







ISOM8110 SLLSFS9A - SEPTEMBER 2023 - REVISED DECEMBER 2023

ISOM811x 3.75-kV_{RMS}, Single-Channel Opto-Emulator with Analog Transistor Output

1 Features

- Drop-in replacement and pin-to-pin upgrade to industry-standard phototransistor optocouplers
- 1-channel LED-emulator input
- Current transfer ratio (CTR) at $I_F = 5$ mA, $V_{CE} = 5$
 - ISOM8110, ISOM8115: 100% to 155%
 - ISOM8111, ISOM8116: 150% to 230%
 - ISOM8112, ISOM8117: 255% to 380%
 - ISOM8113, ISOM8118: 375% to 560%
- High collector-emitter voltage: V_{CE} (max) = 80 V
- Robust isolation barrier
 - Isolation rating: 3750-V_{RMS}
 - Working voltage: 500-V_{RMS}, 707-V_{PK}
 - Surge capability: up to 10-kV
- Temperature range: -55°C to +125°C
- Response time: 3 μ s (typical) at V_{CE} = 10 V, I_C = 2 $mA, R_I = 100 \Omega$
- Safety-related certifications planned:
 - UL 1577 recognition, 3750-V_{RMS} isolation
 - DIN EN IEC 60747-17 (VDE 0884-17) conformity per VDE
 - IEC 62368-1, IEC 61010-1 certifications
 - CQC GB 4943.1 certification

2 Applications

- Switching power supply
- Programmable Logic Controller (PLC)
- Motor drive I/O and position feedback
- Factory automation
- Data acquisition
- HEV/EV battery-management system (BMS)

3 Description

ISOM811x devices are single-channel optocoupler-emulators with LED-emulator input and transistor output. The devices are pin-compatible and drop-in replacements for many traditional optocouplers, allowing enhancement to existing systems with no PCB redesign.

ISOM811x opto-emulators offer significant reliability performance advantages compared optocouplers, including high bandwidth, low turn-off delay, low power consumption, wider temperature ranges, and tight CTR and process controls resulting in small part-to-part skew. Since there is no aging effect or temperature variation to compensate for, the emulated LED input stage consumes less power than optocouplers.

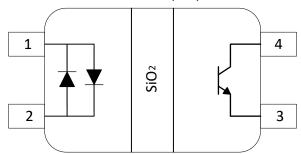
ISOM811x devices are offered in small SOIC-4 packages with 2.54-mm and 1.27-mm pin pitchs, supporting a 3.75-kV_{RMS} isolation rating and DC (ISOM811[0-3]) and bi-directional DC (ISOM811[5-8]) input options. The high performance and reliability of ISOM811x enables these devices to be used in power supply feedback design, motor drives, I/O modules in industrial controllers, factory automation applications, and more.

Device Information

PART NUMBER	PACKAGE ⁽¹⁾	PACKAGE SIZE ⁽³⁾	BODY SIZE (NOM)
ISOM811x	SO-4 (DFG)	7.0 mm × 3.5 mm	4.8 mm × 3.5 mm
	SO-4 (DFH) ⁽²⁾	7.0 mm × 2.7 mm	4.8 mm × 2.7 mm

- For all available packages, see the orderable addendum at the end of the data sheet.
- Preview only
- The package size (length × width) is a nominal value and includes pins, where applicable.

ISOM811(5-8)



Simplified Schematic

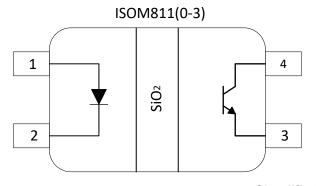




Table of Contents

1 Features1	7.9 Typical Characteristics	11
2 Applications1	8 Parameter Measurement Information	13
3 Description1	9 Detailed Description	15
4 Revision History2	9.1 Overview	
5 Device Selection3	9.2 Functional Block Diagram	15
6 Pin Configuration and Functions3	9.3 Feature Description	16
7 Specifications4	9.4 Device Functional Modes	
7.1 Absolute Maximum Ratings4	10 Application and Implementation	17
7.2 ESD Ratings5	10.1 Application Information	17
7.3 Thermal Information5	10.2 Power Supply Recommendations	21
7.4 Insulation Specifications6	10.3 Layout	21
7.5 Safety-Related Certifications	11 Mechanical, Packaging, and Orderable	
7.6 Safety Limiting Values7	Information	21
7.7 Electrical Characteristics8	11.1 Tape and Reel Information	25
7.8 Switching Characteristics10	·	

4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

С	hanges from Revision * (September 2023) to Revision A (December 2023)	Page
•	Updated to Production Data for ISOM8110	1



5 Device Selection

Table 5-1. Device Selection

PART NUMBER ²	CTR	PACKAGE ¹	PIN PITCH
ISOM8110, ISOM8115	100% to 155%		
ISOM8111, ISOM8116	150% to 230%	→ 4-pin SOIC (DFG), 4-pin SOIC (DFH)	2.54-mm. 1.27-mm
ISOM8112, ISOM8117	255% to 380%		2.34-11111, 1.27-111111
ISOM8113, ISOM8118	375% to 560%		

- 1. DFH package is preview only.
- 2. ISOM8111-3 and ISOM8115-8 are preview only.

6 Pin Configuration and Functions

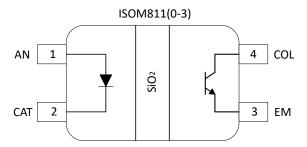


Figure 6-1. ISOM811(0-3) 4-Pin SOIC (Top View)

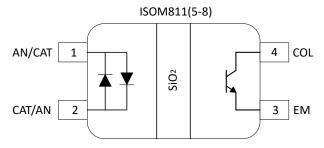


Figure 6-2. ISOM811(5-8) 4-Pin SOIC (Top View)

Table 6-1. Pin Functions

	PIN	TYPE ⁽¹⁾	DESCRIPTION	
NO.	NAME	1175		
1	AN	I	Anode connection of input LED emulator	
2	CAT	I	Cathode connection of input LED emulator	
3	EM	0	nitter for transistor	
4	COL	0	Collector for transistor	

(1) I = Input, O = Output

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7 Specifications

7.1 Absolute Maximum Ratings

See⁽¹⁾ (2)

