

## **DDR Termination Regulator**

### **General Description**

The RTQ2508 is a sink/source tracking termination regulator. It is specifically designed for low-cost and low-external component count systems. The RTQ2508 possesses a high speed operating amplifier that provides fast load transient response and only requires a minimum 10 $\mu$ F x 3 ceramic output capacitor. The RTQ2508 supports remote sensing functions and all features required to power the DDRIII and Low Power DDRIII / DDRIV VTT bus termination according to the JEDEC specification. In addition, the RTQ2508 provides an open-drain PGOOD signal to monitor the output regulation and an EN signal that can be used to discharge VTT during S3 (suspend to RAM) for DDR applications.

The RTQ2508 is available in the thermal efficient package, WDFN-10L 3x3.

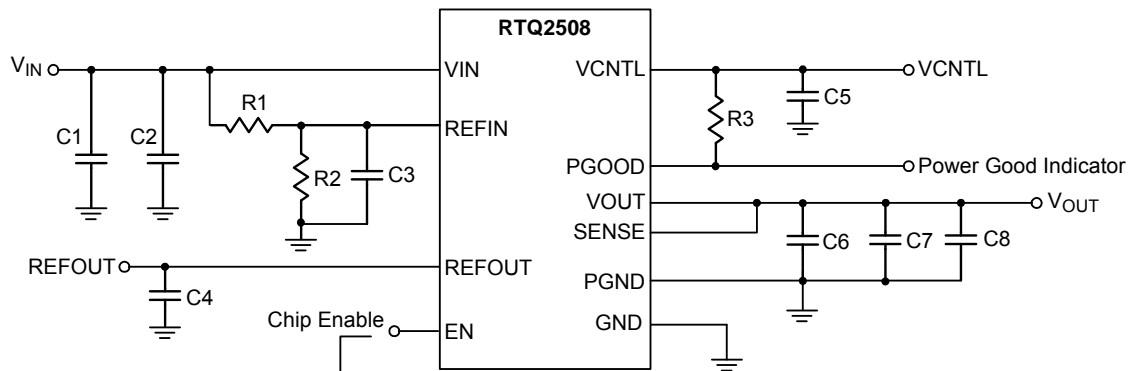
### **Features**

- VIN Input Voltage Range : 1.1V to 3.5V
- VCNTL Input Voltage Range : 2.9V to 5.5V
- Support Ceramic Capacitors
- Power Good Indicator
- 10mA Source/Sink Reference Output
- Meet DDRI, DDRII JEDEC Spec
- Support DDRI, DDRII, DDRIII and Low Power DDRIII/DDRIV VTT Applications
- Soft-Start Function
- UVLO and OCP Protection
- Thermal Shutdown

### **Applications**

- Notebook/Desktop/Server
- Telecom/Datacom, GSM Base Station, LCD-TV/PDP-TV, Copier/Printer, Set-Top Box

## **Simplified Application Circuit**

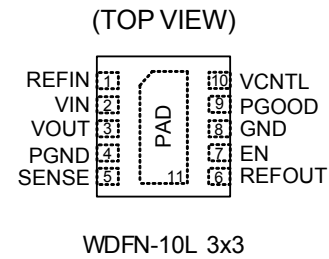


## Ordering Information

RTQ2508 □ □ □

- Pin 1 Orientation\*\*\*  
(2) : Quadrant 2, Follow EIA-481-D
- Package Type  
QW : WDFN-10L 3x3 (W-Type)
- Lead Plating System  
G : Green (Halogen Free and Pb Free)

## Pin Configuration



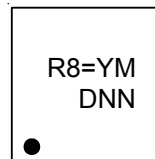
Note :

\*\*\*Empty means Pin1 orientation is Quadrant 1

Richtek products are :

- ▶ RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- ▶ Suitable for use in SnPb or Pb-free soldering processes.

