

1 FEATURES

- Fully compatible with the “ISO 11898” standard
- High speed (up to 1 M/Baud)
- At least 110 nodes can be connected
- Improved Replacement for the TJA1050
- ± 12 kV ESD Protection
- Bus-Fault Protection of -50V to 50V
- Thermally protected
- Transmit Data (TXD) dominant time-out function
- Silent mode in which the transmitter is disabled

2 APPLICATIONS

- Industrial Automation
- High-Speed CAN Bus for Automotive Applications
- SAE J1939 Standard Data Bus Interface
- ISO 11783 Standard Data Bus Interface
- NMEA 2000 Standard Data Bus Interface

3 DESCRIPTION

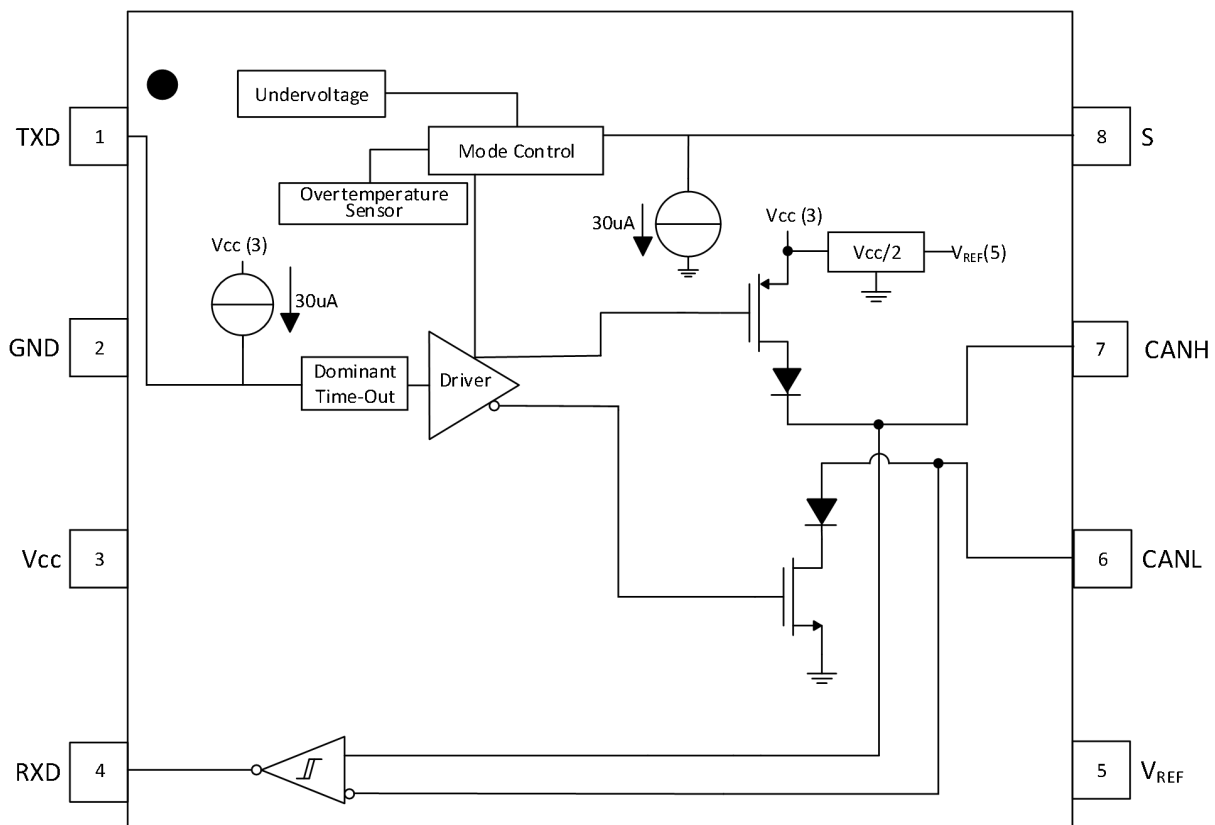
The HMT1050 is the interface between the Controller Area Network (CAN) protocol controller and the physical bus.

The device provides differential transmit capability to the bus and differential receive capability to the CAN controller.

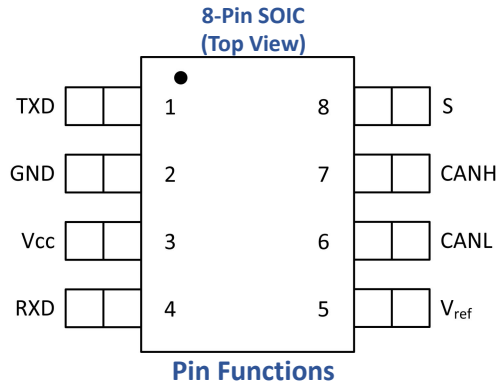
As a CAN transceiver, this device provides differential transmit capability to the bus and differential receive capability to a CAN controller at signaling rates up to 1 megabit per second (M/bps).

The HMT1050 is designed for operation in especially harsh environments. As a result, the device features cross-wire, overvoltage and loss of ground protection from -27 V to 40 V, overtemperature shutdown, a -12 V to 12 V common-mode range, and will withstand voltage transients from -200 V to 200 V according to ISO7637.