			MIN MAX	UNIT
	Maximum Input forward current	ISOM8110, ISOM8111, ISOM8112, ISOM8113	50	mA
I _{F(max)}	Maximum Input forward current	ISOM8115, ISOM8116, ISOM8117, ISOM8118	±50	mA
V	Collector-emitter voltage	ISOM8110, ISOM8111, ISOM8112, ISOM8113	80	V
V _{CEO}	Collector-entitler voltage	ISOM8115, ISOM8116, ISOM8117, ISOM8118	80	V
W	Emitter-collector voltage	ISOM8110, ISOM8111, ISOM8112, ISOM8113	7	V
V _{ECO}	Emilier-collector voltage	ISOM8115, ISOM8116, ISOM8117, ISOM8118	7	V
I _{FP}	Input pulse forward current (1 μs width)	ISOM8110, ISOM8111, ISOM8112, ISOM8113	1	А
I _{FP}	Input pulse forward current (1 μs width)	ISOM8115, ISOM8116, ISOM8117, ISOM8118	±1	Α
V _R	Input reverse voltage at I _R = 10 μA	ISOM8110, ISOM8111, ISOM8112, ISOM8113	7	V
n	Input pouce die sir stiere	ISOM8110, ISOM8111, ISOM8112, ISOM8113	140	mW
P _I	Input power dissipation	ISOM8115, ISOM8116, ISOM8117, ISOM8118	140	mW
	Callactor current	ISOM8110, ISOM8111, ISOM8112, ISOM8113	50	mA
I _C	Collector current	ISOM8115, ISOM8116, ISOM8117, ISOM8118	50	mA
D	Collector newer discinction	ISOM8110, ISOM8111, ISOM8112, ISOM8113	150	mW
P _C	Collector power dissipation	ISOM8115, ISOM8116, ISOM8117, ISOM8118	150	mW
D	Tatal name diagram	ISOM8110, ISOM8111, ISOM8112, ISOM8113	290	mW
P _T	Total power dissipation	ISOM8115, ISOM8116, ISOM8117, ISOM8118	290	mW
т	Ambient town continu	ISOM8110, ISOM8111, ISOM8112, ISOM8113	-55 125	°C
T _A	Ambient temperature	ISOM8115, ISOM8116, ISOM8117, ISOM8118	-55 125	°C
-		ISOM8110, ISOM8111, ISOM8112, ISOM8113	150	°C
TJ	Operating junction temperature	ISOM8115, ISOM8116, ISOM8117, ISOM8118	150	°C

⁽¹⁾ Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under the operational sections of this document. If used outside the listed operational conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

⁽²⁾ All specifications are at T_A = 25 °C unless otherwise noted



7.2 ESD Ratings

			VALUE	UNIT
		Human body model (HBM), per ANSI/ ESDA/JEDEC JS-001, all pins ⁽¹⁾	±2000	
V _(ESD)	Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	±1000	V

- JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process. JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

7.3 Thermal Information

		ISOM811x	
	THERMAL METRIC(1)	DFG (SOIC)	UNIT
		4 PINS	
R _{0JA}	Junction-to-ambient thermal resistance	288.8	°C/W
R ₀ JC(top)	Junction-to-case (top) thermal resistance	173.6	°C/W
R _{0JB}	Junction-to-board thermal resistance	192.9	°C/W
ΨЈТ	Junction-to-top characterization parameter	121.9	°C/W
ΨЈВ	Junction-to-board characterization parameter	190	°C/W

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.



7.4 Insulation Specifications

	PARAMETER	TEST CONDITIONS	VALUE	UNIT
	PARAMETER	TEST CONDITIONS	4-DFG	UNII
IEC 6066	64-1			
CLR	External clearance ⁽¹⁾	Side 1 to side 2 distance through air	> 5	mm
CPG	External creepage ⁽¹⁾	Side 1 to side 2 distance across package surface	> 5	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	>17	μm
CTI	Comparative tracking index	IEC 60112; UL 746A	>400	V
	Material Group	According to IEC 60664-1	II	
		Rated mains voltage ≤ 150 V _{RMS}	I-IV	
	Overvoltage category per IEC 60664-1	Rated mains voltage ≤ 300 V _{RMS}	I-IV	
		Rated mains voltage ≤ 600 V _{RMS}	I-III	
DIN VDE	V 0884-11:2017 ⁽⁶⁾			
V _{IORM}	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	707	V_{PK}
V _{IOWM}	Maximum isolation working voltage	AC voltage (sine wave); time-dependent dielectric breakdown (TDDB) test	500	V_{RMS}
		DC voltage	707	V_{DC}
V_{IOTM}	Maximum transient isolation voltage	$V_{TEST} = V_{IOTM}$, t = 60 s (qualification); $V_{TEST} = 1.2 \times V_{IOTM}$, t = 1 s (100% production)	5303	V _{PK}
V _{IMP}	Maximum impulse voltage (2)	Tested in air, 1.2/50-µs waveform per IEC 62368-1	7200	V _{PK}
V _{IOSM}	Maximum surge isolation voltage ⁽³⁾	V _{ISOM} ≥ 1.3 x V _{IMP} ; tested in oil (qualification test), 1.2/50-µs waveform per IEC 62368-1	10000	V _{PK}
		Method a: After I/O safety test subgroup 2/3, $V_{ini} = V_{IOTM}$, $t_{ini} = 60 \text{ s}$; $V_{pd(m)} = 1.2 \times V_{IORM}$, $t_m = 10 \text{ s}$	≤ 5	
q _{pd}	Apparent charge ⁽⁴⁾	Method a: After environmental tests subgroup 1, V_{ini} = V_{IOTM} , t_{ini} = 60 s; $V_{pd(m)}$ = 1.6 × V_{IORM} , t_m = 10 s	≤ 5	рС
		Method b: At routine test (100% production) and preconditioning (type test), $V_{ini} = 1.2 \times V_{IOTM}$, $t_{ini} = 1$ s; $V_{pd(m)} = 1.875 \times V_{IORM}$, $t_m = 1$ s	≤ 5	
C _{IO}	Barrier capacitance, input to output ⁽⁵⁾	V _{IO} = 0.4 × sin (2 πft), f = 1 MHz	1	pF
		V _{IO} = 500 V, T _A = 25°C	> 10 ¹²	
R _{IO}	Insulation resistance, input to output ⁽⁵⁾	$V_{IO} = 500 \text{ V}, \ 100^{\circ}\text{C} \le T_{A} \le 125^{\circ}\text{C}$	> 10 ¹¹	Ω
		V _{IO} = 500 V at T _S = 150°C	> 109	
	Pollution degree		2	
	Climatic category		40/125/21	
UL 1577	•	·	1	
V _{ISO}	Withstand isolation voltage	V_{TEST} = V_{ISO} , t = 60 s (qualification); V_{TEST} = 1.2 × V_{ISO} , t = 1 s (100% production)	3750	V_{RMS}

- (1) Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves, ribs, or both on a printed circuit board are used to help increase these specifications.
- (2) Testing is carried out in air to determine the surge immunity of the package.
- (3) Testing is carried out in oil to determine the intrinsic surge immunity of the isolation barrier.
- (4) Apparent charge is electrical discharge caused by a partial discharge (pd).
- (5) All pins on each side of the barrier tied together creating a two-pin device.
- (6) This coupler is suitable for *safe electrical insulation only* within the safety ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.

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7.5 Safety-Related Certifications

VDE	CSA	UL	CQC	TUV
	Plan to certify according to IEC 61010-1, IEC 62368-1 and IEC 60601-1	to III 1577 Component	CB4043 1-2011	Plan to certify according to EN 61010-1:2010/A1:2019 and EN 62368-1:2014
Certificate planned	Certificate planned	Certificate planned	Certificate planned	Certificate planned

7.6 Safety Limiting Values

Safety limiting(1) intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry.

