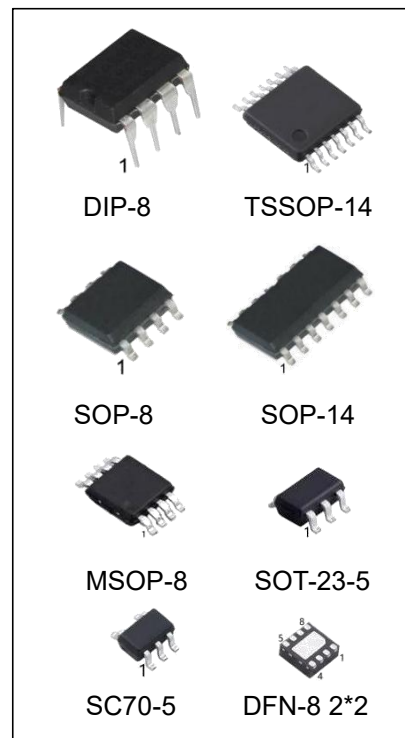


## 1MHz CMOS Rail-to-Rail IO Opamp with RF Filter

### Features

- Single-Supply Operation from +2.1V ~ +5.5V
- Rail-to-Rail Input / Output
- Gain-Bandwidth Product: 1MHz (Typ.)
- Low Input Bias Current: 1pA (Typ.)
- Low Offset Voltage: 3.5mV (Max.)
- Quiescent Current: 40μA per Amplifier (Typ.)
- Operating Temperature: -40°C ~ +125°C
- Embedded RF Anti-EMI Filter
- Small Package:  
LMV551 Available in SOT23-5 and SC70-5 Packages  
LMV552 Available in SOP-8, MSOP-8, DIP-8, DFN-8 2\*2 Packages  
LMV554 Available in SOP-14 and TSSOP-14 Packages



### Ordering Information

DEVICE	Package Type	MARKING	Packing	Packing Qty
LMV551M7/TR	SC70-5(SOT-353)	V551	REEL	3000/reel
LMV551M5/TR	SOT-23-5	V551	REEL	3000/reel
LMV552M/TR	SOP-8	LMV552	REEL	2500/reel
LMV552MM/TR	MSOP-8	V552	REEL	3000/reel
LMV552N	DIP-8	LMV552	TUBE	2000/box
LMV552LQ/TR	DFN-8 2*2	V552	REEL	4000/reel
LMV554M/TR	SOP-14	LMV554	REEL	2500/reel
LMV554MT/TR	TSSOP-14	LMV554	REEL	2500/reel

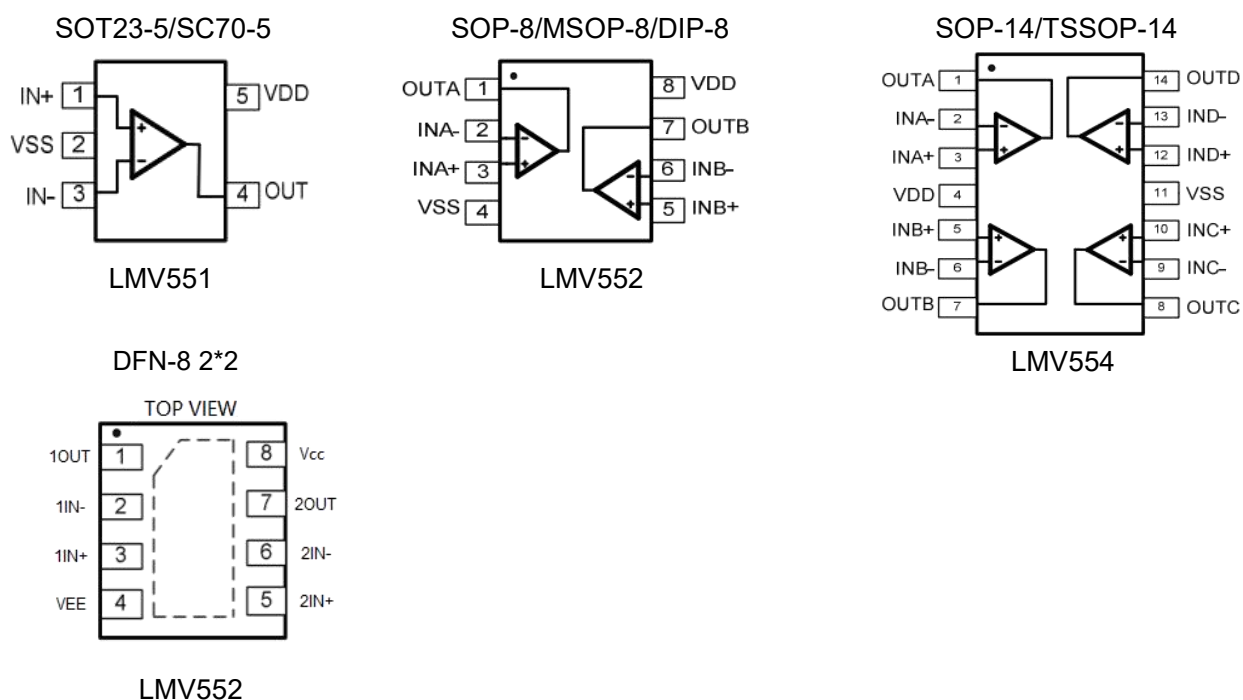
## General Description

The LMV551 family have a high gain-bandwidth product of 1MHz, a slew rate of 0.6V/μs, and a quiescent current of 40μ A/amplifier at 5V. The LMV551 family is designed to provide optimal performance in low voltage and low noise systems. They provide rail-to-rail output swing into heavy loads. The input common mode voltage range includes ground, and the maximum input offset voltage is 3.5mV for LMV551 family. They are specified over the extended industrial temperature range (-40°C to +125°C). The operating range is from 2.1V to 5.5V.

