

Inductor Built-in Negative Output Voltage DC/DC Converters

☆Green Operation Compatible

■GENERAL DESCRIPTION

MYRNP050030W21RA / MYRNP050030B21RA are small coil-integrated negative voltage DC/DC converter IC. The oscillating frequency is a fast 2.5MHz and the small 2.5 x 2.0 x 1.0mm package contributes significantly to space saving in PCB area.

Further, integrating the coil together with the DC/DC simplifies the circuit board layout and minimizes potential noise interference. Compared to a charge pump type solution, the switching method of the MYRNP-W/B series maintains a stable output voltage even when the input voltage fluctuates. In addition, this new mini DC/DC can support larger output current than a charge pump solution.

The PWM controlled MYRNP050030W21RA can be selected for applications where low noise is important, and the PWM/PFM automatic switching controlled MYRNP050030B21RA can be selected for applications where high efficiency at light load current and low noise at high load current is important. The MYRNP-W/B series allows users to select either a PWM control or PWM/PFM automatic

switching control method, which are optimum for applications where low noise and high efficiency are important. Output voltage can be adjusted within the range of -1.2V to -6.0V using externally mounted resistors.

■APPLICATIONS

- Negative power supply for Optical tra
- Negative power supply for AMP
- Negative power supply for LCD
- Negative power supply for CCD
- General purpose Negative power sup

■FEATURES

■TYPICAL APPLICATION CIRCUIT ■ TYPICAL PERFORMANCE

L1

7

 L_x V_{IN} 6

1

SD

2

 C_{L}

CVRE

 R_{FB}

 R_{FB}

FB

VREE

 $\overline{2}$

8

 $\overline{\mathcal{L}}$

 $3 \mid V_{REF}$ CE $4 \mid$

5 GND

 C_{IN}

MYRNP050030W21RA/MYRNP050030B21RA $(V_{N} = 3.7V, V_{\text{out}} = -3.3V)$

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■ BLOCK DIAGRAM

* Diodes inside the circuit are an ESD protection diode and a parasitic diode.

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■PIN CONFIGURATION

(BOTTOM VIEW)

* The dissipation pad should be solder-plated in recommended mount pattern and metal masking to enhance mounting strength and heat release. If the pad needs to be connected to other pins, it should be connected to the GND (No. 5) pin.

■PINASSIGNMENT

FUNCTION

* Please do not leave the CE pin open.

\blacksquare ABSOLUTE MAXIMUM RATINGS $_{\text{Ta}=25^\circ\text{C}}$

* All voltages are described based on the GND pin.

(*1) The maximum value should be either VIN+0.3V or +6.2V in the lowest.

(*2) The power dissipation figure shown is PCB mounted and is for reference only.

(*3) Please do not apply voltage to the VREF pin from outside.

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■ELECTRICAL CHARACTERISTICS

MYRNP050030W21RA, MYRNP050030B21RA Ta=25˚C

*Unless otherwise stated, $V_{IN} = V_{CE} = 3.7V$

(*1) "H" = $V_{IN} \sim V_{IN}$ -1.2V, "L" = +0.1V \sim -0.1V ^(*2) V_{FB(E)} : Effective FB Voltage,

(*3) V_{FB(T)} : Setting FB Voltage(0.5V)

(*4) ON resistance = (V_{IN} – V_{LX} pin measurement voltage) / 100mA

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■TEST CIRCUITS INNOVATOR IN ELECTRON
TEST CIRCU
< Test Circuit No.① >

< Test Circuit No.^② >

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ITYPICAL APPLICATION CIRCUIT

EXTERNAL COMPONENTS SELECTION

【Typical example】

*Take capacitance loss, withstand voltage, rated current and other conditions into consideration when selecting components.

 $*$ 10μF ~ 44μF output capacitor (CL) value is recommended.

When the output capacitor (CL) is large, there is a possibility that the output voltage will be unstable.