HMT1050 Block Diagram



4 Pin Configuration and Functions



| Pin | | I/O | Description |
|------------------|-----|--------|--|
| Name | No. | | |
| TXD | 1 | I | CAN transmit data input (LOW for dominant and HIGH for recessive bus states) |
| GND | 2 | GND | Device ground |
| V _{CC} | 3 | Supply | Transceiver 5-V supply |
| RXD | 4 | O | CAN receive data output (LOW for dominant and HIGH for recessive bus states) |
| V _{ref} | 5 | O | Reference output voltage |
| CANL | 6 | I/O | Low level CAN bus line |
| CANH | 7 | I/O | High level CAN bus line |
| S | 8 | I | Mode-select pin (Logic LOW places the device in high-speed mode and logic HIGH places the device in a listen-only silent mode) |

5 Specifications

5.1 Absolute Maximum Ratings

See Note⁽¹⁾

| Parameter | Description | Min | Max | Unit |
|------------------|---|------|-----|------|
| V _{CC} | Supply Voltage | -0.3 | 7 | V |
| | Voltage at any bus terminal (CANH, CANL, V _{ref}) | -50 | 50 | V |
| V _I | Voltage input, transient pulse ⁽²⁾ (CANH, CANL) | -200 | 200 | V |
| V _I | Voltage input range (TXD, S) | -0.5 | 6 | V |
| T _J | Junction temperature | -55 | 170 | °C |
| T _{stg} | Storage temperature | -40 | 125 | °C |

(1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) Tested in accordance with ISO 7637, test pulses 1, 2, 3a, 3b, 5, 6, and 7.

5.2 ESD Ratings

| | | | Value | Unit |
|---|--|----------------------|--------|------|
| V _(ESD) Electrostatic discharge | Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾ | All pins | ±8000 | V |
| | | Bus terminals vs GND | ±12000 | |
| | Charged-device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾ | | ±2000 | |
| | Machine model (MM) ANSI/ESDS5.2-1996 | | ±400 | |

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

5.3 Recommended Operating Condition

| Parameter | Description | | Min | Max | Unit |
|-----------------------------------|---|---|-----|-----------------|------|
| V _{CC} | Supply voltage | | 4.5 | 5.5 | V |
| V _I or V _{IC} | Voltage at any bus terminal (separately or common mode) | | -24 | 24 | V |
| V _{IH} | High-level input voltage | TXD, S | 2.0 | V _{CC} | V |
| V _{IL} | Low-level input voltage | | 0 | 0.8 | V |
| V _{ID} | Differential input voltage | | -7 | 7 | V |
| I _{OH} | High-level output current | Driver | -70 | | mA |
| | | Receiver | -6 | | |
| I _{OL} | Low-level output current | Driver | | 70 | mA |
| | | Receiver | | 6 | |
| T _J | Junction temperature | See Absolute Maximum Ratings ⁽¹⁾ , 1-Mbps minimum signaling rate with R _L = 54Ω | -40 | 150 | °C |
| | Signaling Rate | | 20 | | kbps |

(1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability

5.4 Driver Electrical Characteristics

over recommended operating conditions (unless otherwise noted)

| Parameter | Description | | Test Conditions | | Min | TYP ⁽¹⁾ | Max | Unit |
|--------------------|---|---|--|---------------------------------|------|--------------------|-----|------|
| V _{O(D)} | Bus output voltage (Dominant) | CANH | V _I = 0V, S at 0V, R _L = 60 Ω, See Figure 1 and Figure 2 | 4.75V < V _{CC} < 5.25V | 3.0 | 3.6 | 4.5 | V |
| | | | | 4.5V < V _{CC} < 5.5V | 2.75 | | 5.2 | |
| | | CANL | V _I = 0V, S at 0V, R _L = 60 Ω, See Figure 1 and Figure 2 | 4.75V < V _{CC} < 5.25V | 0.8 | 1.4 | 1.6 | |
| | | | | 4.5V < V _{CC} < 5.5V | | | 1.6 | |
| V _{O(R)} | Bus output voltage (Recessive) | V _I = 3V, S at 0V, R _L = 60 Ω, See Figure 1 and Figure 2 | 4.75V < V _{CC} < 5.25V | 2 | 2.5 | 3 | V | |
| | | | 4.5V < V _{CC} < 5.5V | 1.8 | 2.4 | 3 | V | |
| V _{OD(D)} | Differential output voltage (Dominant) | V _I = 0V, R _L = 60 Ω, S at 0V, See Figure 1 and Figure 2 , and Figure 3 | 4.75V < V _{CC} < 5.25V | 1.5 | 2.2 | 3 | V | |
| | | | 4.5V < V _{CC} < 5.5V | 1.4 | | 3 | | |
| | | V _I = 0V, R _L = 45 Ω, S at 0V, See Figure 1 and Figure 2 , and Figure 3 | 4.75V < V _{CC} < 5.25V | 1.4 | | 3 | | |
| | | | 4.5V < V _{CC} < 5.5V | 1.3 | | 3 | | |
| V _{OD(R)} | Differential output voltage (Recessive) | V _I = 3V, S at 0V, See Figure 1 and Figure 2 | | -0.012 | | 0.012 | V | |
| | | | V _I = 3V, S at 0V, No Load | -0.5 | | 0.05 | | |

| | | | | | | | |
|---------------------|---|--------------------------------|--------------------------|------|------|-----|---------|
| $V_{OC(SS)}$ | Steady state common-mode output voltage | S at 0V | $4.75V < V_{CC} < 5.25V$ | 2 | 2.5 | 3 | V |
| | | | $4.5V < V_{CC} < 5.5V$ | 1.9 | | 3 | |
| $\Delta V_{OC(SS)}$ | Change in steady-state common-mode output voltage | | | 30 | | | mV |
| I_{IH} | High-level input current, TXD input | V_i at V_{CC} | | -2 | 0 | 2 | μA |
| I_{IL} | Low-level input current, TXD input | V_i at 0V | | -50 | -30 | -10 | |
| $I_{O(off)}$ | Power-off TXD output current | V_{CC} at 0V, TXD at 5V | | | | 1 | mA |
| $I_{OS(SS)}$ | Short-circuit steady-state output current | $V_{CANH} = -12V$, CANL Open | | -130 | -100 | | |
| | | $V_{CANH} = 12V$, CANL Open | | | | 1 | |
| | | $V_{CANL} = -12V$, CANH Open | | -1 | | | |
| | | $V_{CANL} = 12V$, CANH Open | | | 77 | 105 | |
| C_O | Output capacitance | See receiver input capacitance | | | | | |

(1) All typical values are at 25°C with a 5-V supply.