	PARAMETER	TEST CONDITIONS	MIN TYP		MAX	UNIT
SO-4 PA	CKAGE (DFG)					
		$R_{\theta JA}$ = 288.8°C/W, V_F = 1.4 V, T_J = 150°C, T_A = 25°C			300	mA
I _S	Safety limiting input current	R _{0JA} =288.8°C/W, V _{CEO} = 40 V, T _J = 150°C, T _A = 25°C			10.5	mA
		R _{0JA} = 288.8°C/W, V _{CEO} = 24 V, T _J =150°C, T _A = 25°C			17.5	mA
		R _{0JA} = 288.8°C/W, V _{CEO} = 15 V, T _J = 150°C, T _A = 25°C			28	mA
Ps	Safety limiting total power	R _{0JA} = 288.8°C/W, T _J = 150°C, T _A = 25°C			420	mW
T _S	Maximum safety temperature				135	°C

The I_S and P_S parameters represent the safety current and safety power respectively. The maximum limits of I_S and P_S should not be exceeded. These limits vary with the ambient temperature, T_{A} .

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The junction-to-air thermal resistance, $R_{\theta,JA}$, in the table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:

 $T_J = T_A + R_{\theta JA} \times P$, where P is the power dissipated in the device.

 $T_{J(max)} = T_S = T_A + R_{\theta JA} \times P_S$, where $T_{J(max)}$ is the maximum allowed junction temperature. $P_S = I_S \times V_I$, where V_I is the maximum input voltage.



7.7 Electrical Characteristics

All specifications are at T_A = 25 °C unless otherwise noted

	cations are at T _A = 25 °C unless PARAMETER	TEST CONDITIONS	GPN	MIN	TYP	MAX	UNIT
INPUT							
V _F	Input forward voltage	I _F = 5 mA	ISOM8110, ISOM8111, ISOM8112, ISOM8113		1.2	1.4	V
V _F	Input forward voltage	I _F = ±5 mA	ISOM8115, ISOM8116, ISOM8117, ISOM8118		1.2	1.5	V
I _R	Input reverse current	V _R = 5 V	ISOM8110, ISOM8111, ISOM8112, ISOM8113			10	uA
C _{IN}	Input capacitance	At 1 MHz, V _F = 0 V	ISOM8110, ISOM8111, ISOM8112, ISOM8113		35		pF
C _{IN}	Input capacitance	At 1 MHz, V _F = 0 V	ISOM8115, ISOM8116, ISOM8117, ISOM8118		6		pF
OUTPUT	,		1	•			
C _{CE}	Collector-emitter capacitance	1 MHz, V _F = 0 V	ISOM811x		12		pF
V _{CE(SAT)}	Collector-emitter saturation voltage	I _F = 20 mA, I _C = 1 mA	ISOM8110, ISOM8111, ISOM8112, ISOM8113			0.3	V
V _{CE(SAT)}	Collector-emitter saturation voltage	I _F = ±20 mA, I _C = 1 mA	ISOM8115, ISOM8116, ISOM8117, ISOM8118			0.3	٧
Dark Current	Collector dark current	V _{CE} = 20 V, I _F = 0 mA	ISOM811x			100	nA
I _{EC}	Reverse current	V _{EC} = 7 V, I _F = 0 mA	ISOM811x			100	μΑ
IC_OFF	OFF_state collector current	V _F = 0.7 V, V _{CE} = 48 V	ISOM8110, ISOM8111, ISOM8112, ISOM8113			10	uA
IC_OFF	OFF_state collector current	V _F = ±0.7 V, V _{CE} = 48 V	ISOM8115, ISOM8116, ISOM8117, ISOM8118			10	uA
CTR ⁽¹⁾							
			ISOM8110	55	130	195	%
			ISOM8115	55	130	195	%
			ISOM8111	80	180	290	%
CTR	Current Transfer Ratio	$I_C / I_F (T_A = 25^{\circ}C), I_F = 0.5 \text{ mA},$	ISOM8116	80	180	290	%
	Current transier Ratio	V _{CE} = 5 V	ISOM8112	135	300	480	%
			ISOM8117	135	300	480	%
			ISOM8113	195	440	710	%
			ISOM8118	195	440	710	%

All specifications are at T_A = 25 °C unless otherwise noted

_	PARAMETER	TEST CONDITIONS	GPN	MIN	TYP	MAX	UNIT
			ISOM8110	70	120	170	%
			ISOM8115	70	120	170	%
			ISOM8111	110	180	260	%
CTR	Current Transfer Ratio	$I_C / I_F (T_A = 25^{\circ}C), I_F = 2 \text{ mA},$	ISOM8116	110	180	260	%
CIK	Current Hansler Ratio	V _{CE} = 5V	ISOM8112	185	300	430	%
			ISOM8117	185	300	430	%
			ISOM8113	265	440	635	%
			ISOM8118	265	440	635	%
			ISOM8110	100	120	155	%
			ISOM8115	100	120	155	%
			ISOM8111	150	180	230	%
CTR	Current Transfer Ratio	I_{C}/I_{F} (T _A = 25°C), I_{F} = 5 mA,	ISOM8116	150	180	230	%
CIK	Current Hansler Ratio	V _{CE} = 5 V	ISOM8112	255	300	380	%
			ISOM8117	255	300	380	%
			ISOM8113	375	440	560	%
			ISOM8118	375	440	560	%

⁽¹⁾ CTR (%) = $(I_C / I_F) \times 100\%$



7.8 Switching Characteristics

All specifications are at T_A = 25 °C unless otherwise noted

	PARAMETER	TEST CONDITIONS	GPN	MIN TYP MAX	UNIT
AC					
	Rise time, see Figure 8-2	$V_{CC} = 10 \text{ V}, I_{C} = 2 \text{ mA}, R_{L} = 100 \Omega,$	ISOM8110	3.2	μs
t _r	and Figure 8-3	C _L = 50 pF	ISOM8113	1.1	μs
t.	Fall time, see Figure 8-2	$V_{CC} = 10 \text{ V}, I_{C} = 2 \text{ mA}, R_{L} = 100 \Omega,$	ISOM8110	4.0	μs
t _f	and Figure 8-3	C _L = 50 pF	ISOM8113	7.5	μs
			ISOM8110, ISOM8115	5.7	μs
T _{ON}	Turn on time, see Figure 8-2	$V_{CC} = 10 \text{ V}, I_C = 2 \text{ mA}, R_L = 100 \Omega,$	ISOM8111, ISOM8116	9.5	μs
	and Figure 8-3	C _L = 50 pF	ISOM8112, ISOM8117	8.1	μs
			ISOM8113, ISOM8118	20	μs
	Turn off time, see Figure 8-2 and Figure 8-3		ISOM8110, ISOM8115	3.6	μs
-		$V_{CC} = 10 \text{ V}, I_C = 2 \text{ mA}, R_L = 100 \Omega,$	ISOM8111, ISOM8116	2.3	μs
T _{OFF}		C _L = 50 pF	ISOM8112, ISOM8117	1.7	μs
			ISOM8113, ISOM8118	0.68	μs
t _s	Storage time; time required for the output waveform to change from 0% (100%) to 10% (90%) when input is turned on and back off, see, see Figure 8-3	V_{CC} = 5 V, I_F = 1.6 mA, R_L = 4.7 k Ω	ISOM811x	21	μs
			ISOM8110, ISOM8115	680	kHz
BW	Bandwidth, see Figure 8-4	V_{IN_DC} = 5 V, V_{IN_AC} = 1 Vpk, R_{IN} = 2 k Ω V _{CC} = 5 V, R_{LOAD} = 100 Ω ,	ISOM8111, ISOM8116	680	kHz
	and Figure 8-5	C _L = 50 pF, measured at V _{CE} –3dB sinewave	ISOM8112, ISOM8117	680	kHz
			ISOM8113, ISOM8118	680	kHz

