waveforms

# **NOTE**

A: The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  50 kHz, 50% duty cycle,  $t_r \leq$  3 ns,  $t_f \leq$  3 ns,  $Z_O =$  50  $\Omega$ .

B:  $C_L = 15 \text{ pF} \pm 20\%$  and includes instrumentation and fixture capacitance.



# **Parameter Measurement Information (continued)**

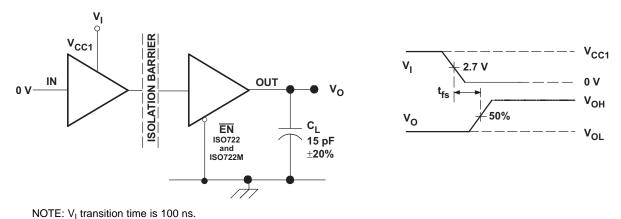
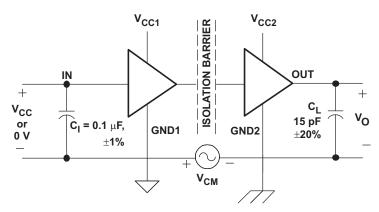


Figure 13. Failsafe Delay Time Test Circuit and Voltage Waveforms

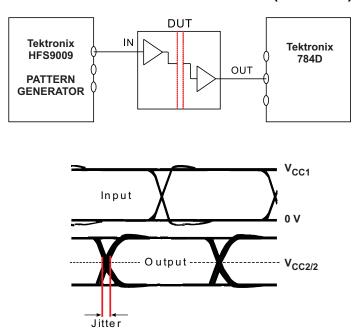


NOTE: Pass/fail criterion is no change in Vo.

Figure 14. Common-Mode Transient-Immunity Test Circuit and Voltage Waveform



# **Parameter Measurement Information (continued)**



NOTE: Bit pattern run length is  $2^{16} - 1$ . Transition time is 800 ps. NRZ data input has no more than five consecutive 1s or 0s.

Figure 15. Peak-to-Peak Eye-Pattern Jitter Test Circuit and Voltage Waveform



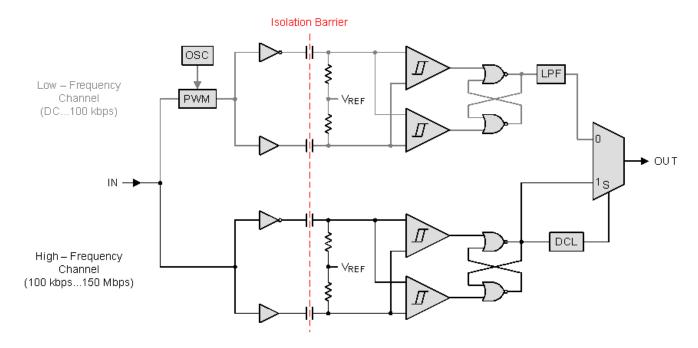
# 9 Detailed Description

#### 9.1 Overview

The isolator in the *Functional Block Diagram* is based on a capacitive isolation barrier technique. The I/O channel of the device consists of two internal data channels, a high-frequency channel (HF) with a bandwidth from 100 kbps up to 150 Mbps, and a low-frequency channel (LF) covering the range from 100 kbps down to DC. In principle, a single ended input signal entering the HF-channel is split into a differential signal via the inverter gate at the input. The following capacitor-resistor networks differentiate the signal into transients, which then are converted into differential pulses by two comparators. The comparator outputs drive a NOR-gate flip-flop whose output feeds an output multiplexer. A decision logic (DCL) at the driving output of the flip-flop measures the durations between signal transients. If the duration between two consecutive transients exceeds a certain time limit, (as in the case of a low-frequency signal), the DCL forces the output-multiplexer to switch from the high-frequency to the low-frequency channel.

Because low-frequency input signals require the internal capacitors to assume prohibitively large values, these signals are pulse-width modulated (PWM) with the carrier frequency of an internal oscillator, creating a sufficiently high-frequency signal capable of passing the capacitive barrier. As the input is modulated, a low-pass filter (LPF) is needed to remove the high-frequency carrier from the actual data before passing the carrier on to the output multiplexer.

## 9.2 Functional Block Diagram





## 9.3 Features Description

Insulation characteristics and regulatory information of ISO72x family is provided in this section.