5.5 Receiver Electrical Characteristics

over recommended operating conditions (unless otherwise noted)

| Parameter | Description | Test Conditions | | Min | TYP ⁽¹⁾ | Max | Unit |
|--------------|---|---|--------------------------|-----|--------------------|-----|------------|
| V_{IT+} | Positive-going input threshold voltage | S at 0V | | | 800 | 900 | mV |
| V_{IT-} | Negative-going input threshold voltage | | | 500 | 650 | | |
| V_{hys} | Hysteresis voltage ($V_{IT+} - V_{IT-}$) | | | 100 | 125 | | |
| V_{OH} | High-level output voltage | $I_O = -2mA$ See Figure 6 | $4.75V < V_{CC} < 5.25V$ | 4 | 4.6 | | V |
| | | | $4.5V < V_{CC} < 5.5V$ | 3.8 | | | |
| V_{OL} | Low-level output voltage | $I_O = 2mA$, See Figure 6 | | | 0.3 | 0.4 | V |
| $I_{I(off)}$ | Power-off bus input current | CANH or CANL = 5V, Other pin at 0V, V_{CC} at 0V, TXD at 0V | | | 165 | 250 | μA |
| $I_{O(off)}$ | Power-off RXD leakage current | V_{CC} at 0V, RXD at 5V | | | | 20 | μA |
| C_I | Input capacitance to ground, (CANH or CANL) | TXD at 3 V, $V_i = 0.4\sin(4E6\pi t) + 2.5$ V | | | 13 | | pF |
| C_{ID} | Differential input capacitance | TXD at 3 V, $V_i = 0.4\sin(4E6\pi t)$ | | | 5 | | |
| R_{ID} | Differential input resistance | TXD at 3V, S at 0V | | 30 | 50 | 80 | k Ω |
| R_{IN} | Input resistance, (CANH or CANL) | | | 15 | 25 | 40 | |
| $R_{I(m)}$ | Input resistance matching $[1 - R_{IN(CANH)} / R_{IN(CANL)}] \times 100\%$ | $V_{(CANH)} = V_{(CANL)}$ | | -3% | 0 | 3% | |

(1) All typical values are at 25°C with a 5V supply.

5.6 Device Switching Characteristics

over recommended operating conditions (unless otherwise noted)

| Parameter | Description | Test Conditions | | Min | TYP | Max | Unit |
|----------------|--|-----------------|--------------------------|-----|-----|-----|------|
| $T_{d(LOOP1)}$ | Total loop delay, driver input to receiver output, Recessive to Dominant | S at 0V | $4.75V < V_{CC} < 5.25V$ | 70 | 100 | 190 | ns |
| | | | $4.5V < V_{CC} < 5.5V$ | 75 | | 115 | |
| $T_{d(LOOP2)}$ | Total loop delay, driver input to receiver output, Dominant to Recessive | S at 0V | $4.75V < V_{CC} < 5.25V$ | 70 | 110 | 190 | |
| | | | $4.5V < V_{CC} < 5.5V$ | 75 | | 155 | |

5.7 Driver Switching Characteristics

over recommended operating conditions (unless otherwise noted)

| Parameter | Description | Test Conditions | Min | TYP | Max | Unit | |
|-------------|--|---------------------------------------|--------------------------|-----|-----|---------|---------|
| t_{PLH} | Propagation delay time, low-to-high-level output | S at 0V, See Figure 4 | 25 | 55 | 110 | ns | |
| t_{PHL} | Propagation delay time, high-to-low-level output | | 25 | 50 | 90 | | |
| t_r | Differential output signal rise time | | 25 | | | | |
| t_f | Differential output signal fall time | | 50 | | | | |
| t_{en} | Enable time from silent mode to dominant | | | | 10 | μ s | |
| $T_{(dom)}$ | Dominant time-out | $\downarrow V_I$ | $4.75V < V_{CC} < 5.25V$ | 300 | 550 | 700 | μ s |
| | | | $4.5V < V_{CC} < 5.5V$ | 280 | | 700 | |

5.8 Receiver Switching Characteristics

over recommended operating conditions (unless otherwise noted)

| Parameter | Description | Test Conditions | Min | TYP | Max | Unit | |
|-----------|--|--|--------------------------|-----|-----|------|----|
| T_{PLH} | Propagation delay time, low-to-high-level output | S at 0V or V_{CC} See Figure 6 | $4.75V < V_{CC} < 5.25V$ | 50 | 75 | 130 | ns |
| T_{PHL} | Propagation delay time, high-to-low-level output | | $4.5V < V_{CC} < 5.5V$ | 55 | | 135 | |
| | | | $4.75V < V_{CC} < 5.25V$ | 25 | 50 | 90 | |
| | | | $4.5V < V_{CC} < 5.5V$ | 30 | | 95 | |
| t_r | Output signal rise time | | 8 | | | | |
| t_f | Output signal fall time | | 8 | | | | |