## Applications

- ASIC Input or Output Amplifier
- Sensor Interface
- Medical Communication
- Smoke Detectors
- Audio Output
- Piezoelectric Transducer Amplifier
- Medical Instrumentation
- Portable Systems

## Pin Configuration



**Figure 1. Pin Assignment Diagram**

## Absolute Maximum Ratings

Condition	Min	Max
Power Supply Voltage (VDD to Vss)	-0.5V	+7.5V
Analog Input Voltage (IN+ or IN-)	Vss-0.5V	VDD+0.5V
PDB Input Voltage	Vss-0.5V	+7V
Operating Temperature Range	-40°C	+125°C
Junction Temperature	+160°C	
Storage Temperature Range	-55°C	+150°C
Lead Temperature (soldering, 10sec)	+245°C	
Package Thermal Resistance (TA=+25°C)		
SOP-8, θJA	125°C/W	
MSOP-8, θJA	216°C/W	
SOT23-5, θJA	190°C/W	
ESD Susceptibility		
HBM	6KV	
MM	300V	

**Note:** Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

## Electrical Characteristics

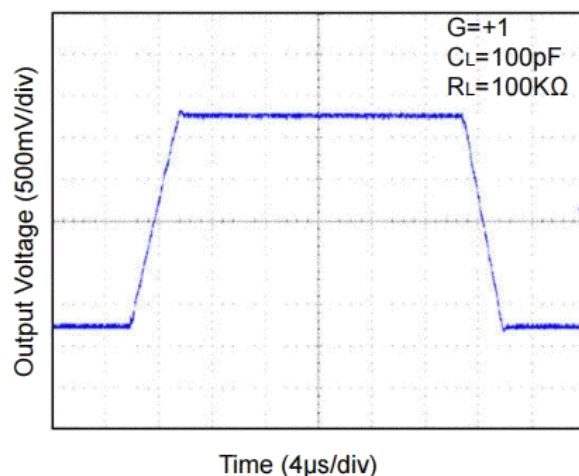
(At  $V_S = +5V$ ,  $R_L = 100k\Omega$  connected to  $V_S/2$ , and  $V_{OUT} = V_S/2$ , unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	LMV551/552/554				
			TYP	MIN/MAX OVER TEMPERATURE			
			+25°C	+25°C	-40°C to +85°C	UNITS	MIN/MAX
INPUT CHARACTERISTICS							
Input Offset Voltage	VOS	VCM = VS/2	0.4	3.5	5.6	mV	MAX
Input Bias Current	IB		1			pA	TYP
Input Offset Current	IOS		1			pA	TYP
Common-Mode Voltage Range	VCM	VS = 5.5V	-0.1 to +5.6			V	TYP
Common-Mode Rejection Ratio	CMRR	VS = 5.5V, VCM = -0.1V to 4V	70	62	62	dB	MIN
		VS = 5.5V, VCM = -0.1V to 5.6V	68	56	55		
Open-Loop Voltage Gain	AOL	RL = 5kΩ, VO = +0.1V to +4.9V	80	70	70	dB	MIN
		RL = 10kΩ, VO = +0.1V to +4.9V	100	90	85		
Input Offset Voltage Drift	ΔVOS/ΔT		2.7			μV/°C	TYP
OUTPUT CHARACTERISTICS							
Output Voltage Swing from Rail	VOH	RL = 100kΩ	4.997	4.990	4.980	V	MIN
	VOL	RL = 100kΩ	3	10	20	mV	MAX
	VOH	RL = 10kΩ	4.992	4.970	4.960	V	MIN
	VOL	RL = 10kΩ	8	30	40	mV	MAX
Output Current	ISOURCE	RL = 10Ω to VS/2	84	60	45	mA	MIN
	ISINK		75	60	45		
POWER SUPPLY							
Operating Voltage Range				2.1	2.5	V	MIN
				5.5	5.5	V	MAX
Power Supply Rejection Ratio	PSRR	VS = +2.5V to +5.5V, VCM = +0.5V	82	60	58	dB	MIN
Quiescent Current / Amplifier	IQ		40	60	80	μA	MAX
DYNAMIC PERFORMANCE (CL = 100pF)							
Gain-Bandwidth Product	GBP		1			MHz	TYP
Slew Rate	SR	G = +1, 2V Output Step	0.6			V/μs	TYP
Settling Time to 0.1%	tS	G = +1, 2V Output Step	5			μs	TYP
Overload Recovery Time		VIN · Gain = VS	2.6			μs	TYP
NOISE PERFORMANCE							
Voltage Noise Density	en	f = 1kHz	27			nV√Hz	TYP
		f = 10kHz	20			nV√Hz	TYP

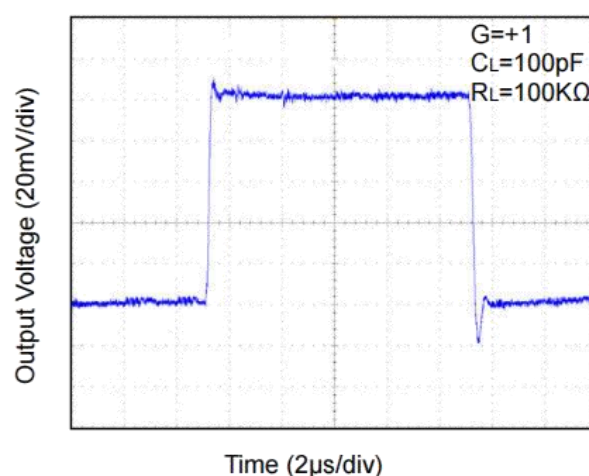
## Typical Performance characteristics

At  $T_A = +25^{\circ}\text{C}$ ,  $V_S = +5\text{V}$ , and  $R_L = 100\text{K}\Omega$  connected to  $V_S/2$ , unless otherwise noted.