#### 9.3.1 Insulation Characteristics

over recommended operating conditions (unless otherwise noted.)

	PARAMETER	TEST CONDITIONS	SPECIFICATIONS	UNIT
DIN V V	DE V 0884-10 (VDE V 0884-10):2006-12 <sup>(</sup>	1)		
V <sub>IORM</sub>	Maximum working insulation voltage		560	Vpeak
		After Input/Output Safety Test Subgroup 2/3 $V_{PR} = V_{IORM} \times 1.2$ , $t = 10$ s, Partial discharge < 5 pC	672	Vpeak
$V_{PR}$	Input to output test voltage	Method a, $V_{PR} = V_{IORM} \times 1.6$ , Type and sample test with t = 10 s, Partial discharge < 5 pC	896	Vpeak
		Method b1, $V_{PR} = V_{IORM} \times 1.875$ , 100% production test with t = 1 s, Partial discharge < 5 pC	1050	Vpeak
V <sub>IOTM</sub>	Transient overvoltage	t = 60 s	4000	Vpeak
Rs	Insulation resistance	V <sub>IO</sub> = 500 V at T <sub>S</sub>	> 10 <sup>9</sup>	Ω
	Pollution degree		2	
UL 1577			<u>'</u>	
.,	Indiation values	V <sub>TEST</sub> = V <sub>ISO</sub> , t = 60 s (qualification)	3535 / 2500	\/m = =   - \/ /m = = -
V <sub>ISO</sub>	Isolation voltage	$V_{TEST} = 1.2 \times V_{ISO}, t = 1 \text{ s } (100\% \text{ production})^{(2)}$	4242 / 3000	Vpeak/Vrms

<sup>(1)</sup> Climatic classification 40/125/21

# 9.3.2 IEC 60664-1 Ratings Table

PARAMETER	TEST CONDITIONS	SPECIFICATION
Basic isolation group	Material group	II
Installation classification	Rated mains voltage ≤150 VRMS	I-IV
IIIStaliation ciassincation	Rated mains voltage ≤300 VRMS	1-111

## 9.3.3 Regulatory Information

VDE	CSA	UL
Certified according to DIN V VDE V 0884-10 (VDE V 0884-10):2006-12 and DIN EN 61010-1 (VDE 0411-1)	Approved according to CSA Component Acceptance Notice 5A and IEC 60950-1	Recognized under UL 1577 Component Recognition Program
Basic Insulation Maximum Transient Overvoltage, 4000 V <sub>PK</sub> Maximum Working Voltage, 560 V <sub>PK</sub> Maximum Surge Voltage, 4000 V <sub>PK</sub>	Evaluated to CSA 60950-1-07 and IEC 60950-1 (2nd Ed) with 2000 $V_{RMS}$ Isolation rating for products with working voltages $\leq$ 125 $V_{RMS}$ for reinforced insulation and $\leq$ 390 $V_{RMS}$ for basic insulation	Single Protection, 2500 V <sub>RMS</sub> <sup>(1)</sup>
Certificate number: 40016131	Master contract number: 220991	File number: E181974

(1) Production tested  $\geq$  3000 V<sub>RMS</sub> for 1 second in accordance with UL 1577.

<sup>(2)</sup> Based on lifetime curve (see the *High-Voltage Lifetime of the ISO72x Family of Digital Isolators* application report, SLLA197); these devices can withstand 4242 Vpeak / 3000 Vrms for > 10,000 s at 150°C.



#### 9.3.4 Package Insulation Characteristics

	PARAMETER	DESCRIPTIONS / TEST CONDITIONS	3	MIN	TYP	MAX	UNIT
1 (404)	Minimum air gan (alaaranaa) (1)	Chartest terminal to terminal distance through air	D-8				
L(101)	Minimum air gap (clearance) (1)	Shortest terminal-to-terminal distance through air	DUB-8	6.1			mm
L (102)	Minimum external tracking	Shortest terminal-to-terminal distance across the	D-8	4			<b>~~</b>
L(102)	(creepage) <sup>(1)</sup>	package surface	DUB-8	6.8			mm
СТІ	Tracking resistance (comparative tracking index)	DIN EN 60112 (VDE 0303-11); IEC 60112	400			V	
DTI	Distance through insulation	Minimum internal gap (internal clearance)		0.008			mm
В	Isolation resistance	Input to output, $V_{IO}$ = 500 V; all pins on each side barrier tied together, creating a two-terminal devices		10 <sup>12</sup>			Ω
R <sub>IO</sub>	Isolation resistance	Input to output, $V_{IO} = 500 \text{ V}$ , $100^{\circ}\text{C} \le T_A < T_A \text{ max}$ .		10 <sup>11</sup>			Ω
C <sub>IO</sub>	Barrier capacitance Input-to-output	$V_I = 0.4 \sin (4 \times 10^6 \pi t)$		1		pF	