5.9 Supply Current

over operating free-air temperature range (unless otherwise noted)

| Parameter | Description | Test Conditions | Min | TYP | Max | Unit | |
|-----------|-----------------------------------|-----------------|--|--------------------------|-----|------|----|
| I_{CC} | 5-V Supply current | Silent mode | S at V_{CC} , $V_I = V_{CC}$ | | 2.5 | 5 | mA |
| | | Dominant | $V_I = 0V$, 60 Ω Load, S at 0V | $4.75V < V_{CC} < 5.25V$ | 50 | 70 | |
| | | | | $4.5V < V_{CC} < 5.5V$ | | 75 | |
| Recessive | $V_I = V_{CC}$, No Load, S at 0V | | 2.5 | 5 | | | |

5.10 S-Pin Characteristics

over recommended operating conditions (unless otherwise noted)

| Parameter | Description | Test Conditions | Min | Typ | Max | Unit |
|-----------|--------------------------|-----------------|-----|-----|-----|---------|
| I_{IH} | High level input current | S at 2V | 20 | 35 | 50 | μ A |
| I_{IL} | Low level input current | S at 0.8V | 15 | 30 | 45 | |

5.11 V_{REF} -Pin Characteristics

over recommended operating conditions (unless otherwise noted)

| Parameter | Description | Test Conditions | Min | Typ | Max | Unit |
|-----------|--------------------------|------------------------------|-------------|-------------|-------------|------|
| V_{REF} | Reference output voltage | $-50 \mu A < I_o < 50 \mu A$ | $0.4V_{CC}$ | $0.5V_{CC}$ | $0.6V_{CC}$ | V |

