

FAN2012B 1.5A Low Voltage Current Mode Synchronous PWM Buck Regulator

Features

- 95% Efficiency, Synchronous Operation
- Adjustable Output Voltage from 0.8V to V_{IN}
- 4.5V to 5.5V Input Voltage Range
- Up to 1.5A Output Current
- Fixed Frequency 1.3 MHz PWM Operation
- 100% Duty Cycle Low Dropout Operation
- Soft Start
- Excellent Load Transient Response
- 3x3mm 6-lead MLP Package

Applications

- Hard Disk Drive
- Set Top Box
- Point of Load Power
- Notebook Computers
- Communications Equipment

Description

The FAN2012B is a high-efficiency, low-noise synchronous PWM current mode DC-DC converter, designed for low-voltage applications. It provides up to 1.5A continuous load current from the 4.5V to 5.5V input. The output voltage is adjustable over a wide range of 0.8V to V_{IN} by means of an external voltage divider.

The FAN2012B has an "Enable Input" and the device can be put in the shutdown mode, in which the ground current falls below 1μ A.

A current mode control loop with a f ast transient response ensures excellent line and load regulation. The fixed 1.3MHz switching frequency enables the user to choo se a small, inexpensive external inductor and capacitor. Filtering is also easily accomplished with very small components.

Protection features include input under-v oltage lockout, short circuit protection and thermal shutdown. Soft-start limits in-rush current during start-up conditions.

The device is available in a 3x3mm 6-lead MLP package, making it possible to build a 1.5 A complete DC/DC converter in a tiny space on the PCB.

Typical Application









Pin Definitions

Pin No.	Pin Name	Pin Description
P1	AGND	Analog Ground. P1 must be soldered to the PCB ground.
1	FB	Feedback Input. Adjustable voltage option, connect this pin to the resistor divider.
2	PGND	Power Ground. This pin is connected to the internal MOSFET switches. This pin must be externally connected to AGND.
3	SW	Switching Node. This pin is connected to the internal MOSFET switches.
4	PVIN	Supply Voltage Input. This pin is connected to the internal MOSFET switches.
5	V _{IN}	Supply Voltage Input.
6	EN	Enable Input. Logic high enables the chip and logic low disables the chip, reducing the supply current to less than $1\mu A$. Do not float this pin.

Absolute Maximum Ratings (Note1)

Parameter		Min	Max	Unit
V _{IN}		-0.3	7	V
PV _{IN} and any other pin		-0.3	V _{IN}	V
Thermal Resistance-Junction to Tab (θ_{JC}), 3mmx3mm 6-lead MLP (Not	Thermal Resistance-Junction to Tab (θ_{JC}), 3mmx3mm 6-lead MLP (Note 2)		8	°C/W
Lead Soldering Temperature (10 seconds)			260	°C
Storage Temperature		-65	150	°C
Junction Temperature		-40	150	°C
Electrostatic Discharge (ESD) Protection Level (Note 3)	HBM	4		kV
	CDM	2		

Recommended Operating Conditions

Parameter	Min	Тур	Мах	Unit
Supply Voltage Range	4.5		5.5	V
Output Voltage Range, Adjustable Version	0.8		V _{IN}	V
Output Current			1500	mA
Inductor (Note 4)		3.3		μH
Input Capacitor (Note 4)		10		μF
Output Capacitor (Note 4)		4 x 10		μF
Operating Ambient Temperature Range	0		+85	°C

Notes:

- Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This
 is a stress rating only and functional operation of the device at these or any other conditions above those indicated
 in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for
 extended periods may affect device reliability. Absolute maximum ratings apply individually only, not in combination.
 Unless otherwise specified, all other voltages are referenced to AGND.
- Junction to ambient thermal resistance, θ_{JA}, is a strong function of PCB material, board thickness, thickness and number of copper planes, number of via used, diameter of via used, available copper surface, and attached heat sink characteristics.
- 3. Using Mil Std. 883E, method 3015.7(Human Body Model) and EIA/JESD22C101-A (Charge Device Model).
- 4. Refer to the applications section for further details.