<sup>(1)</sup> Creepage and clearance requirements are applied according to the specific equipment isolation standards of an application. Maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed circuit board do not reduce this distance.

#### 9.3.5 Safety Limiting Values

Safety limiting intends to prevent potential damage to the isolation barrier upon failure of input or output circuitry. A failure of the I/O can allow low resistance to ground or the supply, and without current limiting, dissipate sufficient power to overheat the die and damage the isolation barrier, potentially leading to secondary system failures.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Cofety input output or cumply current	$\theta_{JA} = 263$ °C/W, $V_I = 5.5$ V, $T_J = 170$ °C, $T_A = 25$ °C			100	A
IS	Safety input, output, or supply current	$\theta_{JA} = 263^{\circ}\text{C/W}, \ V_{I} = 3.6 \ \text{V}, \ T_{J} = 170^{\circ}\text{C}, \ T_{A} = 25^{\circ}\text{C}$			153	mA
T <sub>S</sub>	Maximum case temperature				150	°C

The safety-limiting constraint is the absolute maximum junction temperature specified in the absolute maximum ratings table. The power dissipation and junction-to-air thermal impedance of the device installed in the application hardware determines the junction temperature. The junction-to-air thermal resistance in the *Thermal Information* table is that of a device installed in the JESD51-3, Low Effective Thermal Conductivity Test Board for Leaded Surface Mount Packages and is conservative.

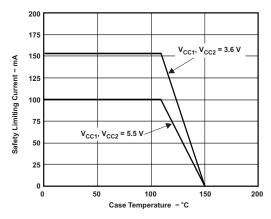


Figure 16.  $\theta_{JC}$  Thermal Derating Curve per VDE

Creepage and clearance on a printed circuit board become equal according to the measurement techniques shown in the *Isolation Glossary* in the *Related Documentation* section. Techniques such as inserting grooves and/or ribs on a printed circuit board are used to help increase these specifications.



#### 9.4 Device Functional Modes

Functional modes of ISO72x are shown in Table 1 and Table 2.

**Table 1. ISO721 Functional Table** 

V <sub>CC1</sub>	V <sub>CC2</sub>	INPUT (IN)	OUTPUT (OUT)
		Н	Н
PU	PU	L	L
		Open	Н
PD	PU	X	Н
X	PD	X	Undetermined

Table 2. ISO722 Functional Table

V <sub>CC1</sub>	V <sub>CC2</sub>	INPUT (IN)	OUTPUT ENABLE (EN)	OUTPUT (OUT)
		Н	L or open	Н
PU	PU	L	L or open	L
PU		X	Н	Z
		Open	L or open	Н
PD	PU	X	L or open	Н
PD	PU	X	Н	Z
X	PD	X	X	Undetermined