6 Parameter Measurement Information

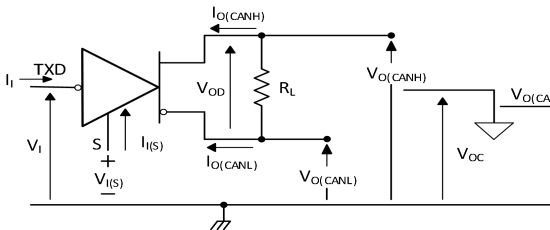


Figure 1. Driver Voltage, Current, and Test Definition

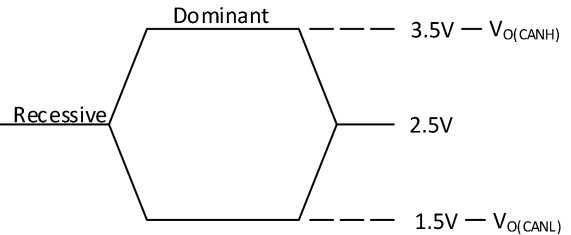


Figure 2. Bus Logic State Voltage Definitions

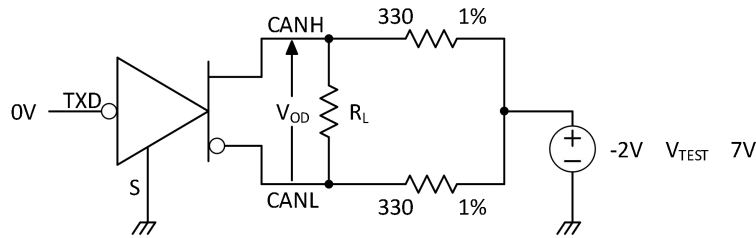


Figure 3. Driver VOD Test Circuit

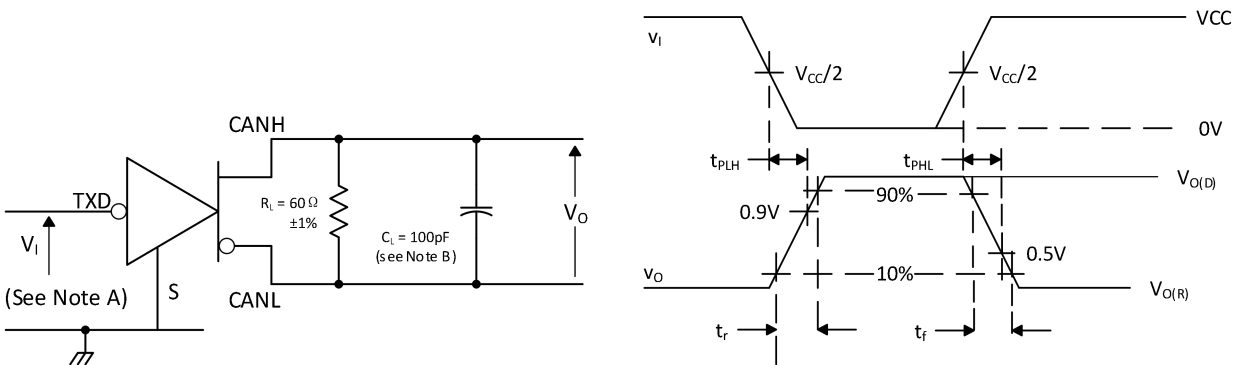


Figure 4. Driver Test Circuit and Voltage Waveforms

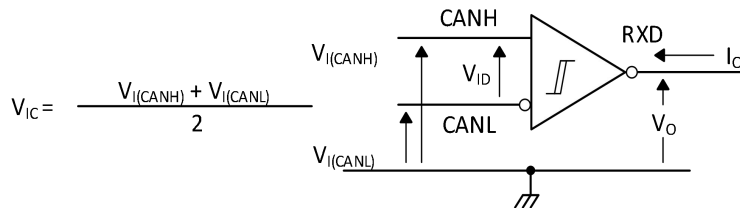
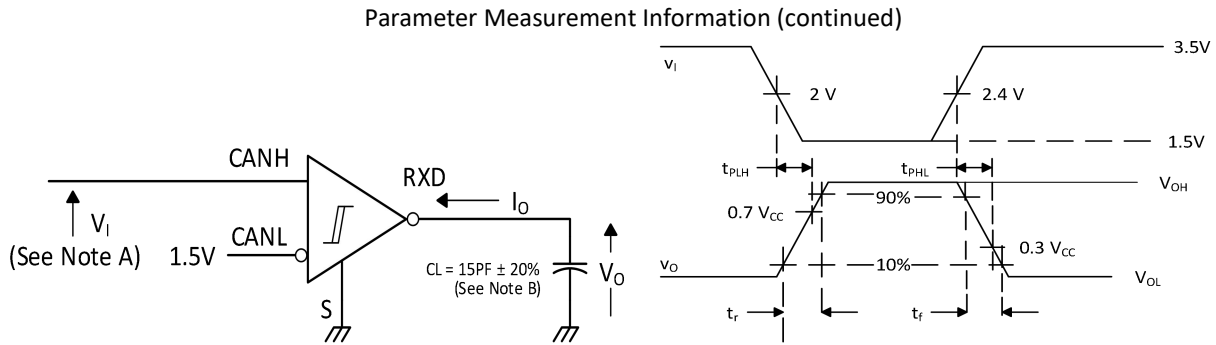


Figure 5. Receiver Voltage and Current Definitions



A. The input pulse is supplied by a generator having the following characteristics: PRR \leq 125 kHz, 50% duty cycle, $t_r \leq 6$ ns, $t_f \leq 6$ ns, $Z_o = 50 \Omega$.

B. CL includes instrumentation and fixture capacitance within $\pm 20\%$.

Figure 6. Receiver Test Circuit and Voltage Waveforms

7 Specifications

7.1 Overview

The HMT1050 CAN transceiver is compatible with the ISO1189-2 High Speed CAN (Controller Area Network) physical layer standard. It is designed to interface between the differential bus lines in controller area network and the CAN protocol controller at data rates up to 1 M/bps.

7.2 Feature Description

7.2.1 Mode control

7.2.1.1 Normal Mode

Select the normal mode of the device operation by setting the S pin low. The CAN bus driver and receiver are fully operational and the CAN communication is bidirectional. The driver is translating a digital input on TXD to a differential output on CANH and CANL. The receiver is translating the differential signal from CANH and CANL to a digital output on RXD.

7.2.1.2 Silent Mode

Activate silent mode (receive only) by setting the S pin high. The CAN driver is turned off while the receiver remains active and RXD outputs the received bus data.

7.2.2 TXD Dominant Timeout(DTO)

During normal mode, the mode where the CAN driver is active, the TXD DTO circuit prevents the transceiver from blocking network communication in the event of a hardware or software failure where TXD is held dominant longer than the timeout period t_{TXD_DTO} . The DTO circuit is triggered on a falling edge on the driver input, TXD. The DTO circuit disables the CAN bus driver if no rising edge is seen on TXD before the timeout period expires. This frees the CAN bus for communication between other nodes on the network. The CAN driver is re-enabled when a rising edge is seen on the driver input, TXD, thus clearing the TXD DTO condition. The receiver and RXD pin still reflect the CAN bus, and the bus pins are biased to recessive level during a TXD DTO.

7.2.3 Thermal Shutdown

The HMT1050 has a thermal shutdown feature that turns off the driver outputs when the junction temperature nears 190°C. This shutdown prevents catastrophic failure from bus shorts, but does not protect the circuit from possible damage. The user should strive to maintain recommended operating conditions and not exceed absolute-maximum ratings at all times. If an HMT1050 is subjected to many, or long-duration faults that can put the device into thermal shutdown, it should be replaced.

7.2.4 V_{REF}

A reference voltage of $V_{CC}/2$ is available through the V_{REF} output pin. The V_{REF} voltage should be tied to the common mode point in a split termination network to help stabilize the output common mode voltage. See [Figure 5](#) for more application specific information on properly terminating the CAN bus. If the V_{REF} output pin is not used it can be left floating.

7.2.5 Operating Temperature Range

The HMT1050 is characterized for operation from -40°C to 125°C.

7.3 Device Functional Modes

Table 1. Driver

| INPUTS | | OUTPUTS | | BUS STATE |
|--------------------|------------------|---------------------|---------------------|-----------|
| TXD ⁽¹⁾ | S ⁽¹⁾ | CANH ⁽¹⁾ | CANL ⁽¹⁾ | |
| L | L or Open | H | L | DOMINANT |
| H | X | Z | Z | RECESSIVE |
| Open | X | Z | Z | RECESSIVE |
| X | H | Z | Z | RECESSIVE |

(1) H = high level; L = low level; X = irrelevant; Z = high impedance

Table 2. Receiver

| DIFFERENTIAL INPUTS $V_{ID} = V(CANH) - V(CANL)$ | OUTPUT RXD ⁽¹⁾ | BUS STATE |
|---|---------------------------|-----------|
| $V_{ID} \geq 0.9V$ | L | DOMINANT |
| $0.5V < V_{ID} < 0.9V$ | ? | ? |
| $V_{ID} \leq 0.5V$ | H | RECESSIVE |
| Open | H | RECESSIVE |

(1) H = high level; L = low level; X = irrelevant; ? = indeterminate; Z = high impedance

8 Application And Implementation

8.1 Application Information

The CAN bus has two states during powered operation of the device; dominant and recessive. A dominant bus state is when the bus is driven differential, corresponding to a logic low on the D and R pin. A recessive bus state is when the bus is biased to $V_{CC}/2$ via the high-resistance internal resistors R_{IN} and R_{ID} of the receiver, corresponding to a logic high on the D and R pins. See [Figure 7](#) and [Figure 8](#).