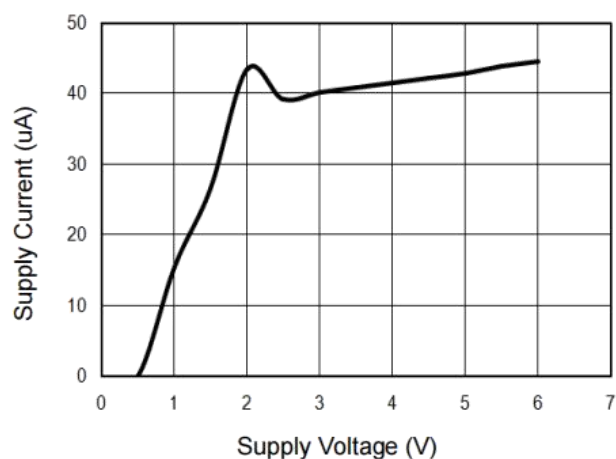
Large-Signal Step Response



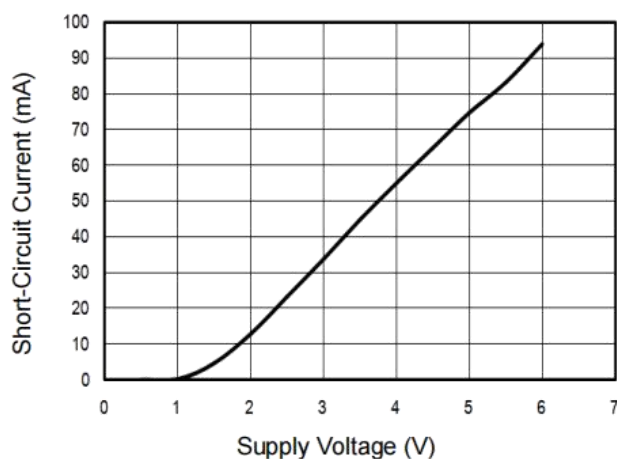
Small-Signal Step Response



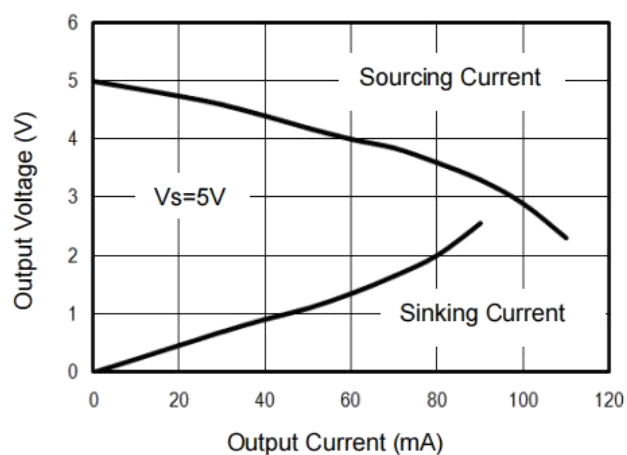
Supply Current vs. Supply Voltage



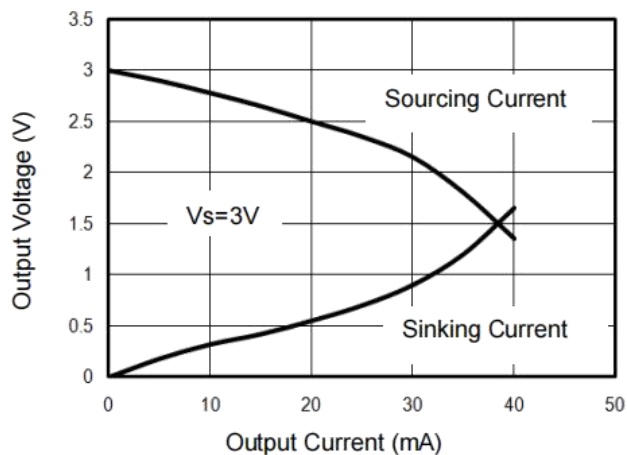
Short-Circuit Current vs. Supply Voltage



Output Voltage vs. Output Current



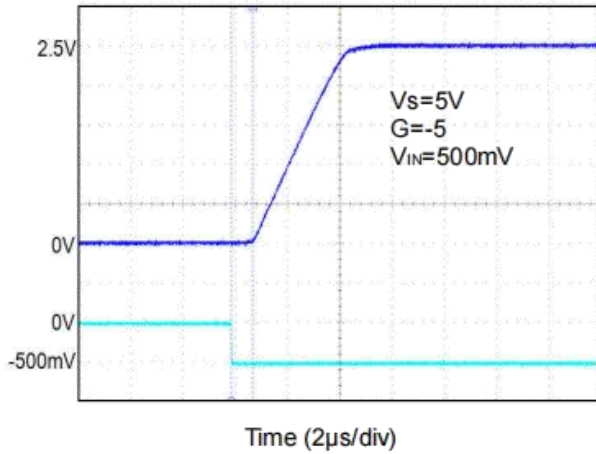
Output Voltage vs. Output Current



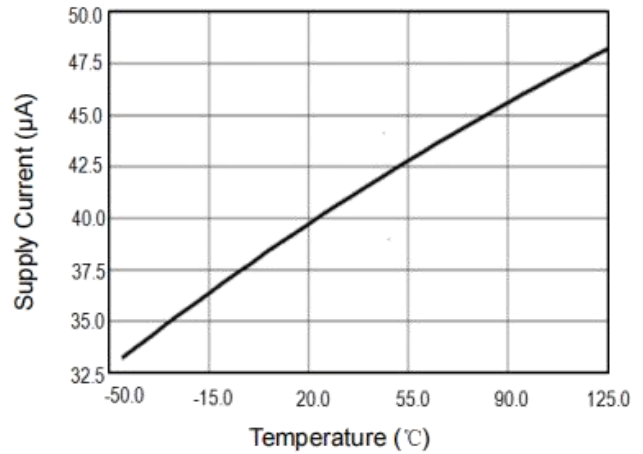
## Typical Performance characteristics

At TA=+25°C, VS=+5V, and RL=100KΩ connected to VS/2, unless otherwise noted.