#### 9.4.1 Device I/O Schematic

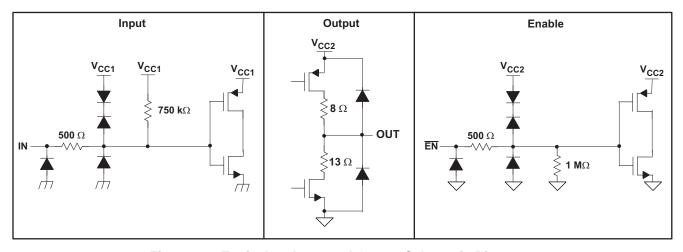


Figure 17. Equivalent Input and Output Schematic Diagrams



# 10 Application and Implementation

#### NOTE

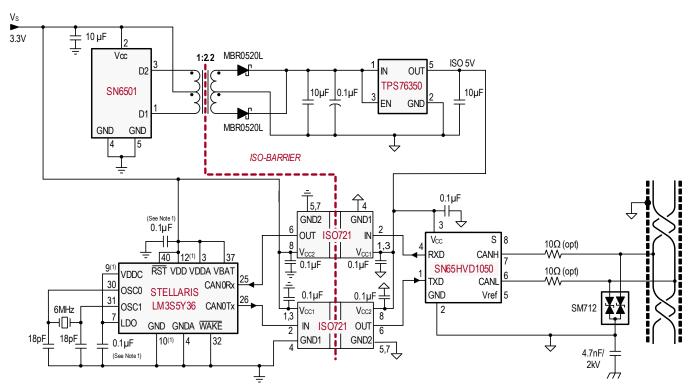
Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

# 10.1 Application Information

The ISO72x devices use single-ended TTL or CMOS-logic-switching technology. The supply voltage range of the devices is from 3 V to 5.5 V for both supplies,  $V_{CC1}$  and  $V_{CC2}$ . When designing with digital isolators, due to the single-ended design structure, digital isolators do not conform to any specific interface standard and are only intended for isolating single-ended CMOS or TTL digital signal lines. The isolator is typically placed between the data controller ( $\mu$ C or UART), and a data converter or a line transceiver, regardless of the interface type or standard.

#### 10.2 Typical Application

The ISO721 device can be used with Texas Instruments' microcontroller, CAN transceiver, transformer driver, and low-dropout voltage regulator to create an Isolated CAN Interface as shown in Figure 18.



(1) Multiple pins and capacitors omitted for clarity purpose.

Figure 18. Isolated CAN Interface



### Typical Application (continued)

#### 10.2.1 Design Requirements

Unlike optocouplers which need external components to improve performance and provide bias (or limit current), the ISO72x devices need only two external bypass capacitors to operate.

#### 10.2.2 Detailed Design Procedure

Typical ISO721 circuit hook-up is shown in Figure 19.

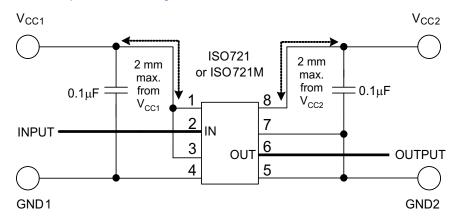


Figure 19. Typical ISO721 Circuit Hook-up

The ISO72x isolators have the same functional pinout as those of most other vendors as shown in Figure 20, and they are often pin-for-pin drop-in replacements. The notable differences in the products are propagation delay, signaling rate, power consumption, and transient protection rating. Table 3 is used as a guide for replacing other isolators with the ISO72x family of single channel isolators.

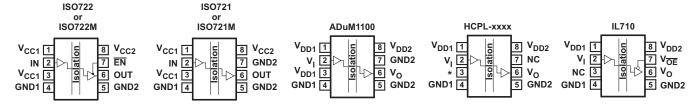


Figure 20. Pin Cross Reference

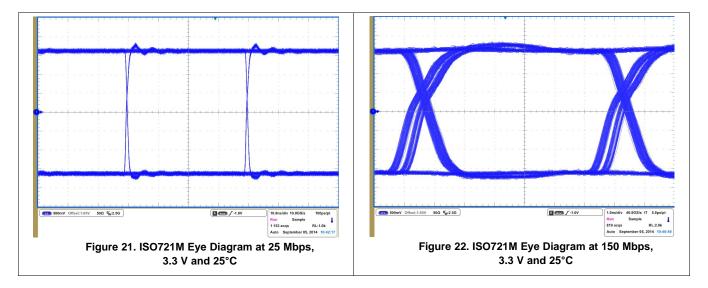
**Table 3. Cross Reference** 

							PII	N 7	
ISOLATOR	PIN 1	PIN 2	PIN 3	PIN 4	PIN 5	PIN 6	ISO721 OR ISO721M	ISO722 OR ISO722M	PIN 8
ISO721 (1)(2)	V <sub>CC1</sub>	IN	V <sub>CC1</sub>	GND1	GND2	OUT	GND2	ĒΝ	V <sub>CC2</sub>
ADuM1100 <sup>(1)(2)</sup>	V <sub>DD1</sub>	VI	$V_{DD1}$	GND1	GND2	Vo	GN	ID2	$V_{DD2}$
HCPL-xxxx	V <sub>DD1</sub>	VI	*Leave Open <sup>(3)</sup>	GND1	GND2	Vo	NC <sup>(4)</sup>		$V_{DD2}$
IL710	V <sub>DD1</sub>	VI	NC <sup>(5)</sup>	GND1	GND2	Vo	V <del>OE</del>		$V_{DD2}$