 $*$ If a tantalum or electrolytic capacitor is used for the output capacitor (C_L), ripple voltage will increase, and there is a possibility that operation will become unstable. Test fully using the actual device.

*When Schottky Diodes, which have a large junction capacity are used, there is a possibility that the output voltage will be unstable.

<Output voltage (VOUTSET) setting>

Output voltage can be set by adding an external resistor.

Output voltage is set by the following equation according to RFB1, RFB2, VFB and VREF.

 $V_{\text{OUTSET}} = V_{FB} - R_{FB1} / R_{FB2} \times (V_{REF} - V_{FB})$

Please select within 100k $\Omega \leq R_{FB1} + R_{FB2} \leq 500 \text{k}\Omega$ range.

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TYPICAL APPLICATION CIRCUIT

EXTERNAL COMPONENTS SELECTION (Continued)

 \le Setting soft start time (tss) $>$

Soft start time is determined by the capacity of the C_{VREF} connected to the V_{REF} terminal. Please select the capacitance value of C_{VREF} within the range of 0.47μ F ~ 10 μ F referring to the below graph.

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■OPERATIONAL EXPLANATION

This IC consists of a standard voltage reference, error amp, ramp wave circuit, oscillator circuit, PWM comparator, PWM/PFM controller, Pch driver transistor, current sensing circuit, UVLO circuit, VREF startup circuit and etc.

Control method is a current mode control method which allows for the use of low ESR ceramic capacitors.

MYRNP050030W21RA / MYRNP050030B21RA block diagram

Inductor Built-in Negative Output Voltage DC/DC Converters

■OPERATIONAL EXPLANATION (Continued)

<Normal Operation>

The FB terminal voltage divided by the output voltage is compared with the VREF voltage by the error amp. Phase compensation is applied to the error amp output, which is then forwarded to the PWM comparator. At the PWM comparator the error amp output and ramp wave are compared to determine the ON time during PWM control.

The MYRNP050030W21RA (PWM control) is switched using a constant switching frequency (fosc) independent of the output current. During light load current, the ON time is short, and the IC operates in a non-continuous mode. As the output current increases, the ON time becomes longer, and the IC operates in a continuous mode.

At high load currents, the ON time depends heavily on the input voltage, output voltage, and output current, and the maximum ON time (toNMAX) restriction determines the maximum output current that can flow under the conditions of each input voltage and output voltage.

Refer to the typical performance characteristics for the maximum output current under each condition.

The MYRNP050030B21RA (PWM/PFM automatic switching control) turns ON the Pch driver transistor until the coil current reaches the PFM current (IPFM) and to lower the switching frequency during light load current. This operation reduces loss during light loads to achieve high efficiency from light to high load currents.

As the output current grows larger, the switching frequency increases proportional to the output current, and when the switching frequency reaches the fosc to switch from PFM control to PWM control the switching frequency is fixed.

MYRNP050030B21RA: Example of operation at light load current MYRNP050030B21RA: Example of operation at high load currents

Further, the phase compensation circuit optimizes the error amp frequency characteristics and is used to phase compensate the Pch driver transistor current feedback signal. This achieves output voltage stability even when low ESR capacitors, such as ceramic capacitors are used.

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■OPERATIONAL EXPLANATION (Continued)

<CE Function>

When a "H" voltage (V_{CEH}) is input to the CE terminal, it operates normally after the output voltage is started by the soft start function.

When a "L" voltage (V_{CEL}) is input to the CE terminal, it goes to the stand-by state, the quiescent current is suppressed to the stand-by current I_{STB} (TYP.0 μA) level and the Pch driver transistor turns OFF.

<UVLO Function>

When the V_{IN} terminal voltage drops below the UVLO detect voltage level (VUVLOD), the UVLO function operates and turns off the Pch driver transistor to prevent any erroneous pulse output due to possible unstable action of the internal circuit.

When the VIN terminal voltage increases above the UVLO release voltage level (VUVLOR), the UVLO function is released. After the UVLO function is released, the soft start function starts the output voltage and the IC operates normally.