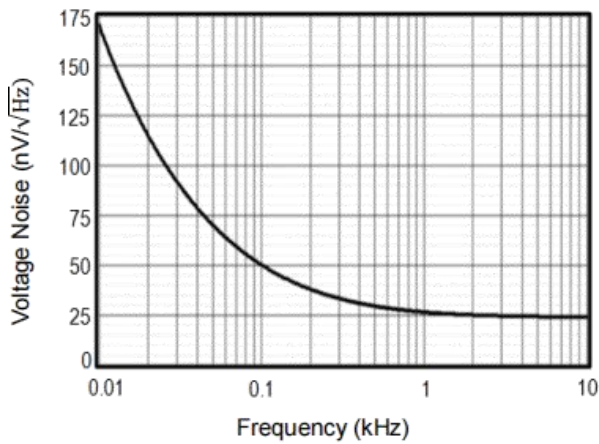
Overload Recovery Time



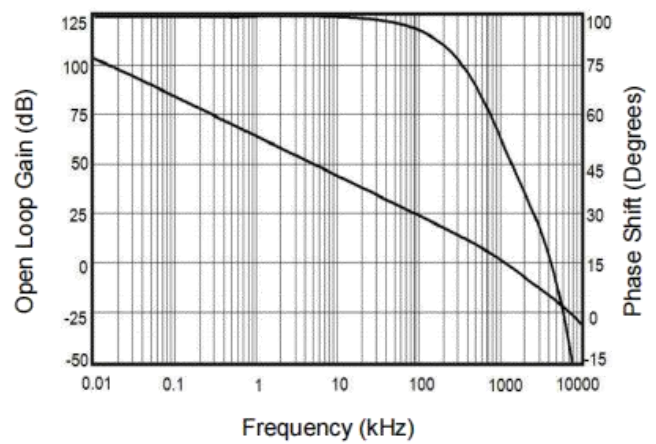
Supply Current vs. Temperature



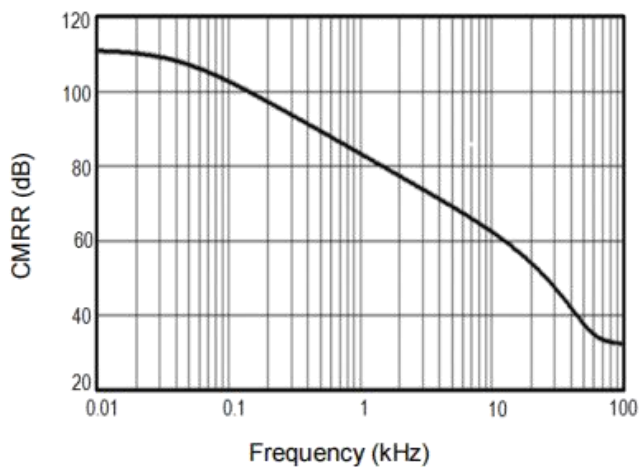
Input Voltage Noise Spectral Density vs. Frequency



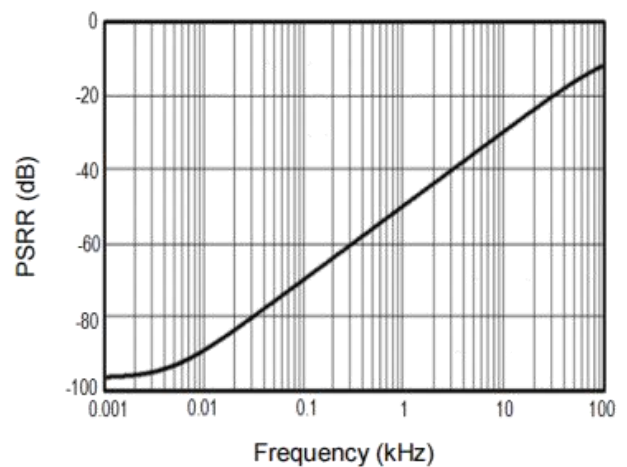
Open Loop Gain, Phase Shift vs. Frequency at +5V



CMRR vs. Frequency



PSRR vs. Frequency



## Application Note

### Size

LMV551 family series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The small footprints of the LMV551 family packages save space on printed circuit boards and enable the design of smaller electronic products.

### Power Supply Bypassing and Board Layout

LMV551 family series operates from a single 2.1V to 5.5V supply or dual  $\pm 1.05\text{V}$  to  $\pm 2.75\text{V}$  supplies. For best performance, a  $0.1\mu\text{F}$  ceramic capacitor should be placed close to the VDD pin in single supply operation. For dual supply operation, both VDD and VSS supplies should be bypassed to ground with separate  $0.1\mu\text{F}$  ceramic capacitors.

### Low Supply Current

The low supply current (typical  $40\mu\text{A}$  per channel) of LMV551 family will help to maximize battery life. They are ideal for battery powered systems

### Operating Voltage

LMV551 family operates under wide input supply voltage (2.1V to 5.5V). In addition, all temperature specifications apply from  $-40^\circ\text{C}$  to  $+125^\circ\text{C}$ . Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-Ion battery lifetime

### Rail-to-Rail Input

The input common-mode range of LMV551 family extends  $100\text{mV}$  beyond the supply rails ( $\text{VSS}-0.1\text{V}$  to  $\text{VDD}+0.1\text{V}$ ). This is achieved by using complementary input stage. For normal operation, inputs should be limited to this range.