## Marking Information



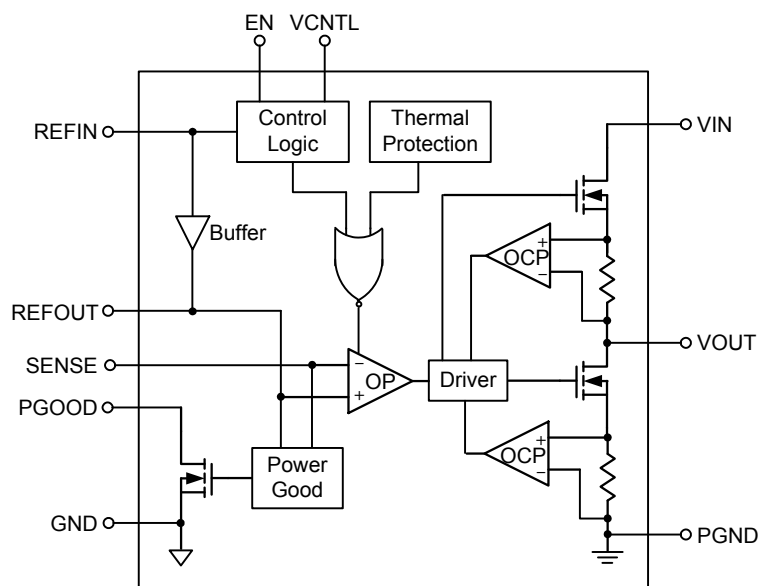
R8= : Product Code

YMDNN : Date Code

## Functional Pin Description

Pin No.	Pin Name	Pin Function
1	REFIN	Reference input.
2	VIN	Power input of the regulator.
3	VOUT	Power output of the regulator.
4	PGND	Power ground of the regulator.
5	SENSE	Voltage sense input for the regulator. Connect to positive terminal of the output capacitor or the load.
6	REFOUT	Reference output. Connect to GND through a 0.1μF ceramic capacitor.
7	EN	Enable control input. For DDR VTT application, connect EN to SLP_S3. For other applications, use EN as the ON/OFF function.
9	PGOOD	Power good open-drain output. Connect a pull-up resistor between this pin and VCNTL pin.
10	VCNTL	Control voltage input. Connect this pin to the 3.3V or 5V power supply. A ceramic decoupling capacitor with a value between 1μF and 4.7μF is required.
8	GND	Analog ground. Connect to negative terminal of the output capacitor.
11 (Exposed Pad)	PAD	Exposed pad. The exposed pad is internally unconnected and must be soldered to a large PGND plane. Connect this PGND plane to other layers with thermal vias to help dissipate heat from the device.

## Functional Block Diagram



## Operation

The RTQ2508 is a linear sink/source DDR termination regulator with current capability up to 3A. The RTQ2508 builds in a high-side N-MOSFET which provides current sourcing and a low-side N-MOSFET which provides current sinking. All the control circuits are supplied by the power VCNTL. In normal operation, the error amplifier OP adjusts the gate driving voltage of the power MOSFET to achieve SENSE voltage well tracking the REFIN voltage.

Both the source and sink currents are detected by the internal sensing resistor, and the OCP function will work to limit the current to a designed value when overload happens. Furthermore, the current will be folded back to be one half if VOUT is out of the power good window.

### Buffer

This function provides REFOUT output equal to  $V_{REFIN}$  with 10mA source/sink current capability.

### Power Good

When the SENSE voltage is in the power good window and lasts for a certain delay time, then the PGOOD pin will be high impedance and the PGOOD voltage will be pulled high by the external resistor.

### Over-Current Protection

The device continuously monitors the output current to protect the pass transistor against abnormal operations. The current limit ( $I_{LIM}$ ) level reduces by one-third when the output voltage is not within the powergood window. This reduction is a non-latch protection.

### Control Logic

This block includes VCNTL UVLO, REFIN UVLO and Enable/Disable functions, and provides logic control to the whole chip.

### Thermal Protection

Both the high-side and low-side power MOSFETs will be turned off when the junction temperature is higher than typically 160°C, and be released to normal operation when junction temperature falls below 120°C typically.

**Absolute Maximum Ratings** (Note 1)

• Supply Voltage, VIN, VCNTL	-----	–0.3V to 6V
• Input Voltage, EN, REFIN, SENSE	-----	–0.3V to 6V
• Output Voltage, VOUT, REFOUT, PGOOD	-----	–0.3V to 6V
• Power Dissipation, PD @ TA = 25°C		
WDFN-10L 3x3	-----	2.5W
• Package Thermal Resistance (Note 2)		
WDFN-10L 3x3, $\theta_{JA}$	-----	40°C/W
WDFN-10L 3x3, $\theta_{JC}$	-----	7.5°C/W
• Lead Temperature (Soldering, 10 sec.)	-----	260°C
• Junction Temperature	-----	150°C
• Storage Temperature Range	-----	–65°C to 150°C
• ESD Susceptibility (Note 3)		
HBM (Human Body Model)	-----	2kV