Product Folder Links: ISOM8110

10

7.9 Typical Characteristics

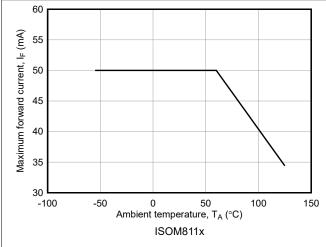


Figure 7-1. Maximum Forward Current vs Ambient Temperature

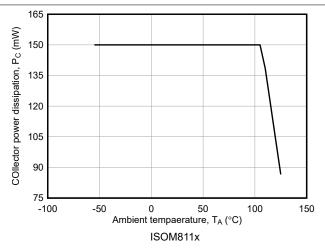


Figure 7-2. Maximum Collector Power Dissipation vs Ambient Temperature

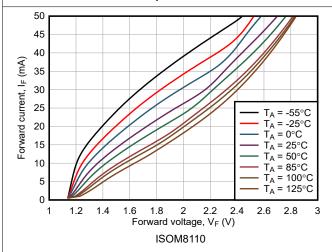


Figure 7-3. Forward Voltage vs Forward Current

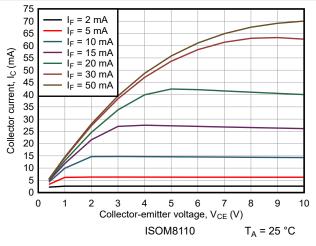


Figure 7-4. Collector Current vs Collector-Emitter Voltage

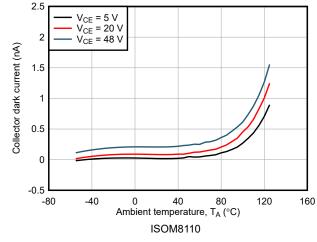


Figure 7-5. Collector Dark Current vs Ambient Temperature

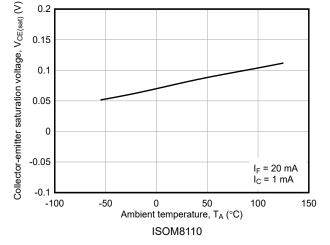
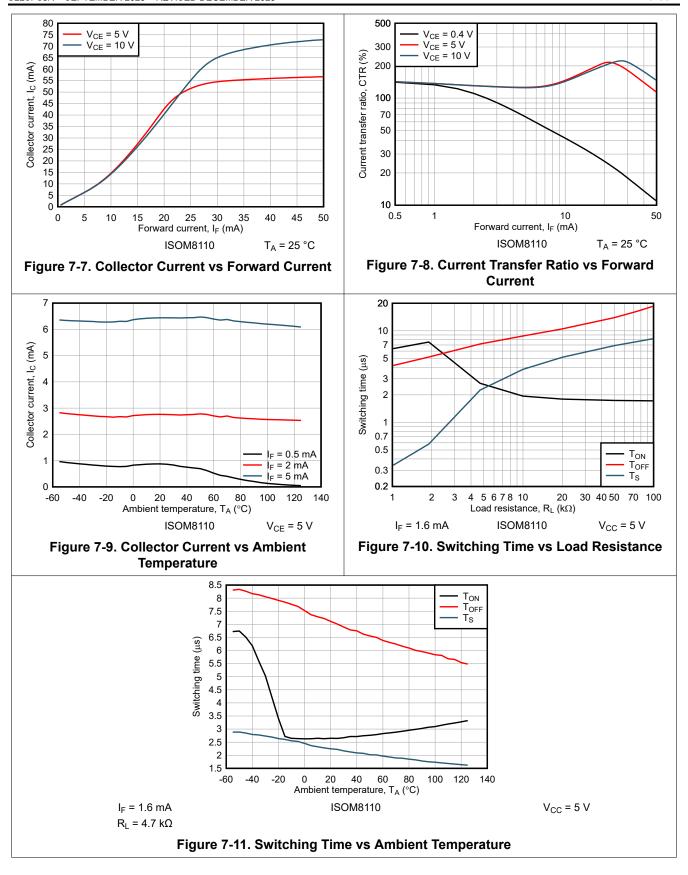


Figure 7-6. Collector-Emitter Saturation Voltage vs
Ambient Temperature





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8 Parameter Measurement Information

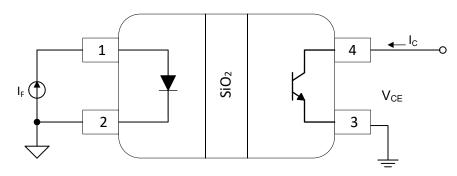


Figure 8-1. ISOM811x Test Circuit for CTR

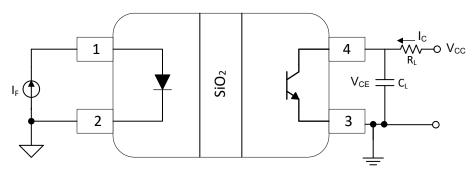


Figure 8-2. ISOM811x Test Circuit for Switching Timing

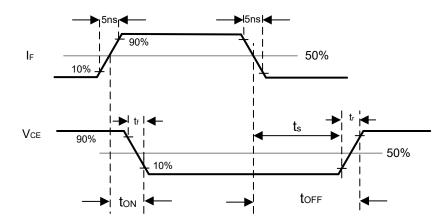


Figure 8-3. ISOM811x Switching Timing Waveforms

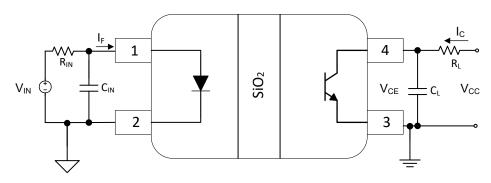


Figure 8-4. ISOM811(0-3) Test Circuit for Bandwidth

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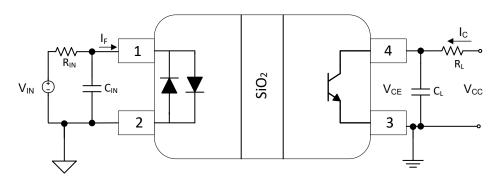


Figure 8-5. ISOM811(5-8) Test Circuit for Bandwidth

9 Detailed Description

9.1 Overview

The ISOM811x opto-emulators are pin-compatible, single-channel, drop-in replacements for many traditional optocouplers. While standard optocouplers use an LED as the input stage, ISOM811x uses an emulated LED as the input stage. The input and output stages are isolated by Tl's proprietary silicon dioxide-based (SiO₂) isolation barrier. This isolation technology makes ISOM811x resistant to the wear-out effects found in optocouplers that degrade performance with increasing temperature, forward current, and device age. Ordering options include four different ranges of current transfer ratio (CTR) and input options supporting uni-polar and bi-polar DC flow.