Electrical Characteristics

 V_{IN} = 4.5V to 5.5V, V_{OUT} = 1.2V, I_{OUT} = 200mA, C_{IN} = 10µF, C_{OUT} = 4x10µF, L = 3.3µH, T_A = 0°C to +85°C, unless otherwise noted. Typical values are at T_A = 25°C.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
V _{IN}	Input Voltage		4.5		5.5	V
Ι _Q	Quiescent Current	I _{OUT} = 0mA		7	10	mA
V	UV/LO Threshold	V _{IN} Rising	3.5	3.7	3.9	V
V _{UVLO}	UVLO Threshold	Hysteresis		150		mV
R _{ON_PMOS}	PMOS On Resistance	$V_{IN} = V_{GS} = 5V$		150	290	mΩ
R _{ON_NMOS}	NMOS On Resistance	$V_{IN} = V_{GS} = 5V$		150	290	mΩ
I _{LIMIT}	P-Channel Current Limit	4.5V < V _{IN} < 5.5V	2.2	2.6	3.5	А
		Rising Temperature		150		°C
	Over-Temperature Protection	Hysteresis		20		°C
f _{SW}	Switching Frequency		1000	1300	1600	kHz
R _{LINE}	Line Regulation	$V_{IN} = 4.5 \text{ to } 5.5 \text{V}, I_{OUT} = 100 \text{mA}$		0.16		%/V
R _{LOAD}	Load Regulation	$0mA \le I_{OUT} \le 1500mA$		0.2	0.5	%
N	Output Voltage During Load	I _{OUT} from 700mA to 100mA			5	%
V _{OUT}	Transition ⁽⁵⁾	I _{OUT} from 100mA to 700mA	-5			%
I _{LEAK}	Reverse Leakage Current into Pin SW	V_{IN} = Open, EN = GND, V_{SW} = 5.5V		0.1	1.0	μA
V _{REF}	Reference Voltage			0.8		V
V _{OUT}	Output Voltage Accuracy	V_{IN} = 4.5 to 5.5V, 0mA $\leq I_{OUT} \leq$ 1500mA	-2		2	%

Note:

5. Please refer to the load transient response test waveform shown in Figure 3.

Additional Electrical Characteristics

 V_{IN} = 4.5V to 5.5V and T_A = 0°C to +85°C, unless otherwise noted. Typical values are at T_A = 25°C.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
I _{SD}	Shutdown Mode Supply Current	VEN=0V		0.1	1.0	μA
I _{BIAS}	EN Bias Current				0.1	μA
V_{ENH}	EN High Voltage		1.3			V
V _{ENL}	EN Low Voltage				0.4	V





Typical Performance Characteristics

 T_A = 25°C, C_{IN} = 10µF, C_{OUT} = 40µF, L = 3.3µH, V_{IN} = 5V, unless otherwise noted.



Switching Frequency vs. Ambient Temperature







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Block Diagram





Detailed Operation Description

The FAN2012B is a step-down pulse width modulated (PWM) current mode converter with a typical switching frequency of 1.3 MHz. At the beginning of each clock cycle, the P-channel transistor is turned on. The inductor current ramps up and is monitored via an inter nal circuit. The P-channel switch is turned off when the sensed current causes the PWM comparator to trip when the output voltage is in regula tion or when the inducto r current reaches the current limit (set internally to typically 2600mA). After a minimum dead time, the N-channel transistor is turned on and the inductor current ramps down. As the clock cycle is completed, the N-channel switch is turned off and the next clock cycle starts. The duty cycle is solely given by the ratio of output voltage and input v oltage. Therefore, the converter runs with a minimum duty cycle when output voltage is at minimum and input voltage is at maximum.