**Recommended Operating Conditions** (Note 4)

• Control Input Voltage, VCNTL	-----	2.9V to 5.5V
• Supply Input Voltage, VIN	-----	1.1V to 3.5V
• Junction Temperature Range	-----	–40°C to 125°C

**Electrical Characteristics**

(VIN = 1.5V, VEN = VCNTL = 3.3V, VREFIN = VSENSE = 0.75V, COUT = 10μF x 3, TJ = –40°C to 85°C, unless otherwise specified)

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
<b>Supply Current</b>						
VCNTL Supply Current	IVCNTL	VEN = VCNTL, no load	--	0.7	1	mA
VCNTL Shutdown Current	ISHDN_VCNTL	VEN = 0V, VREFIN = 0V, no load	--	65	80	μA
		VEN = 0V, VREFIN > 0.4V, no load	--	200	400	
VIN Supply Current	IVIN	VEN = VCNTL, no load	--	1	50	μA
VIN Shutdown Current	ISHDN_VIN	VEN = 0V, no load	--	0.1	50	μA
<b>Output</b>						
VTT Output Voltage	VOUTO	VIN = 1.5V, VREFIN = 0.75V, IOUT = 0A	--	0.75	--	V
		VIN = 1.35V, VREFIN = 0.675V, IOUT = 0A	--	0.675	--	
		VIN = 1.2V, VREFIN = 0.6V, IOUT = 0A	--	0.6	--	
VTT Output Voltage Offset	VOUT_OS	IOUT < ±2A, VLDOIN = 1.5V, VOUT_OS = VOUT – VOUTO	–25	--	25	mV
		IOUT < ±2A, VLDOIN = 1.35V, VOUT_OS = VOUT – VOUTO	–25	--	25	
		IOUT < ±2A, VLDOIN = 1.2V, VOUT_OS = VOUT – VOUTO	–25	--	25	

Parameter		Symbol	Test Conditions	Min	Typ	Max	Unit
VOUT Source Current Limit		I <sub>LIM_VOUT_SR</sub>	VOUT in PGOOD window	3	4.5	--	A
VOUT Sink Current Limit		I <sub>LIM_VOUT_SK</sub>	VOUT in PGOOD window	3	4.5	--	A
VOUT Discharge Resistance		R <sub>DISCHARGE</sub>	V <sub>REFIN</sub> = 0V, V <sub>OUT</sub> = 0.3V, V <sub>EN</sub> = 0V	--	18	25	Ω
Power Good Comparator							
PGOOD Threshold		V <sub>TH_PGOOD</sub>	V <sub>SENSE</sub> lower threshold with respect to REFOUT	--	−20	--	%
			V <sub>SENSE</sub> upper threshold with respect to REFOUT	--	20	--	
			PGOOD hysteresis	--	5	--	
PGOOD Start-Up Delay		T <sub>PGDELAY1</sub>	Start-up rising delay, V <sub>SENSE</sub> within PGOOD range	--	2	--	ms
Output Low Voltage		V <sub>LOW_PGOOD</sub>	I <sub>PGOOD</sub> = 4mA	--	--	0.4	V
PGOOD Falling Delay		T <sub>PGDELAY2</sub>	Falling delay, V <sub>SENSE</sub> is out of PGOOD range	--	10	--	μs
Leakage Current		I <sub>LEAKAGE_PGOOD</sub>	V <sub>SENSE</sub> = V <sub>REFIN</sub> (PGOOD high impedance), V <sub>PGOOD</sub> = V <sub>CNTL</sub> + 0.2V	--	--	1	μA
REFIN and REFOUT							
REFIN Input Current		I <sub>REFIN</sub>	V <sub>EN</sub> = V <sub>CNTL</sub>	--	--	1	μA
REFIN Voltage Range		V <sub>REFIN</sub>		0.5	--	1.8	V
REFIN Under-Voltage Lockout		V <sub>UVLO_REFIN</sub>	REFIN rising	360	390	420	mV
			Hysteresis	--	20	--	
REFOUT Voltage Tolerance to V <sub>REFIN</sub>		V <sub>TOL_REFOUT</sub>	−10mA < I <sub>REFOUT</sub> < 10mA, V <sub>REFIN</sub> = 0.75V	−15	--	15	mV
			−10mA < I <sub>REFOUT</sub> < 10mA, V <sub>REFIN</sub> = 0.675V	−15	--	15	
			−10mA < I <sub>REFOUT</sub> < 10mA, V <sub>REFIN</sub> = 0.6V	−15	--	15	
REFOUT Source Current Limit		I <sub>LIM_REFOUT_SR</sub>	V <sub>REFOUT</sub> = 0V	10	40	--	mA
REFOUT Sink Current Limit		I <sub>LIM_REFOUT_SK</sub>	V <sub>REFOUT</sub> = REFIN + 1V	10	40	--	mA
UVLO/EN							
UVLO Threshold		V <sub>UVLO_VCNTL</sub>	Rising	2.5	2.7	2.85	V
			Hysteresis	--	120	--	mV
EN Input Voltage	Logic-High	V <sub>IN_H</sub>		1.7	--	--	V
	Logic-Low	V <sub>IN_L</sub>		--	--	0.3	
Thermal Shutdown							
Thermal Shutdown Threshold		T <sub>SD</sub>	Shutdown temperature (Note 5)	--	160	--	°C
			Hysteresis (Note 5)	--	15	--	