### Rail-to-Rail Output

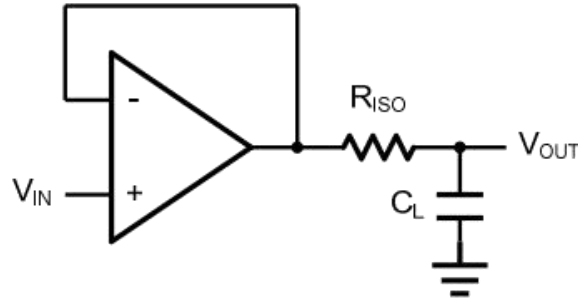
Rail-to-Rail output swing provides maximum possible dynamic range at the output. This is particularly important when operating in low supply voltages. The output voltage of LMV551 family can typically swing to less than  $5\text{mV}$  from supply rail in light resistive loads ( $>100\text{k}\Omega$ ), and  $30\text{mV}$  of supply rail in moderate resistive loads ( $10\text{k}\Omega$ ).

### Capacitive Load Tolerance

The LMV551 family is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain.

Figure 2. shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

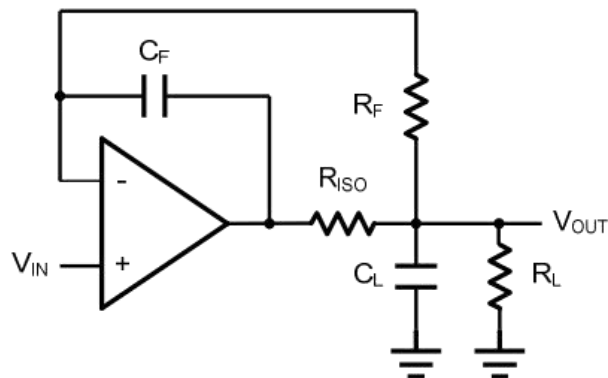




**Figure 2. Indirectly Driving a Capacitive Load Using Isolation Resistor**

The bigger the  $R_{ISO}$  resistor value, the more stable  $V_{OUT}$  will be. However, if there is a resistive load  $R_L$  in parallel with the capacitive load, a voltage divider (proportional to  $R_{ISO}/R_L$ ) is formed, this will result in a gain error.

The circuit in Figure 3 is an improvement to the one in Figure 2.  $R_F$  provides the DC accuracy by feed-forward the  $V_{IN}$  to  $R_L$ .  $C_F$  and  $R_{ISO}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of  $C_F$ . This in turn will slow down the pulse response.



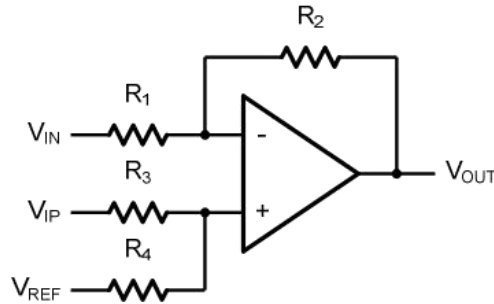
**Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy**



## Typical Application Circuits

### Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common to the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4. shown the differential amplifier using LMV551 family.



**Figure 4. Differential Amplifier**

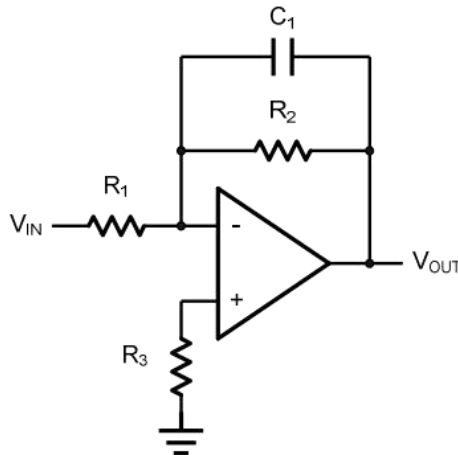
$$V_{OUT} = \left( \frac{R_1 + R_2}{R_3 + R_4} \right) \frac{R_4}{R_1} V_{IN} - \frac{R_2}{R_1} V_{IP} + \left( \frac{R_1 + R_2}{R_3 + R_4} \right) \frac{R_3}{R_1} V_{REF}$$

If the resistor ratios are equal (i.e.  $R_1=R_3$  and  $R_2=R_4$ ), then

$$V_{OUT} = \frac{R_2}{R_1} (V_{IP} - V_{IN}) + V_{REF}$$

### Low Pass Active Filter

The low pass active filter is shown in Figure 5. The DC gain is defined by  $-R_2/R_1$ . The filter has a -20dB/decade roll-off after its corner frequency  $f_C = 1/(2\pi R_3 C_1)$ .



**Figure 5. Low Pass Active Filter**

## Differential Amplifier

The triple LMV551 family can be used to build a three -op-amp differential amplifier as shown in Figure 6. The amplifier in Figure 6 is a high input impedance differential amplifier with gain of  $R_2/R_1$ . The two differential voltage followers assure the high input impedance of the amplifier.

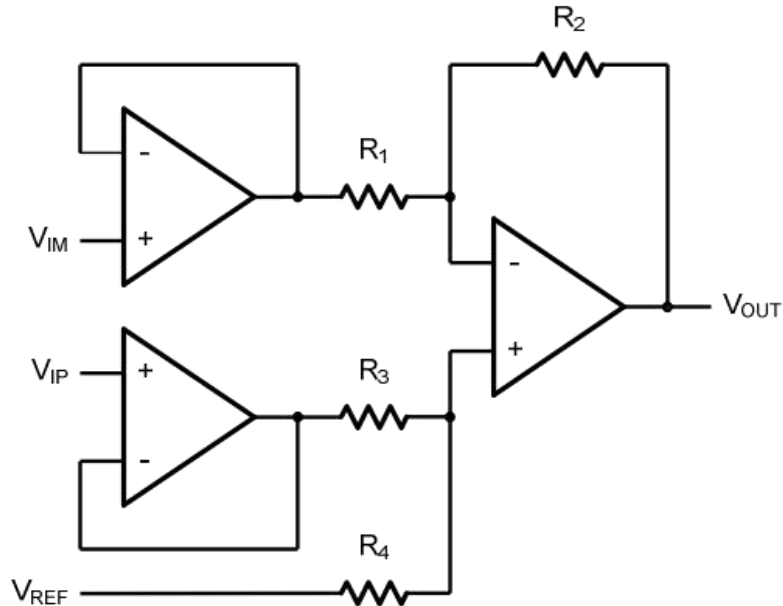
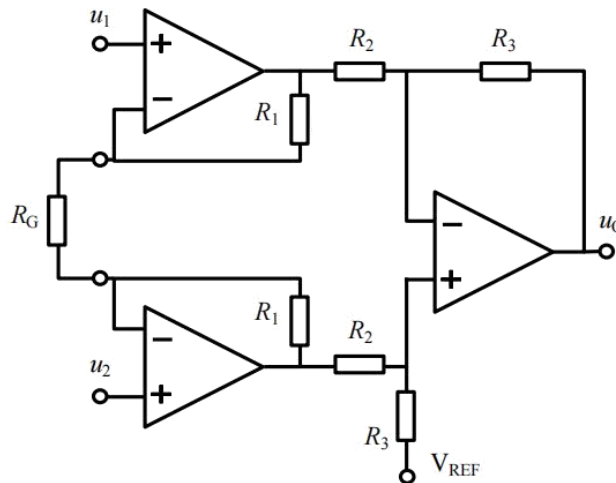