The ISOM811x family of devices isolate DC and bi-directional DC signals and offer performance, reliability, and flexibility advantages not available with traditional optocouplers.

The functional block diagram of ISOM811x devices are shown in Section 9.2. The input signal is transmitted across the isolation barrier using an on-off keying (OOK) modulation scheme. The transmitter sends a high-frequency carrier across the barrier that contains information on how much current is flowing through the input pins. The receiver demodulates the signal after advanced signal conditioning and produces the signal through the output stage. These devices also incorporate advanced circuit techniques to maximize bandwidth and minimize radiated emissions. Figure 9-3 shows conceptual details of how the OOK scheme works.

9.2 Functional Block Diagram

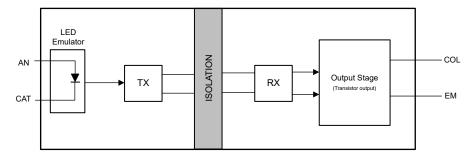


Figure 9-1. Conceptual Block Diagram of an Opto-emulator ISOM811(0-3)

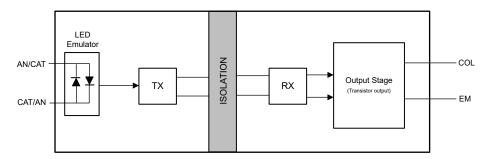


Figure 9-2. Conceptual Block Diagram of an Opto-emulator ISOM811(5-8)



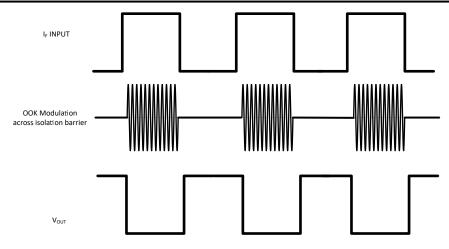


Figure 9-3. On-off Keying (OOK) Based Modulation Scheme

9.3 Feature Description

The ISOM811x devices isolate DC and bi-directional DC signals. ISOM811x has an open-collector output with four different CTR options. All devices support an isolation withstand voltage of $3750~V_{RMS}$ between side 1 and side 2.

9.4 Device Functional Modes

Table 9-1 lists the functional modes for the ISOM811x devices.

Table 9-1. Function Table

CTR ¹	PART NUMBER	Input type
100% to 155%	ISOM8110	DC
100% to 135%	ISOM8115	Bidirectional DC
150% to 230%	ISOM8111	DC
150% to 230%	ISOM8116	Bidirectional DC
255% to 380%	ISOM8112	DC
255% to 360%	ISOM8117	Bidirectional DC
2750/ to 5600/	ISOM8113	DC
375% to 560%	ISOM8118	Bidirectional DC

1. $I_F = 5 \text{ mA}, T_A = 25 ^{\circ}\text{C}, V_{CE} = 5 \text{ V}.$

10 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

10.1 Application Information

The ISOM811x devices are single-channel opto-emulators with LED-emulator input and transistor output. The devices use on-off keying modulation to transmit data across the isolation barrier. The input stage is isolated from the driver stage by Tl's proprietary silicon dioxide-based (SiO₂) isolation barrier which provides robust isolation. With wider temperature ratings than traditional optocouplers, ISOM811x opto-emulators can provide reliable signal isolation in harsh environments.

The ISOM811x devices are capable of sinking current when subjected to an external load being connected to the device. Like typical transistor output optocouplers, the output current will depend on the input current level (I_F) and the current transfer ratio (CTR). With multiple CTR options (100% - 560%), low input current, high bandwidth, low turn-off delay, low power consumption, and wider temperature range, ISOM811x devices are ideal for use in a variety of industries such as factory automation, building automation, e-mobility, automotive, avionics, medical, and power delivery.

10.1.1 Typical Application

ISOM811x opto-emulators are commonly used in the feedback control loops of isolated power supplies. These devices are used to solve the problem of feeding back current while isolating the primary and secondary domains to regulate the output voltage.

In power supplies, the output voltage is isolated from main input voltage using a transformer (for example: flyback converter). For analog power supply units, the controller IC is usually on the primary side of the transformer. For closed loop control, it is necessary to measure the output voltage on the secondary side and feed it back to the controller on the primary. The most common way of achieving this is using an opto-emulator such as ISOM811x, error amplifier (commonly TL431), and a voltage comparator to form a feedback loop across the isolation barrier

Figure 10-1 illustrates a typical isolated power supply. In this implementation, the output voltage is sensed by an error amplifier via the resistor divider (R1 and R2). Depending on the voltage level that it senses, the TL431 can drive the current of the ISOM811x higher or lower which is then compared to a voltage reference. The information is passed across the isolation barrier through ISOM811x to the primary side, where the PWM control circuit modulates the power stage to regulate the output voltage. The TL431 and ISOM811x play an important role for stable feedback and control loop.

The ISOM811x devices enable improvements in transient response, reliability, and stability as compared to commonly used optocoupler as the CTR is stable over wide temperature range of -55°C to 125°C providing a small, low-cost, highly reliable, and easy-to-design solution.



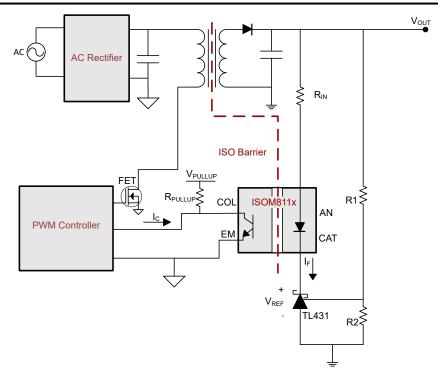


Figure 10-1. Typical Isolated Power Supply Application Using ISOM811x

10.1.1.1 Design Requirements

To design with ISOM811x devices, use the parameters listed in Table 10-1.

 PARAMETER
 VALUE

 Input forward current range, I_F
 0.5 mA (min), 50 mA (max)

 Current transfer ratio at I_F = 5 mA, CTR
 100% to 155%

 Collector current tolerance, I_C
 50 mA (max)

 Collector-emitter voltage (saturation), V_{CE(SAT)}
 0.3 V (max)

 Input forward voltage, V_F
 1.2 V (typ)

Table 10-1. Design Parameters

10.1.1.2 Detailed Design Procedure

This section presents the design procedure for using the ISOM811x opto-emulators. External components should be selected to operate ISOM811x within the *Recommended Operating Conditions*. The following recommendations on component selection focus on the design of a typical feedback control loop for an isolated flyback converter.