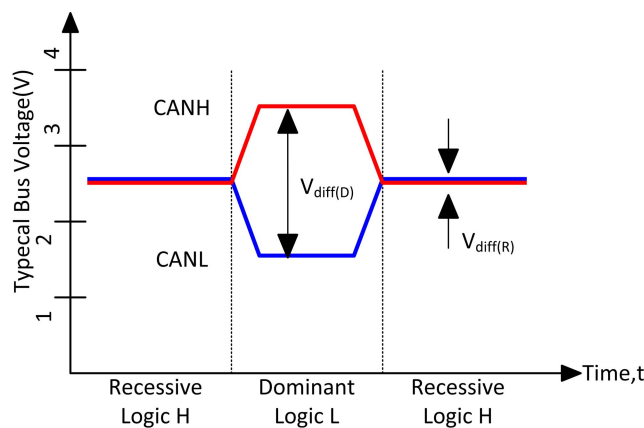


Figure 7. Bus States

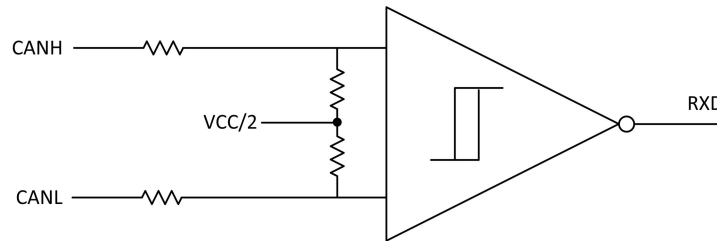


Figure 8. Simplified Recessive Common Mode Bias and Receiver

These CAN transceivers are typically used in applications with a host microprocessor or FPGA that includes the link layer portion of the CAN protocol. The different nodes on the network are typically connected through the use of a 120Ω characteristic impedance twisted-pair cable with termination on both ends of the bus.

8.2 Typical Application

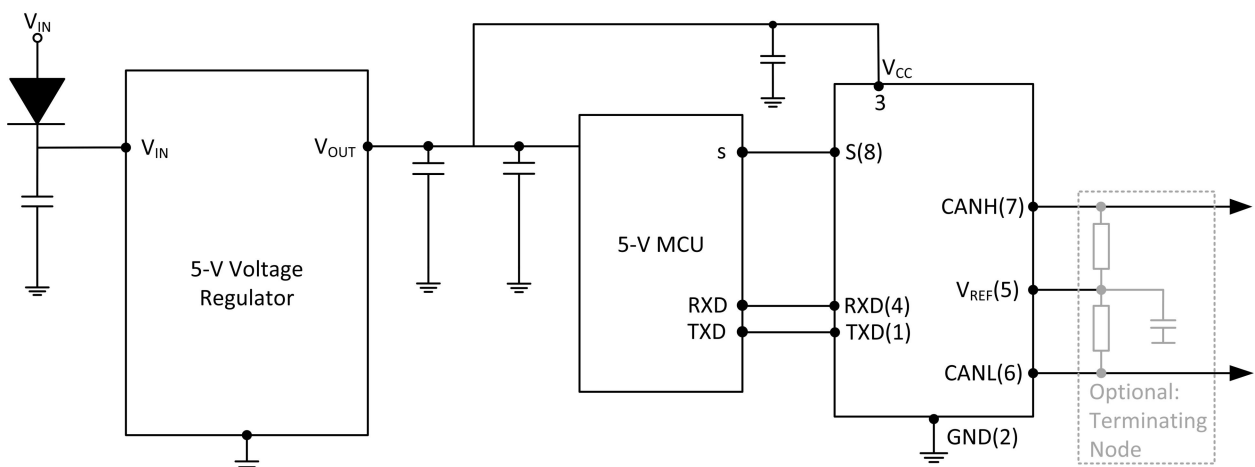


Figure 9. Typical Application Schematic

8.2.1 Design Requirements

8.2.1.1 Bus Loading, Length, and Number of Nodes

The ISO 11898 Standard specifies up to 1 M/bps data rate, maximum bus length of 40 meters, maximum drop line (stub) length of 0.3 meters and a maximum of 30 nodes. However, with careful network design, the system may have longer cables, longer stub lengths, and many more nodes to a bus. Many CAN organizations and standards have scaled the use of CAN for applications outside the original ISO 11898 standard. They have made system level trade-offs for data rate, cable length, and parasitic loading of the bus. Examples of some of these specifications are ARINC825, CANopen, CAN Kingdom, DeviceNet and NMEA200.