Figure 6. Differential Amplifier

## Instrumenton Amplifier



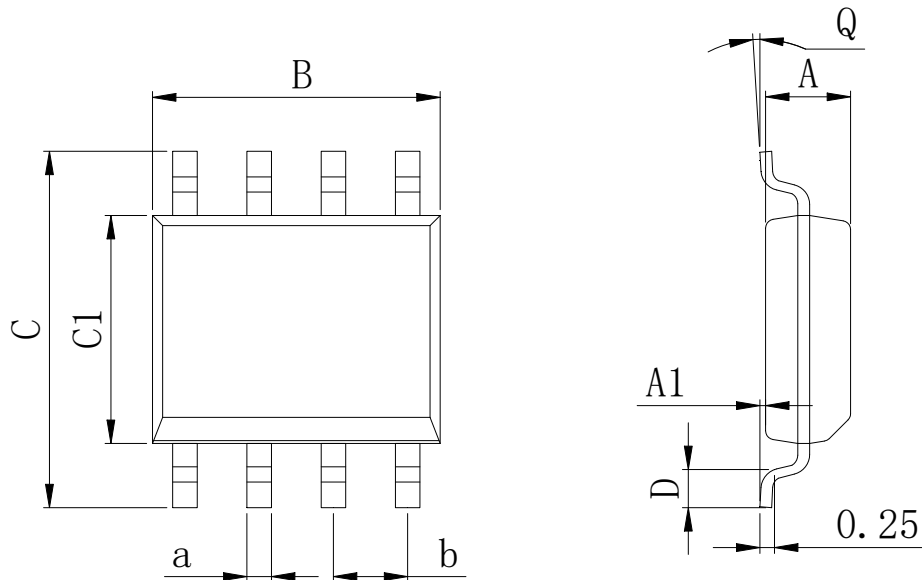
$$G = \frac{R_3}{R_2} \times \frac{R_G + 2R_1}{R_G}$$

Figure 7. Instrumenton Amplifier

By an external resistance  $R_G$ . Note that  $R_G$  can be suspended but can not short circuit. The  $V_{REF}$  pin, used to control the central position of the output voltage. When dual power supply, it is generally grounded. When a single power supply, it is generally connected to 1 / 2 power supply voltage. When  $R_G$  is open circuit, and  $R_2=R_3$ , gain  $G=1$

## Physical Dimensions

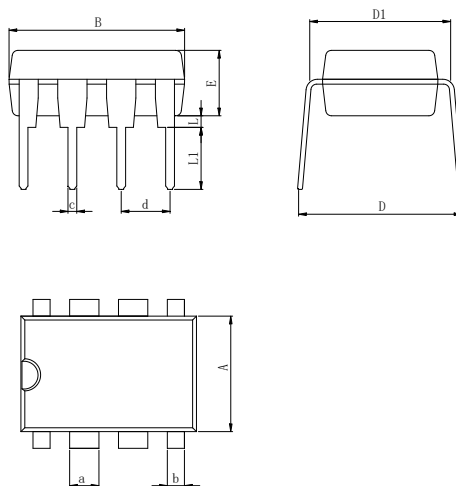
### SOP-8 (150mil)



Dimensions In Millimeters(SOP-8)

Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	1.35	0.05	4.90	5.80	3.80	0.40	0°	0.35	1.27 BSC
Max:	1.55	0.20	5.10	6.20	4.00	0.80	8°	0.45	

### DIP-8

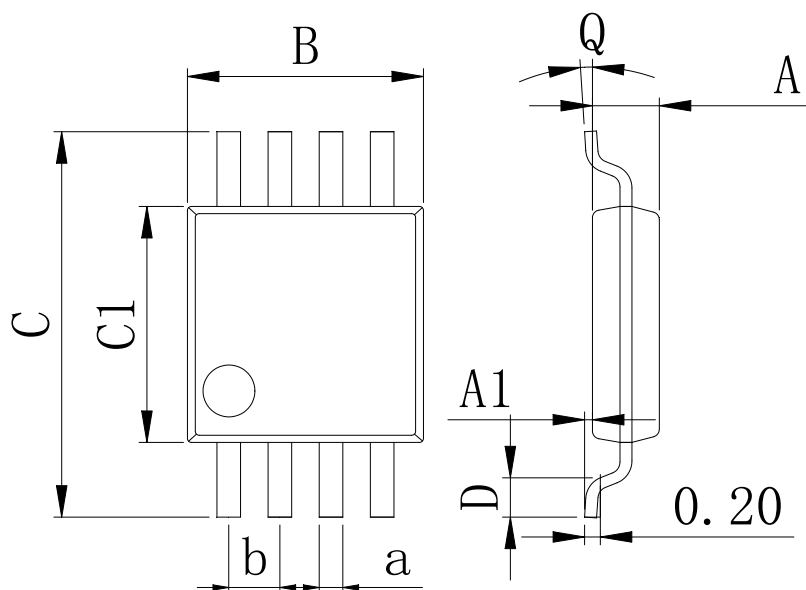


Dimensions In Millimeters(DIP-8)