When using an optocoupler in a feedback control loop for an isolated power supply, many variables can affect how to properly use the optocoupler, including the output voltage of the power supply and the type of controller the feedback signal is being sent to. For this example, let's assume the output voltage of this power supply, V_{OUT} , is 5 V, and the PWM controller being used has an integrated error amplifier with a COMP pin that acts as the output of this amplifier.

Sizing R_{PULLUP}

The transistor output of ISOM811x will operate in active, saturation, reverse, and cut-off regions, just like a regular transistor. To ensure the output does not get damaged when it is saturated, the minimum value of R_{PULLUP} can be calculated for a given pull-up voltage, V_{PULLUP} , in Equation 1 below:



$$R_{PULLUP} > \frac{V_{PULLUP} - V_{CE(SAT)}}{I_{C(MAX)}} \tag{1}$$

For the example of a feedback loop application, we can calculate the minimum required value for R_{PULLUP} for a given V_{PULLUP} of 10 V, the max output voltage of the error amplifier ($V_{COMP(MAX)}$) of 2.5 V, and the max output current of the error amplifier is internally clamped at 1.6 mA. The equation to calculate R_{PULLUP} is shown in Equation 2 below:

$$R_{PULLUP} > \frac{V_{PULLUP} - V_{COMP(MAX)}}{I_{COMP(CLAMP)}} = \frac{10 V - 2.5 V}{1.6 mA} = 4.66 k\Omega$$
 (2)

Sizing R_{IN}

The input side of ISOM811x is current-driven. To limit the amount of current flowing into the AN pin, it is recommended that a series resistor, R_{IN} , is used in series with the input as shown in Figure 10-1.

Depending on how the ISOM811x device is being used, the value of R_{IN} can vary quite a bit. However, at a high level, to make sure the input does not get damaged, the minimum value of R_{IN} can be calculated for a given input voltage, V_{IN} , in Equation 3 below:

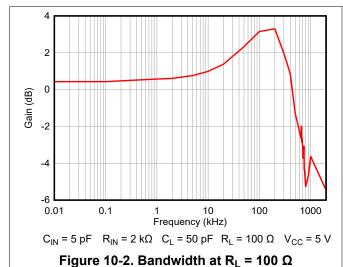
$$R_{IN} > \frac{V_{IN} - V_F}{I_{C(MAX)}} \tag{3}$$

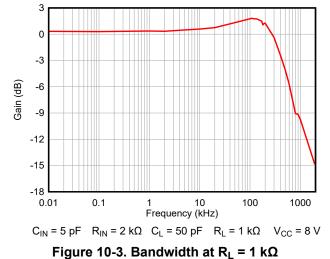
However, in the use case of a feedback loop, R_{IN} directly affects the mid-band gain of the loop. Let's assume the TL431 has been configured to give a reference voltage, V_{REF} , of 2.5 V and R_{PULLUP} is 5 k Ω . Equation 4 is used to calculate the maximum value of R_{IN} ensuring that V_{COMP} voltage on the primary side can be pulled to the saturation voltage of the ISOM811x, $V_{CE(SAT)}$.

$$R_{IN} < \frac{(V_{OUT} - V_{REF} - V_F) \times R_{PULLUP} \times CTR_{MIN}}{V_{PULLUP} - V_{CE}(SAT)} = \frac{(5 V - 2.5 V - 1.2 V) \times 5 k\Omega \times 100\%}{10 V - 0.3 V} = 670 \Omega$$
 (4)

10.1.1.3 Application Curves

The following curves show ISOM8110 bandwidth performance over different loading conditions where V_{IN} = 5 V_{DC} + 2 V_{PK} . See Figure 8-4 for setup details.

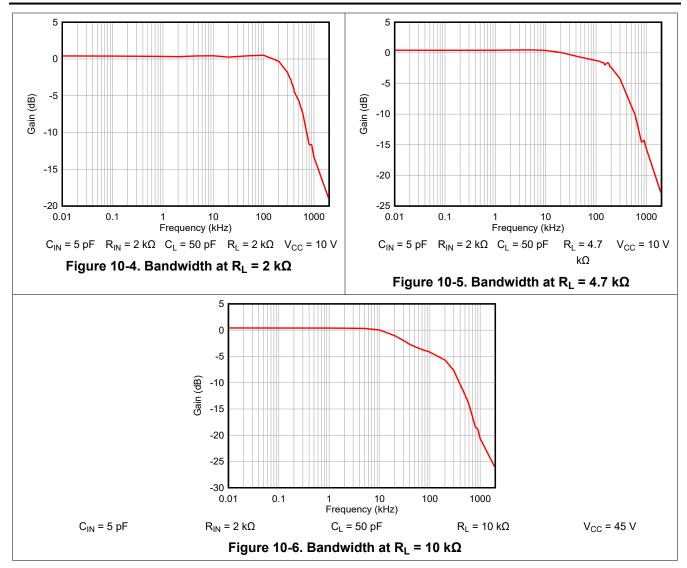




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10.2 Power Supply Recommendations

ISOM811x does not require a dedicated power supply to operate since there is no supply pin. Take care to not violate recommended I/O specifications for proper device functionality.

10.3 Layout

10.3.1 Layout Guidelines

- The device connections to ground should be tied to the PCB ground plane using a direct connection or two vias to help minimize inductance.
- The connections of capacitors and other components to the PCB ground plane should use a direct connection or two vias for minimum inductance.