- Note 1.** Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
- Note 2.**  $\theta_{JA}$  is measured under natural convection (still air) at  $T_A = 25^{\circ}\text{C}$  with the component mounted on a high effective-thermal-conductivity four-layer test board on a JEDEC 51-7 thermal measurement standard.  $\theta_{JC}$  is measured at the exposed pad of the package.
- Note 3.** Devices are ESD sensitive. Handling precaution is recommended.
- Note 4.** The device is not guaranteed to function outside its operating conditions.
- Note 5.** Guarantee by design.

## Typical Application Circuit

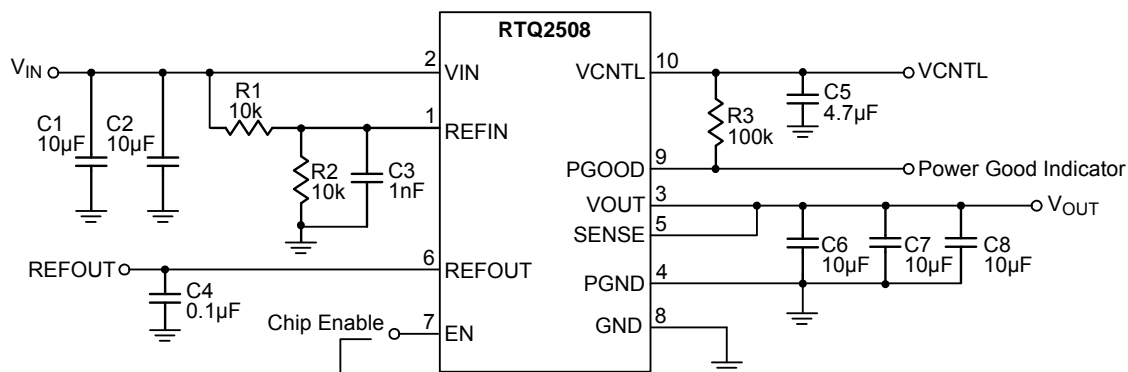
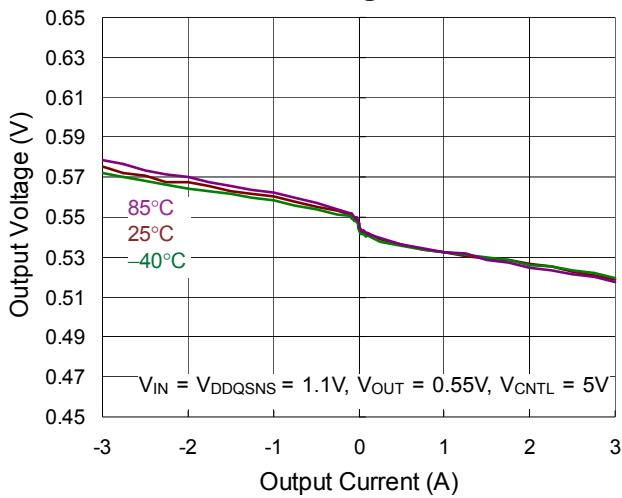