Symbol:	A	B	D	D1	E	L	L1	a	b	c	d
Min:	6.10	9.00	8.10	7.42	3.10	0.50	3.00	1.50	0.85	0.40	2.54 BSC
Max:	6.68	9.50	10.9	7.82	3.55	0.70	3.60	1.55	0.90	0.50	

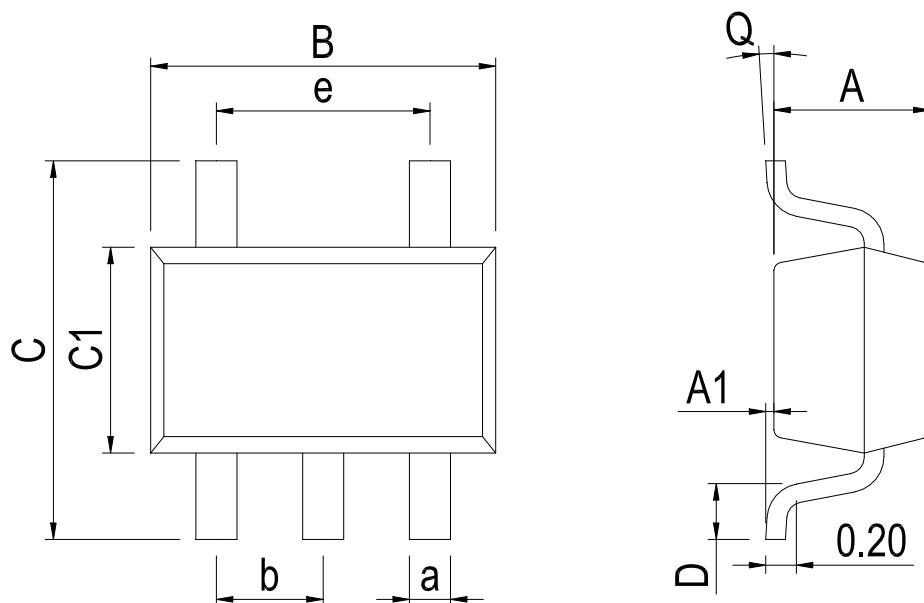
## Physical Dimensions

### MSOP-8



Dimensions In Millimeters(MSOP-8)									
Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	0.80	0.05	2.90	4.75	2.90	0.35	0°	0.25	0.65 BSC
Max:	0.90	0.20	3.10	5.05	3.10	0.75	8°	0.35	

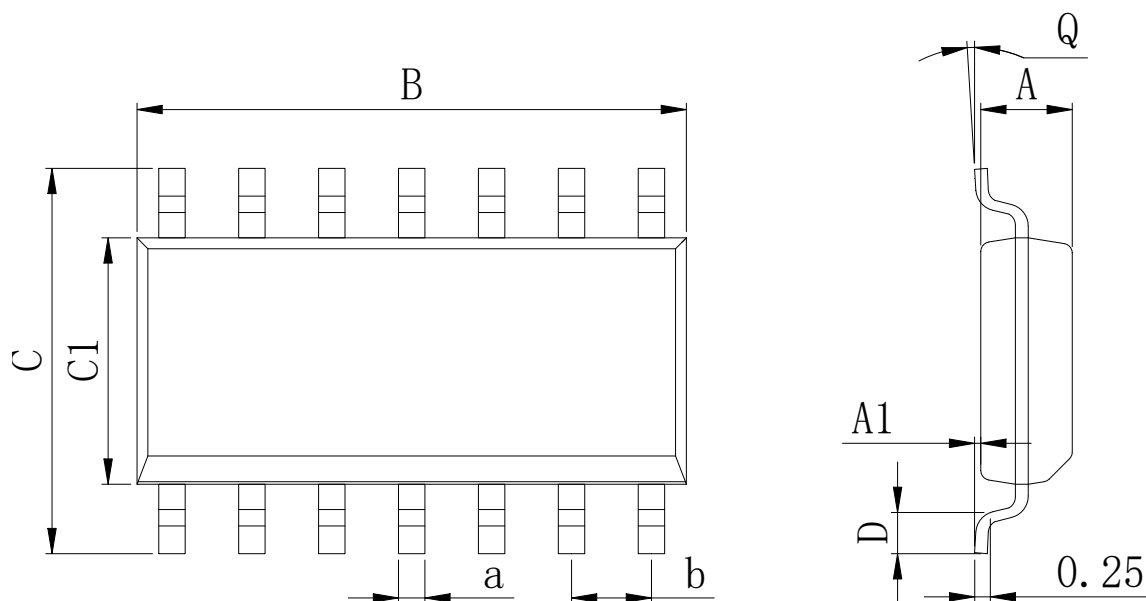
### SOT-23-5



Dimensions In Millimeters(SOT-23-5)										
Symbol:	A	A1	B	C	C1	D	Q	a	b	e
Min:	1.05	0.00	2.82	2.65	1.50	0.30	0°	0.30	0.95 BSC	1.90 BSC
Max:	1.15	0.15	3.02	2.95	1.70	0.60	8°	0.40		