10.3.2 Layout Example

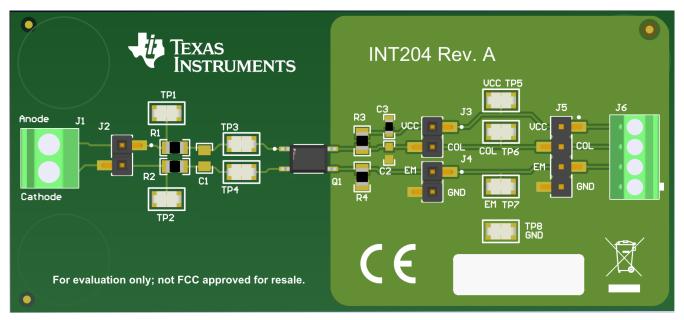


Figure 10-7. Layout Example of ISOM811x with a Single Layer Board

11 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



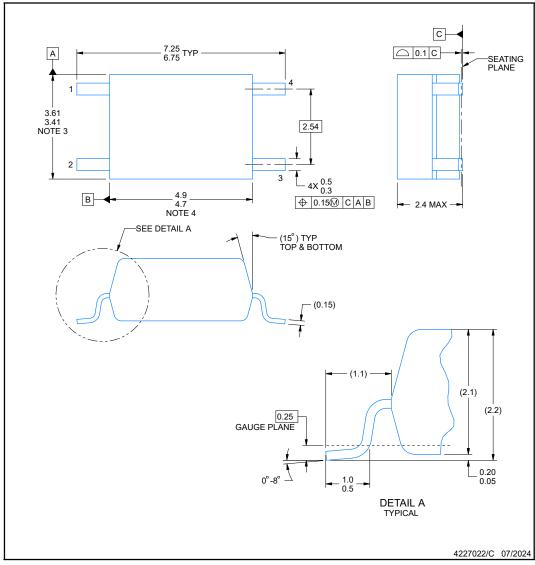
DFG0004A



PACKAGE OUTLINE

SOIC - 2.4 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

- All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
 This drawing is subject to change without notice.
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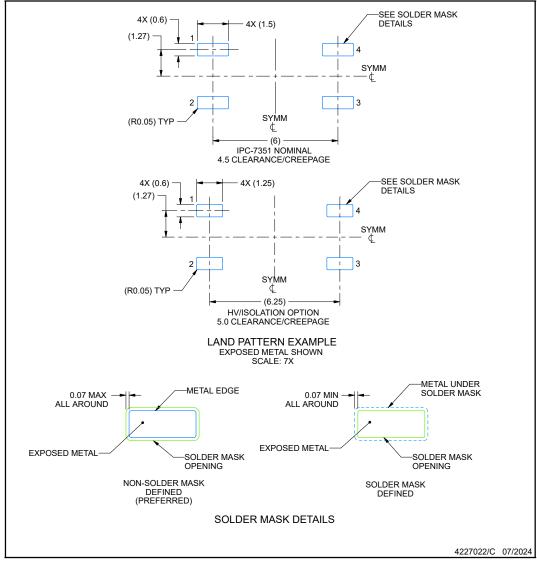


EXAMPLE BOARD LAYOUT

DFG0004A

SOIC - 2.4 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

- 5. Publication IPC-7351 may have alternate designs.6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



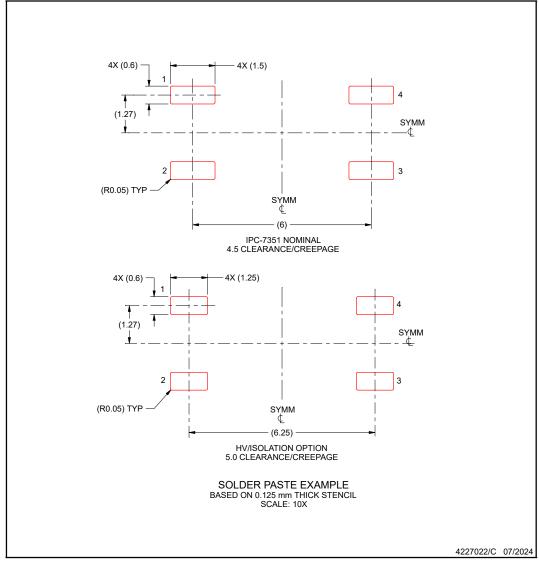


EXAMPLE STENCIL DESIGN

DFG0004A

SOIC - 2.4 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

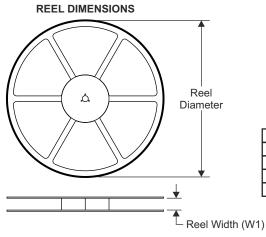
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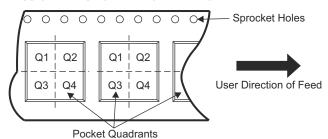
11.1 Tape and Reel Information



TAPE DIMENSIONS KO P1 BO W Cavity A0

A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers
	•

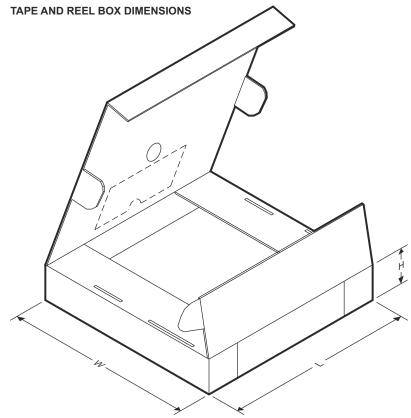
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISOM8110DFGR	SO-4	DFG	4	2000	330.0	12.4	8.0	3.8	2.7	12.0	12.0	Q1

Product Folder Links: ISOM8110





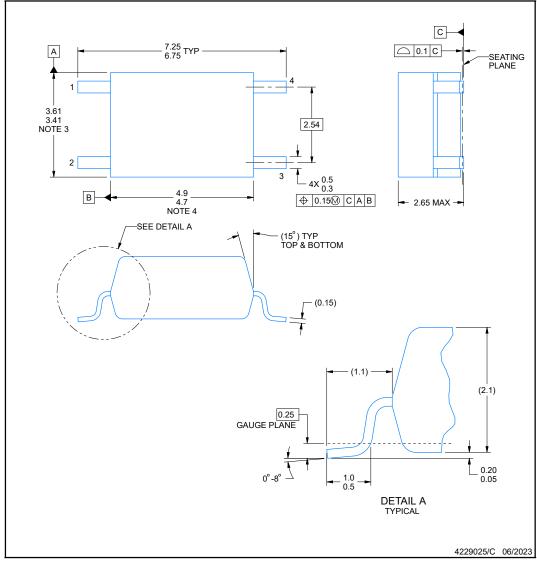
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISOM8110DFGR	SO-4	DFG	4	2000	356.0	356.0	35.0

PACKAGE OUTLINE

DFG0004A-C01

SOIC - 2.65 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

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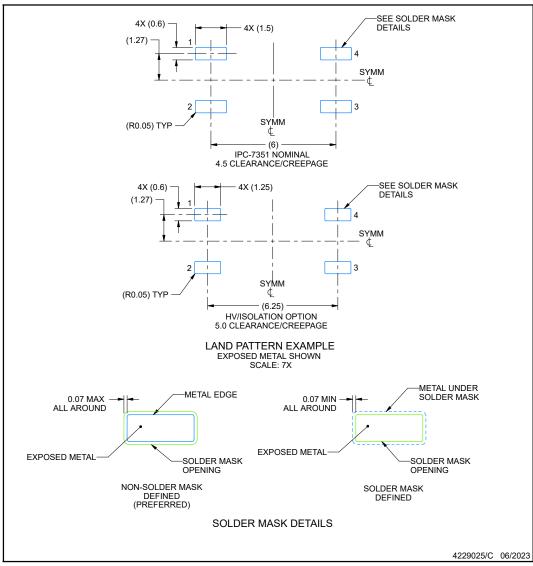


EXAMPLE BOARD LAYOUT

DFG0004A-C01

SOIC - 2.65 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

- 5. Publication IPC-7351 may have alternate designs.6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