100% Duty Cycle Operation

As the input voltage approaches the output voltage and the duty cycle exceeds the typical 95%, the converter turns the P-channel transistor continuously on. In this mode the output voltage is equal to the input voltage minus the voltage drop across the Pchannel transistor:

 V_{OUT} = $V_{IN} - I_{LOAD} \times (R_{DS_ON} + R_L)$, where

 R_{DS_ON} = P-channel switch ON resistance I_{LOAD} = Output current R_L = Inductor DC resistance

UVLO and Soft Start

The reference and the circuit remain reset until the V_{IN} crosses its UVLO threshold.

The FAN2012B has an internal soft-start circuit that limits the inrush current during startup. This prevents possible voltage drops of the input v oltage and eliminates the output voltage overshoot. The soft-start is im plemented as a digital circuit increasing the switch current in four steps to the P-channel current limit (2600mA). Typical start-up time for a 40μ F output capacitor and a load current of 1500mA is 800µs.

Short Circuit Protection

The switch peak current is limit ed cycle by cycle to a typical value of 2600mA. In the event of an output voltage short circuit, the device operates with a frequency of 400kHz and minimum duty cycle, therefore the average input curre nt is typically 350mA.

Thermal Shutdown

When the die temperature exceeds 150° C, a reset occurs and will remain in effect until the die cools to 130° C, at that time the circuit will be allowed to restart.

Applications Information

Setting the Output Voltage

The internal reference is 0.8V. The output is divided down by a voltage divider, R1 and R2 to the FB pin. The output voltage is:

$$V_{\text{OUT}} = V_{\text{REF}} \left(1 + \frac{R_1}{R_2} \right)$$

According to this equation, a nd assuming desired output voltage of 1.5096V, and given R2 = 10K Ω , the calculated value of R1 is 8.87K Ω .

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Inductor Selection

The inductor parameters directly related to device performances are saturation current and DC resistance. FAN2011/2012B operates with a typical inductor value of 3.3μ H. The lower the DC resistance, the higher the efficiency. For saturation current, the inductor should be rated higher than the maximum load current plus half of the inductor ripple current that is calculated as follows:

$$\Delta I_{L} = V_{OUT} \times \frac{1 - (V_{OUT} / V_{IN})}{L \times f}$$

where:

 ΔI_L = Inductor Ripple Current

f = Switching Frequency

L = Inductor Value

Recommended inductors are suggested in Table 1:

Inductor Value	Vendor	Part Number
3.3µH	Panasonic	ELL6PM3R3N
3.3µH	Murata	LQS66C3R3M04

Table 1: Recommended Inductors

Capacitors Selection

For best performances, a low ESR input capacitor is required. A ceramic capacitor of at least $10\mu F$, placed as close to the V_{IN} and AGND pins of the de vice is recommended. The output capacitor determines the output ripple and the transient response.

Capacitor Value	Vendor	Part Number
10µF	Taiyo Yuden	JMK212BJ106MG
		JMK316BJ106KL
	TDK	C2012X5ROJ106K
		C3216X5ROJ106M
	Murata	GRM32ER61C106K

Table 2: Recommended Capacitors

PCB Layout Recommendations

The inherently high peak currents and switching frequency of power supplies require a careful PCB layout design. Therefore, use wide tr aces for high current paths and place the input capacitor, the inductor, and the output capacitor as close as possible to the integrated circuit terminals. In order to minimize voltage stress to the device resulting from ever present switching spikes, use an input bypass capacitor with low ESR. Use of an external Schottky diode with its an ode connected to SW node and cathode connected to PV_{IN} further reduces switching spikes. Note that the peak amplitude of the switching spikes depends upon the load current; the higher the load current, the higher the switching spikes.

The resistor divider that sets the output voltage should be routed away from the inductor to avoid RF coupling. The ground plane at the bottom side of the PCB acts as an electromagnetic shield to reduce EMI. The recommended PCB layout is shown below in Figure 5.



Figure 5. Recommended PCB Layout



Ordering Information

Product Number	Output Voltage	Package Type	Order Code
FAN2012B	Adjustable	3x3mm 6-Lead MLP	FAN2012BMPX

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