## Physical Dimensions

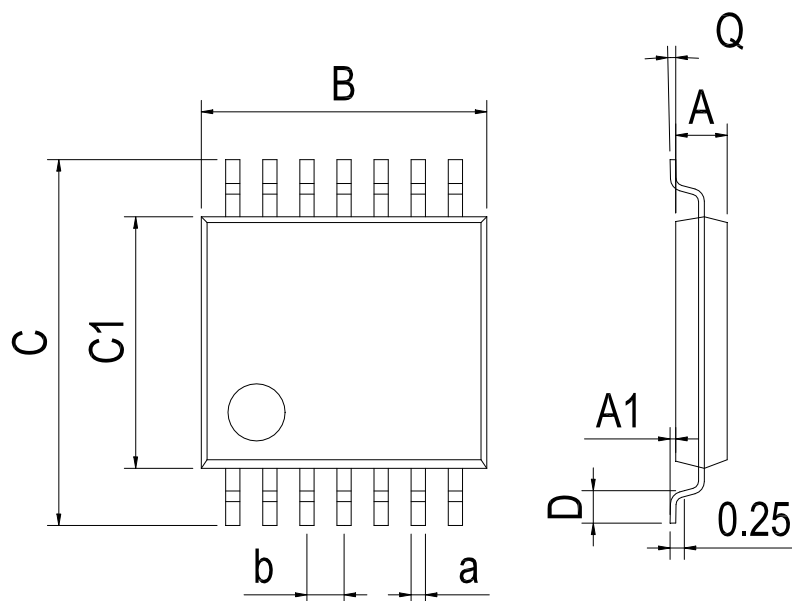
### SOP-14



Dimensions In Millimeters(SOP-14)

Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	1.35	0.05	8.55	5.80	3.80	0.40	0°	0.35	1.27 BSC
Max:	1.55	0.20	8.75	6.20	4.00	0.80	8°	0.45	

### TSSOP-14

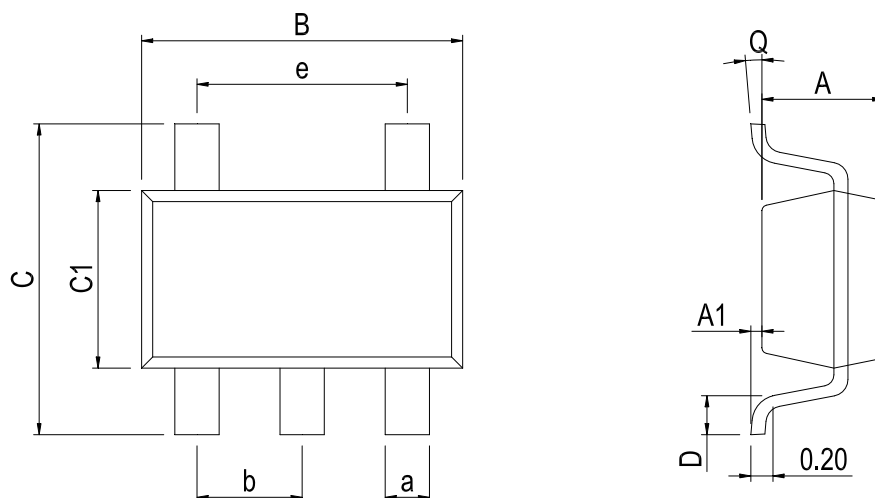


Dimensions In Millimeters(TSSOP-14)

Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	0.85	0.05	4.90	6.20	4.30	0.40	0°	0.20	0.65 BSC
Max:	0.95	0.20	5.10	6.60	4.50	0.80	8°	0.25	

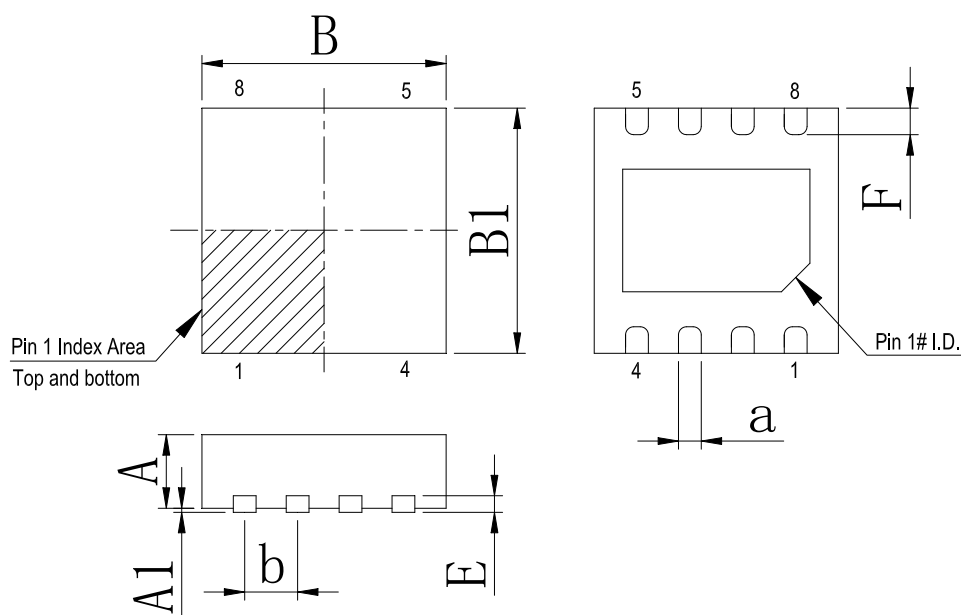
## Physical Dimensions

SC70-5



Dimensions In Millimeters(SC70-5)										
Symbol:	A	A1	B	C	C1	D	Q	a	b	e
Min:	0.90	0.00	2.00	2.15	1.15	0.26	0°	0.15	0.65	1.30 BSC
Max:	1.00	0.15	2.20	2.45	1.35	0.46	8°	0.35	BSC	

DFN-8 2\*2



Dimensions In Millimeters(DFN-8 2*2)								
Symbol:	A	A1	B	B1	E	F	a	b
Min:	0.85	0	1.90	1.90	0.15	0.25	0.18	0.50TYP
Max:	0.95	0.05	2.10	2.10	0.25	0.45	0.30	

## Revision History

DATE	REVISION	PAGE
2018-3-16	New	1-16
2024-1-18	Add New model DFN-8 2*2	1



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