The UVLO function operates even if the VIN terminal momentarily drops below the UVLO detect voltage.

In addition, whilst the UVLO function is in operation, rather than being in a stand-by state, the IC is in a switching operation stopped state, so the internal circuit is still operating.

<Soft Start Function>

This gently starts up the output voltage when the IC starts up and the UVLO function is released to suppress the inrush current. The VREF startup circuit operates after the "H" voltage (VCEH) is input to the CE terminal and after the UVLO function is released. The VREF startup circuit charges the CVREF with current and can gently raise the VREF voltage and FB voltage. In response to this, the output voltage is lowered proportionally to the increase in the VREF voltage and FB voltage. This action makes it possible to prevent input current inrush and to smoothly lower the output voltage.

The output voltage startup time (soft start time) is determined by the capacity of the C_{VREF} connected to the VREF terminal.

In the stand-by state and during the UVLO function operation, the charge accumulated in the C_{VREF} is discharged and the VREF voltage is made to be 0V.

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■OPERATIONAL EXPLANATION (Continued)

<Current Limit Function>

The current limit circuit monitors the current flowing to the Pch driver transistor to restrict overcurrent. The current limit function operates as follows.

- 1) The current flowing to the Pch driver transistor is increased, and when the current limit value of ILIM=1100mA (TYP.) is reached, the current limit state is entered and the Pch driver transistor is turned OFF.
- 2) The Pch driver transistor is turned OFF for a period of 4μs (TYP.), and the coil current is greatly decreased. During this time, lowering the coil current that has reached the current limit lowers the input current and output current while the current is restricted.
- 3) Other switching operations are performed, and when the output voltage is a load resistance that does not reach the set voltage, the coil current increases and the current limit function operates again.
- 4) Operations 1) to 3) are repeated during the current limit state period.
- 5) When the load resistance increases much more than the load resistance during current limit detection, the current limit state is released and the IC automatically returns to normal operation.

Inductor Built-in Negative Output Voltage DC/DC Converters

■NOTE ON USE

1) For temporary, transitional voltage drop or voltage rising phenomenon, the IC is liable to malfunction should the ratings be exceeded.

2) Switching regulators like this DC/DC converter generate spike noise and ripple voltage. This greatly affects the surrounding components (Schottky diodes, capacitors, peripheral component circuit board layout etc.). When making a design, please be sure to sufficiently check this in an actual device.

3) The DC/DC converter characteristics greatly depend not only on the characteristics of this IC but also on those of externally connected components, so refer to EXTERNAL COMPONENTS SELECTION and the specifications of each component and be careful when selecting the components. Be especially careful of the characteristics of the capacitor used for the load capacity CL and use a capacitor with B characteristics (JIS Standard) or an X7R/X5R (EIA Standard) ceramic capacitor.

4) The maximum output current of this IC is determined by the current limit value and the maximum ON time restrictions, and this depends greatly on the input voltage and output voltage. Further, when the input voltage is low and during low temperature, there is a possibility that the maximum ON time decreases and the maximum output current drops. For the maximum output current, please refer to the typical performance characteristics of "Maximum Output Current vs. Output Voltage."

5) With MYRNP050030W21RA, there is a possibility that the switching frequency will decline when the input voltage is high and the load current is light.

6) When Schottky Diodes, which have a large junction capacity, are used or when the CL output capacity is large, there is a possibility that the output voltage will be unstable.

7) When there is steep output current fluctuation, there could be a large drop in the output voltage that can cause the duty to increase which in turn triggers the operation of the current limit function.

8) If the IC is started under a condition where the output current is large, there is a possibility that the inrush current will increase and the current limit function may operate.

9) When the input voltage is lowered below the UVLO detect voltage level for a short time, there are times when it is not possible to discharge the CVREF charge. When the input voltage is started again in this state, the shortening of the soft start time at startup could trigger the current limit function.