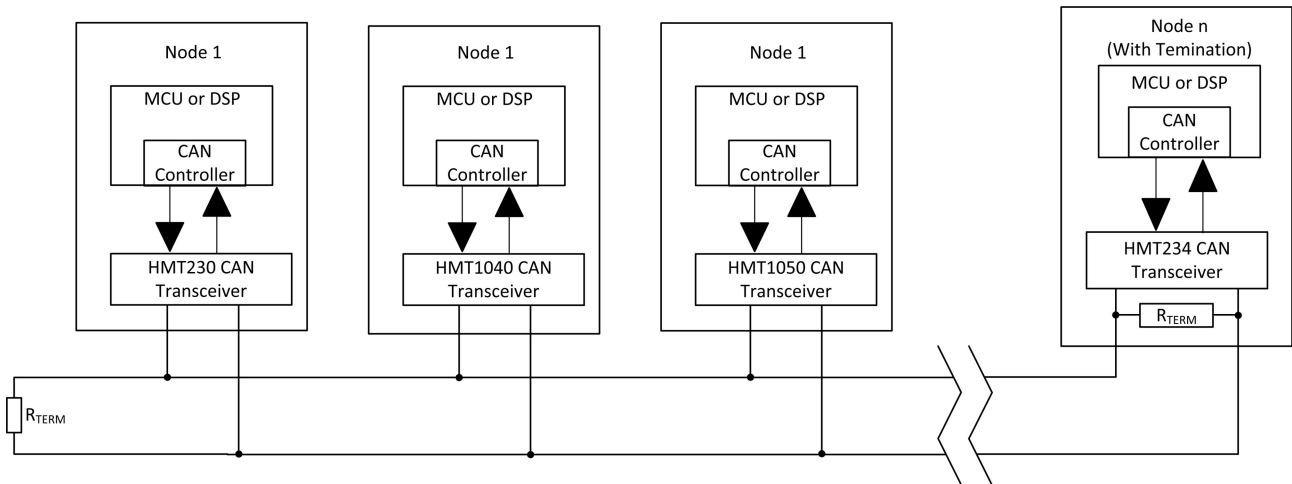


Figure 10. Typical CAN Bus

A high number of nodes requires a transceiver with high input impedance and wide common mode range such as the HMT1050 CAN transceiver. ISO 11898-2 specifies the driver differential output with a 60-Ω load (two 120Ω termination resistors in parallel) and the differential output must be greater than 1.5 V. The HMT1050 device is specified to meet the 1.5-V requirement with a 60-Ω load, and additionally specified with a differential output voltage minimum of 1.2 V across a common mode range of -2 V to 7 V via a 330-Ω coupling network. This network represents the bus loading of 90 HMT1050 transceivers based on their minimum differential input resistance of 30kΩ. Therefore, the HMT1050 supports up to 90 transceivers on a single bus segment with margin to the 1.2-V minimum differential input voltage requirement at each node.

For CAN network design, margin must be given for signal loss across the system and cabling, parasitic loadings, network imbalances, ground offsets and signal integrity thus a practical maximum number of nodes may be lower. Bus length may also be extended beyond the original ISO 11898 standard of 40 meters by careful system design and data rate tradeoffs. For example, CANopen network design guidelines allow the network to be up to 1-km with changes in the termination resistance, cabling, less than 64 nodes and significantly lowered data rate.

This flexibility in CAN network design is one of the key strengths of the various extensions and additional standards that have been built on the original ISO 11898 CAN standard.

8.2.1.2 CAN Termination

The ISO 11898 standard specifies the interconnect to be a twisted pair cable (shielded or unshielded) with 120Ω characteristic impedance (Z_0). Resistors equal to the characteristic impedance of the line should be used to terminate both ends of the cable to prevent signal reflections. Unterminated drop lines (stubs) connecting nodes to the bus should be kept as short as possible to minimize signal reflections. The termination may be on the cable or in a node, but if nodes may be removed from the bus the termination must be carefully placed so that it is not removed from the bus.

Termination is typically a 120Ω resistor at each end of the bus. If filtering and stabilization of the common mode voltage of the bus is desired, then split termination may be used (see [Figure 11](#)). Split termination uses two 60Ω resistors with a capacitor in the middle of these resistors to ground. Split termination improves the electromagnetic emissions behavior of the network by eliminating fluctuations in the bus common mode voltages at the start and end of message transmissions.

Care should be taken when determining the power ratings of the termination resistors. A typical worst case fault condition is if the system power supply and ground were shorted across the termination resistance which would result in much higher current through the termination resistance than the current limit of the CAN transceiver.

Standard Termination Split Termination

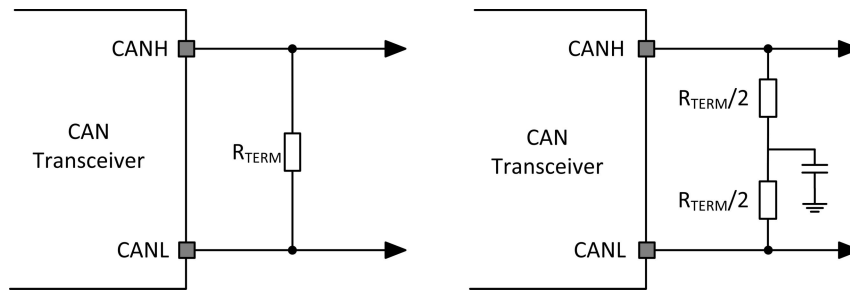


Figure 11. CAN Termination

8.2.1.3 Loop Propagation Delay

Transceiver loop delay is a measure of the overall device propagation delay, consisting of the delay from the driver input (TXD pin) to the differential outputs (CANH and CANL pins), plus the delay from the receiver inputs (CANH and CANL) to its output (RXD pin).

9 Layout

9.1 Layout Guidelines

In order for the PCB design to be successful, start with design of the protection and filtering circuitry. Because ESD and EFT transients have a wide frequency bandwidth from approximately 3-MHz to 3-GHz, high frequency layout techniques must be applied during PCB design. On chip IEC ESD protection is good for laboratory and portable equipment but is usually not sufficient for EFT and surge transients occurring in industrial environments. Therefore robust and reliable bus node design requires the use of external transient protection devices at the bus connectors. Placement at the connector also prevents these harsh transient events from propagating further into the PCB and system.

Use V_{CC} and ground planes to provide low inductance.