- Pin 1 should be used as V<sub>CC1</sub>. Pin 3 can also be used as V<sub>CC1</sub> or left open, as long as pin 1 is connected to V<sub>CC1</sub>.
- Pin 5 should be used as GND2. Pin 7 can also be used as GND2 or left open, as long as pin 5 is connected to GND2.
- Pin 3 of the HCPL devices must be left open. This is not a problem when substituting an ISO72x device, because the extra V<sub>CC1</sub> on pin 3 can be left an open circuit as well.
- An HCPL device pin 7 must be left floating (open) or grounded when an ISO722 or ISO722M device is to be used as a drop-in replacement. If pin 7 of the ISO722 or ISO722M device is placed in a high logic state, the output of the device is disabled.
- Pin 3 of the IL710 must not be tied to ground on the circuit board because this shorts the ISO72x V<sub>CC1</sub> to ground. The IL710 pin 3 can only be tied to V<sub>CC</sub> or left open to drop in an ISO72x device.



#### 10.2.3 Application Curves



# 11 Power Supply Recommendations

To ensure reliable operation at all data rates and supply voltages, a  $0.1-\mu F$  bypass capacitor should be placed at input and output supply pins ( $V_{CC1}$  and  $V_{CC2}$ ). The capacitors should be placed as close to the supply pins as possible. If only a single primary-side power supply is available in an application, isolated power can be generated for the secondary-side with the help of a transformer driver such as Texas Instruments SN6501 data sheet. For such applications, detailed power supply design and transformer selection recommendations are available in the data sheet, SN6501 *Transformer Driver for Isolated Power Supplies* (SLLSEA0).

#### 12 Layout

# 12.1 Layout Guidelines

A minimum of four layers is required to accomplish a low EMI PCB design as shown in Figure 23. Layer stacking should be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane, and low-frequency signal layer.

- Routing the high-speed traces on the top layer avoids the use of vias (and the introduction of their inductances) and allows for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Placing a solid ground plane next to the high-speed signal layer establishes controlled impedance for transmission line interconnects and provides an excellent low-inductance path for the return current flow.
- Placing the power plane next to the ground plane creates additional high-frequency bypass capacitance of approximately 100 pF/in<sup>2</sup>.
- Routing the slower speed control signals on the bottom layer allows for greater flexibility as these signal links usually have margin to tolerate discontinuities such as vias.

If an additional supply voltage plane or signal layer is needed, add a second power/ground plane system to the stack to keep it symmetrical. Adding a second plane system makes the stack mechanically stable and prevents it from warping. Also the power and ground plane of each power system can be placed closer together, thus increasing the high-frequency bypass capacitance significantly. For detailed layout recommendations, see the Application Note *Digital Isolator Design Guide* (SLLA284).



# **Layout Guidelines (continued)**

#### 12.1.1 PCB Material

For digital circuit boards operating below 150 Mbps, (or rise and fall times higher than 1 ns), and trace lengths of up to 10 inches, use standard FR-4 epoxy-glass as PCB material. FR-4 (Flame Retardant 4) meets the requirements of Underwriters Laboratories UL94-V0, and is preferred over cheaper alternatives due to its lower dielectric losses at high frequencies, less moisture absorption, greater strength and stiffness, and its self-extinguishing flammability-characteristics.

# 12.2 Layout Example

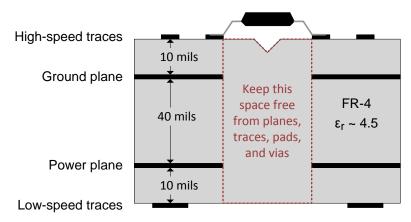


Figure 23. Recommended Layer Stack



# 13 Device and Documentation Support

# 13.1 Documentation Support

#### 13.1.1 Related Documentation

Transformer Driver for Isolated Power Supplies, SLLSEA0.