10) Under the condition where the input voltage is close to 1V, there is a possibility that the UVLO function will not operate.

11) Murata places an importance on improving our products and their reliability. We request that users incorporate fail-safe designs and post-aging protection treatment when using Murata products in their systems.

12) The proper position of mounting is based on the coil terminal

Inductor Built-in Negative Output Voltage DC/DC Converters

■NOTE ON USE (Continued)

- 13) Note on board layout
	- 1. In order to stabilize V_{IN} voltage level, we recommend that a by-pass capacitor (C_N) be connected as close as possible to the V_{IN} & GND pins.
	- 2. Please mount each external component as close to the IC as possible.
	- 3. Wire external components as close to the IC as possible and use thick, short connecting traces to reduce the circuit impedance.
	- 4. Make sure that the PCB GND traces are as thick as possible, as variations in ground potential caused by high ground currents at the time of switching may result in instability of the IC.
	- 5. This series' internal driver transistors bring on heat because of the output current and ON resistance of Pch driver transistors.
	- 6. As precautions on mounting, please set the mounting position accuracy within 0.05 mm.

●Recommended Pattern Layout

14) Appearance(Coil)

1. Coils are compliant with general surface mount type chip coil (inductor) specifications and may have scratches, flux contamination and the like.

Inductor Built-in Negative Output Voltage DC/DC Converters

■TYPICAL PERFORMANCE CHARACTERISTICS

MYRNP050030W21RA VOUT =-3.3V CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ 80 $V_{\mathbb{N}}$ =5.0V Vin = Vc∈
Ta = 25℃ 70 $V_{\text{IN}} = 3.7V$ 60 Efficiency: EFFI (%) Efficiency : EFFI (%) $V_{\text{IN}}=2.7V$ 50 40 30 20 10 0 10 100 1 10 100 1000 Output Current : $I_{OUT}(mA)$

CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ 80 70 60 $(%)$

80

MYRNP050030B21RA VOUT =-3.3V

 $V_{IN} = V_{CE}$
Ta = 25℃

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Inductor Built-in Negative Output Voltage DC/DC Converters

■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(2) Output Voltage vs. Output Current

MYRNP050030W21RA VOUT = -3.3V CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ

MYRNP050030B21RA VOUT = -3.3V CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ

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TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(3) Ripple Voltage vs. Output Current

"PicoBK™" MYRNP-W / MYRNP-B series η Mum η

Inductor Built-in Negative Output Voltage DC/DC Converters

■TYPICAL PERFORMANCE CHARACTERISTICS (Continued) IUAL FERFURIVIAINUE U Inducto
AL PERFORMANCE
Im Output Current vs. Output Voltage ϵ **XCL30xA052** ^CIN=10μF(GRM188D71A106KA73), CVREF= 1μF(GRM155C71A105KE11)

(4) Maximum Output Current vs. Output Voltage $(5) V_{REF}$ Voltage vs. Ambient Temperature

(5) V_{REF} Voltage vs. Ambient Temperature

Nent Lemperar COL Version of Temperature
MYRNP050030*21RA

0.44 0.46 0.48 0.50 0.52 0.54 -50 -25 FB Voltage: V_{REF} (V) 0 25 50 75 100 125 Ambient Temperature: Ta(℃) **MYRNP050030*21RA** $MYRNP050030^*21RA$ CIN=10μF, CL=10μF, CVREF=1μF $V_{IN} = V_{CE} = 3.7V$ $\frac{2}{\pi}$ \tilde{a} \mathcal{L} -50 -25 0 25 50 75 100 125

(6) FB Voltage vs. Ambient Temperature **(8)** (7) Supply Current vs. Ambient Temperature Urrent vs. Ambient Temperature