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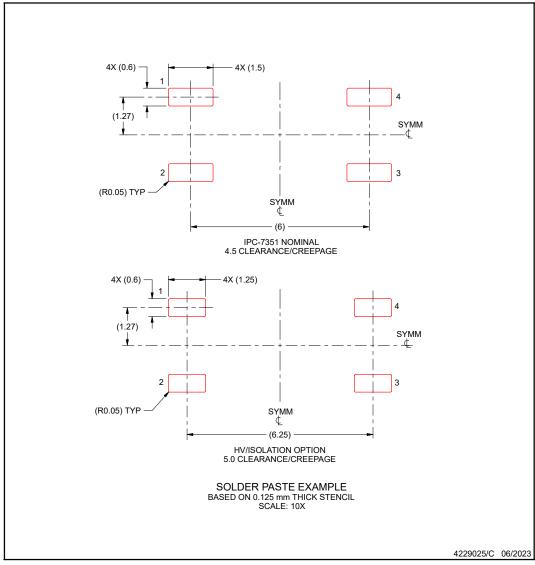


EXAMPLE STENCIL DESIGN

DFG0004A-C01

SOIC - 2.65 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)



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PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
							(6)				
ISOM8110DFGR	ACTIVE	SOIC	DFG	4	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-55 to 125	8110	Samples

(1) The marketing status values are defined as follows:

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NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

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- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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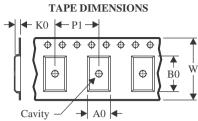
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PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

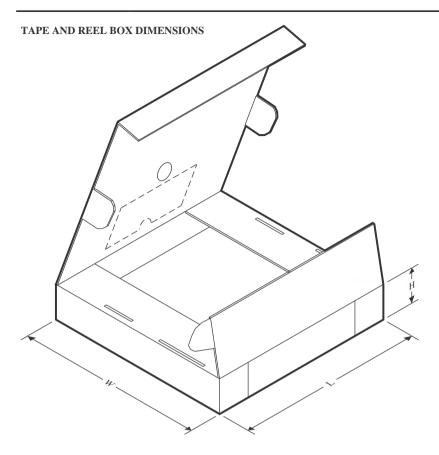


*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISOM8110DFGR	SOIC	DFG	4	2000	330.0	12.4	8.0	3.8	2.7	12.0	12.0	Q1

PACKAGE MATERIALS INFORMATION

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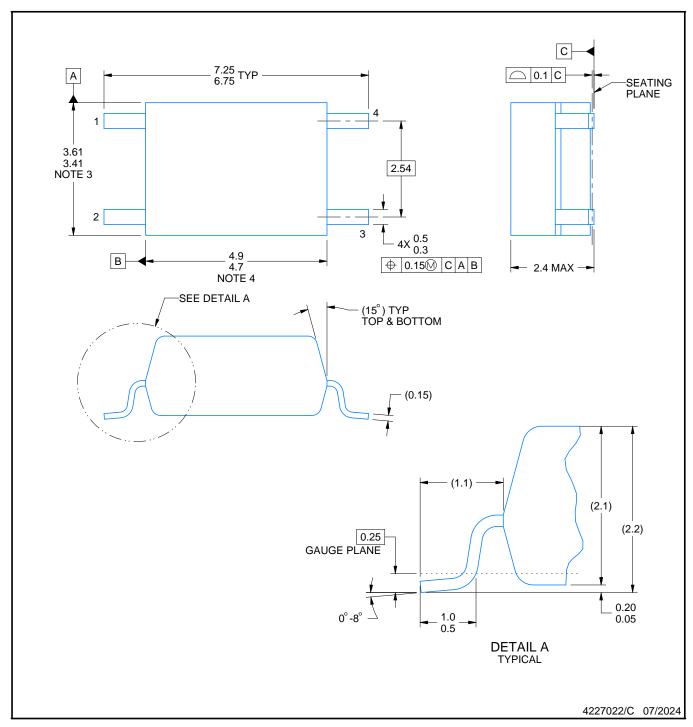


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISOM8110DFGR	SOIC	DFG	4	2000	356.0	356.0	35.0



SMALL OUTLINE INTEGRATED CIRCUIT

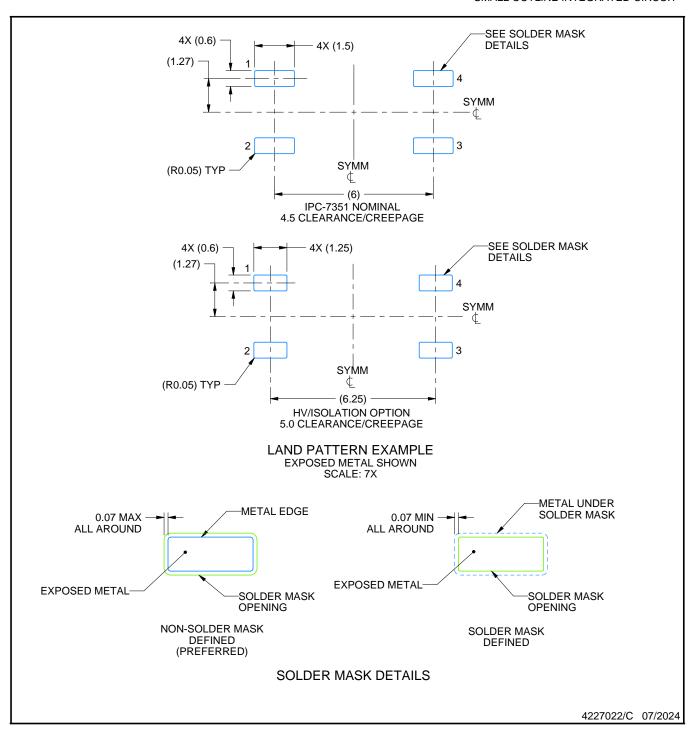


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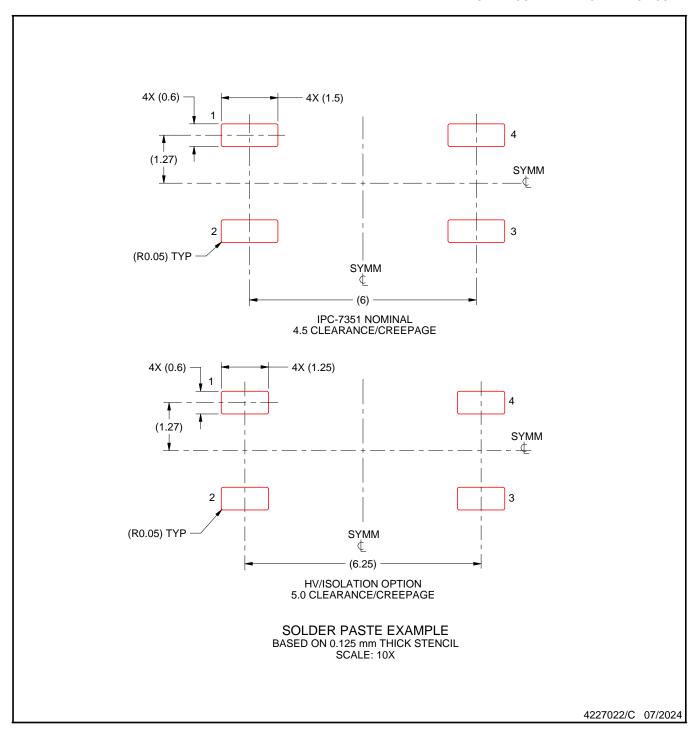


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SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

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