Digital Isolator Design Guide, SLLA284.

Isolation Glossary, SLLA353

## 13.2 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

**Table 4. Related Links** 

PARTS	PRODUCT FOLDER	SAMPLE & BUY	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
ISO721	Click here	Click here	Click here	Click here	Click here
ISO721M	Click here	Click here	Click here	Click here	Click here
ISO722	Click here	Click here	Click here	Click here	Click here
ISO722M	Click here	Click here	Click here	Click here	Click here

#### 13.3 Trademarks

SDS is a trademark of Honeywell.

DeviceNet is a trademark of Open Devicenet Vendors Association, Inc.

All other trademarks are the property of their respective owners.

# 13.4 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

# 13.5 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

# 14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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29-Jul-2018

# **PACKAGING INFORMATION**

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish (6)	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
ISO721D	ACTIVE	SOIC	Drawing	8	75	Green (RoHS	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	ISO721	
1307210	ACTIVE	3010	Б	0	75	& no Sb/Br)	CO NII DAO	Level-1-2000-OINLIIVI	-40 to 125	150721	Sample
ISO721DG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	ISO721	Sample
ISO721DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	ISO721	Sample
ISO721DRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	ISO721	Sample
ISO721DUB	OBSOLETE	SOP	DUB	8		TBD	Call TI	Call TI	-40 to 125	ISO721	
ISO721DUBR	LIFEBUY	SOP	DUB	8		TBD	Call TI	Call TI	-40 to 125	ISO721	
ISO721MD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	IS721M	Sample
ISO721MDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	IS721M	Sample
ISO721MDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	IS721M	Sample
ISO721MDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	IS721M	Sample
ISO722D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	ISO722	Sample
ISO722DG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	ISO722	Sample
ISO722DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	ISO722	Sample
ISO722DRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	ISO722	Sample
ISO722MD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	IS722M	Sample
ISO722MDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	IS722M	Sample
ISO722MDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	IS722M	Sample

<sup>(1)</sup> The marketing status values are defined as follows:

# PACKAGE OPTION ADDENDUM



29-.lul-2018

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead/Ball Finish Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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#### OTHER QUALIFIED VERSIONS OF ISO721, ISO721M, ISO722:

Automotive: ISO721-Q1, ISO721-Q1, ISO722-Q1

■ Enhanced Product: ISO721M-EP

Military: ISO721M

NOTE: Qualified Version Definitions:



# **PACKAGE OPTION ADDENDUM**

29-Jul-2018

- Automotive Q100 devices qualified for high-reliability automotive applications targeting zero defects
- Enhanced Product Supports Defense, Aerospace and Medical Applications
- Military QML certified for Military and Defense Applications

# PACKAGE MATERIALS INFORMATION

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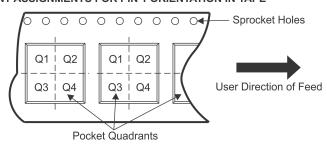
# TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

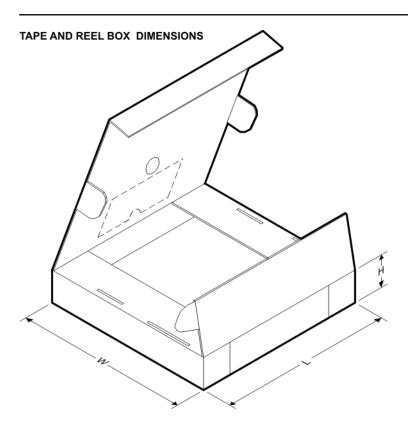
# QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

All dimensions are nomina												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISO721DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
ISO721MDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
ISO722DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
ISO722MDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

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\*All dimensions are nominal

7 till dillitorioriorio di o mominidi							
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISO721DR	SOIC	D	8	2500	367.0	367.0	38.0
ISO721MDR	SOIC	D	8	2500	367.0	367.0	38.0
ISO722DR	SOIC	D	8	2500	367.0	367.0	38.0
ISO722MDR	SOIC	D	8	2500	367.0	367.0	38.0



Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.

4207614/E



# D (R-PDSO-G8)

# PLASTIC SMALL OUTLINE



NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AA.



# D (R-PDSO-G8)

# PLASTIC SMALL OUTLINE



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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