Design the bus protection components in the direction of the signal path. Do not force the transient current to divert from the signal path to reach the protection device. An example placement of the Transient Voltage Suppression (TVS) device indicated as D1 (either bidirectional diode or varistor solution) and bus filter capacitors C5 and C7 are shown in [Figure 12](#).

The bus transient protection and filtering components should be placed as close to the bus connector, J1, as possible. This prevents transients, ESD and noise from penetrating onto the board and disturbing other devices.

Bus termination: [Figure 12](#) shows split termination. This is where the termination is split into two resistors, R5 and R6, with the center or split tap of the termination connected to ground via capacitor C6. Split termination provides common mode filtering for the bus. When termination is placed on the board instead of directly on the bus, care must be taken to ensure the terminating node is not removed from the bus as this will cause signal integrity issues of the bus is not properly terminated on both ends.

Bypass and bulk capacitors should be placed as close as possible to the supply pins of transceiver, examples C2, C3 (V_{CC}).

Use at least two vias for V_{CC} and ground connections of bypass capacitors and protection devices to minimize trace and via inductance.

To limit current of digital lines, serial resistors may be used. Examples are R1, R2, R3 and R4.

To filter noise on the digital IO lines, a capacitor may be used close to the input side of the IO as shown by C1 and C4.

Because the internal pullup and pulldown biasing of the device is weak for floating pins, an external 1k Ω to 10k Ω pullup or pulldown resistor should be used to bias the state of the pin more strongly against noise during transient events.

Pin 1: If an open-drain host processor is used to drive the TXD pin of the device an external pullup resistor between 1k Ω and 10k Ω should be used to drive the recessive input state of the device.

Pin 5: V_{REF} should be connected to the center point of a split termination scheme to help stabilize the common mode voltage to $V_{CC}/2$. If V_{REF} is unused it should be left floating.

Pin 8: Is shown assuming the mode pin, S, will be used. If the device will only be used in normal mode, R3 is not needed and the pads of C4 could be used for the pulldown resistor to GND.

9.2 Layout Example

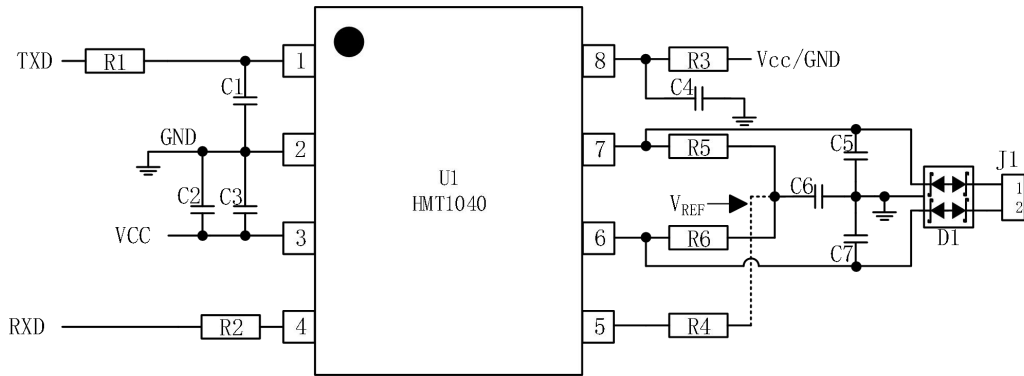
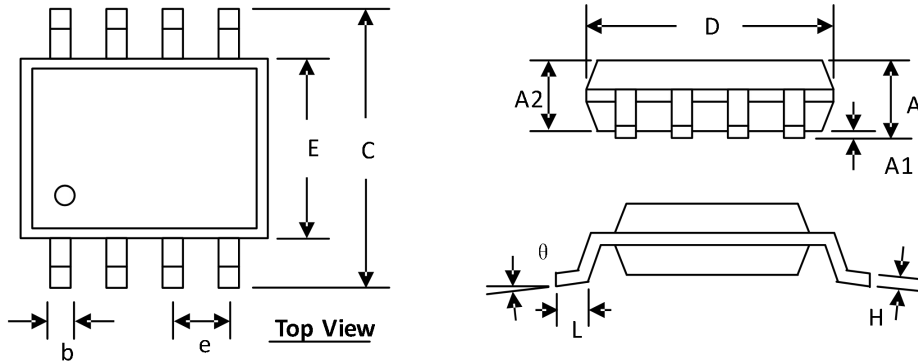


Figure 12. Layout Example

PACKAGE DIMENSION SOP-8L



| SYMBOLS | DIMENSION (MM) | | DIMENSION (INCH) | |
|---------|----------------|-------|------------------|-------|
| | MIN | MAX | MIN | MAX |
| A | 1.300 | 1.752 | 0.051 | 0.069 |
| A1 | 0.000 | 0.203 | 0.000 | 0.008 |
| A2 | 1.350 | 1.550 | 0.053 | 0.061 |
| b | 0.330 | 0.510 | 0.013 | 0.020 |
| C | 5.790 | 6.200 | 0.228 | 0.244 |
| D | 4.700 | 5.110 | 0.185 | 0.201 |
| E | 3.800 | 4.000 | 0.150 | 0.157 |
| e | 1.270 BSC | | 0.050 BSC | |
| H | 0.170 | 0.254 | 0.007 | 0.010 |
| L | 0.400 | 1.270 | 0.016 | 0.050 |
| θ | 0° | 8° | 0° | 8° |

Order Information

| Order number | Package | Marking information | Operation Temperature Range | MSL Grade | Ship, Quantity | Green |
|--------------|---------|---------------------|-----------------------------|-----------|----------------|-------|
| HMT1050T | SOP8-L | HMT1050T | -40 to 125°C | 3 | T&R, 2500 | RoHS |