 F CIN=10μF, CL=10μF, CVREF=1μF 50
40
30
20 500 **ר** $V_{IN} = V_{CE} = 5.5V$ $V_{FB} = V_{FB(T)} \times 0.975$ Supply Current: lop (µA) Supply Current : l_{DD} (μA) 400 300 300 $\overline{1}$ 200 \blacksquare 100 l -50 -25 0 25 50 75 100 125 0 -50 -25 0 25 50 75 100 125 25 50 $\frac{1}{5}$ 100 Ambient Temperature: Ta(℃)

MYRNP050030*21RA

Inductor Built-in Negative Output Voltage DC/DC Converters

TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(10) PFM Switch Current vs. Ambient Temperature (11) Maximum Current Limit vs. Ambient Temperature

(12) Oscillation Frequency vs. Ambient Temperature (13) Maximum ON Time vs. Ambient Temperature

(14) Minimum OFF Time vs. Ambient Temperature (15) Lx SW "H" ON Resistance vs. Ambient Temperature

Inductor Built-in Negative Output Voltage DC/DC Converters

■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(16) Lx SW "L" Leakage Current vs. Ambient Temperature (17) CE "H" Voltage vs. Ambient Temperature (16) Lx SW "L" Leakage Current vs. Ambient Temperature (17) CE "H" Voltage vs. Ambient Temperature

(18) CE "L" Voltage vs. Ambient Temperature

Inductor Built-in Negative Output Voltage DC/DC Converters

■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(19) Rising Output Voltage

CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ

CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ

MYRNP050030W21RA

CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ

MYRNP050030B21RA

CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ

Inductor Built-in Negative Output Voltage DC/DC Converters

■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(20) Load Transient Response

MYRNP050030W21RA

CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ

CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ

500μs/div

MYRNP050030W21RA

CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ

 $V_{\text{in}} = V_{\text{CE}} = 3.7V$, $I_{\text{out}} = 10 \text{mA} \rightarrow 100 \text{mA}$ Ta = 25° C, V_{OUT} = -3.3V Ta = 25° C, V_{OUT} = -3.3V **VOUT (500mV/div) IOUT (100mA/div) VLX (5V/div) 500μs/div**

^VOUT MYRNP050030B21RA

CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ

^VOUT MYRNP050030B21RA

CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ

MYRNP050030W21RA

CIN=10μF, CL=10μF, CVREF=1μF, RFB1=150kΩ, RFB2=43kΩ

Inductor Built-in Negative Output Voltage DC/DC Converters

■PACKAGING INFORMATION

●Packaging (2.0mm × 2.5mm, h=1.0mm)

*Implementation is recommended within accuracy 0.05mm

●Reference Pattern Layout (unit:mm) ●Reference Metal Mask Design (unit:mm)

Inductor Built-in Negative Output Voltage DC/DC Converters

■PACKAGING INFORMATION

Power Dissipation

Power dissipation data for the package is shown in this page. The value of power dissipation varies with the mount board conditions. Please use this data as one of reference data taken in the described condition.

2. Power Dissipation vs. Ambient Temperature(85℃) Board Mount (Tjmax=125℃)

3. Power Dissipation vs. Ambient Temperature(105℃)

Inductor Built-in Negative Output Voltage DC/DC Converters

Represents products series

FB voltage

Represents integer of Oscillation frequency

, ⑤ represents production lot number

~09、0A~0Z、11~9Z、A1~A9、AA~AZ、B1~ZZ in order. (G, I, J, O, Q, W excluded)

Note: No character inversion used.

"PicoBK™" MYRNP-W / MYRNP-B series Inductor Built-in Negative Output Voltage DC/DC Converters

- 1. The product and product specifications contained herein are subject to change without notice to improve performance characteristics. Consult us, or our representatives before use, to confirm that the information in this datasheet is up to date.
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- 5. Although we make continuous efforts to improve the quality and reliability of our products; nevertheless Semiconductors are likely to fail with a certain probability. So in order to prevent personal injury and/or property damage resulting from such failure, customers are required to incorporate adequate safety measures in their designs, such as system fail safes, redundancy and fire prevention features.
- 6. Our products are not designed to be Radiation-resistant.
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