Table 1. Recommended External Components

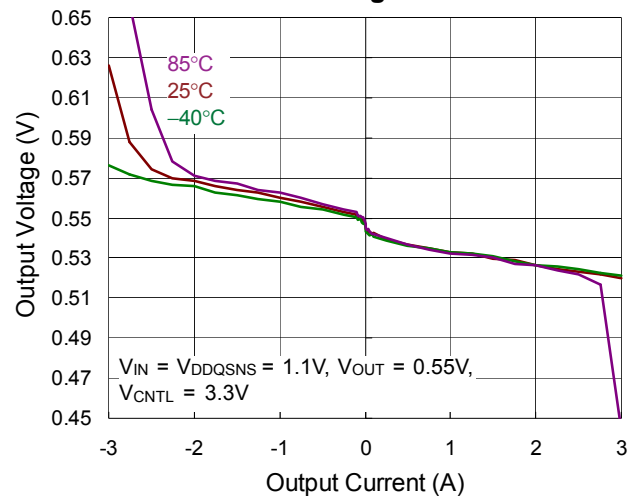
Component	Description	Vendor P/N
C1, C2, C6, C7, C8	10 $\mu$ F, 6.3V, X7R, 0805	GRM21BR70J106KE76L (Murata) CGA4J1X7R0J106K125AC (TDK)
C3	1nF, 50V, X7R, 0603	GCD188R71H102KA01D (Murata) CGA3E2X7R1H102K080AA (TDK)
C4	0.1 $\mu$ F, 16V, X7R, 0603	GCJ188R71C104KA01D (Murata)
C5	4.7 $\mu$ F, 6.3V, X5R, 0603	GRT188R60J475ME01D (Murata) CGB3B3X5R0J475M055AB(TDK)

## Typical Operating Characteristics

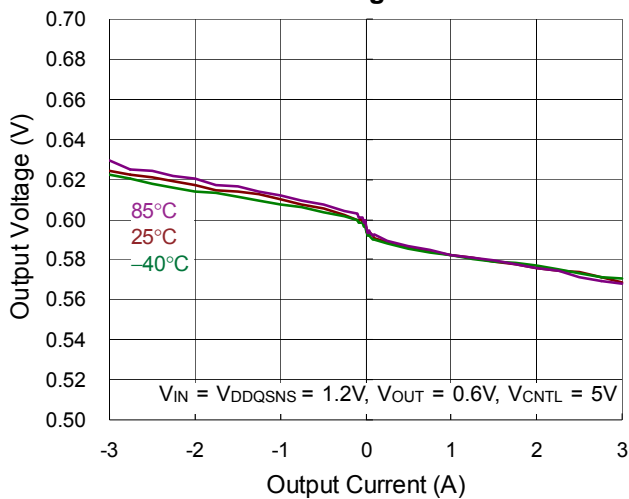
### Load Regulation



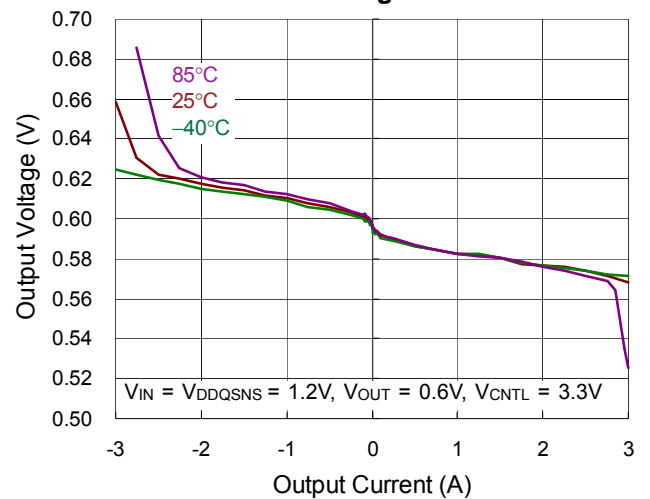
### Load Regulation



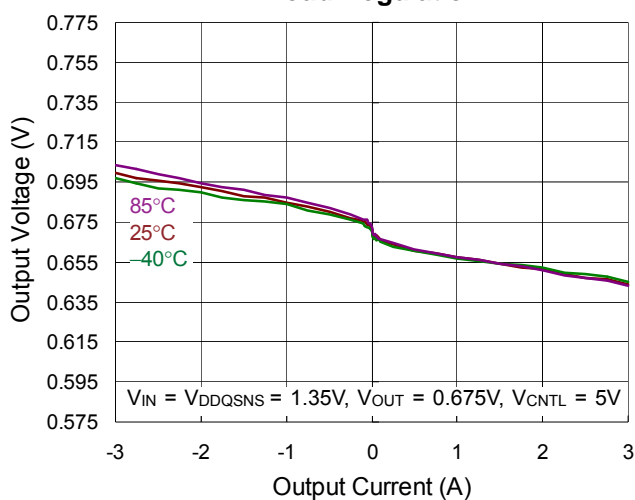
### Load Regulation



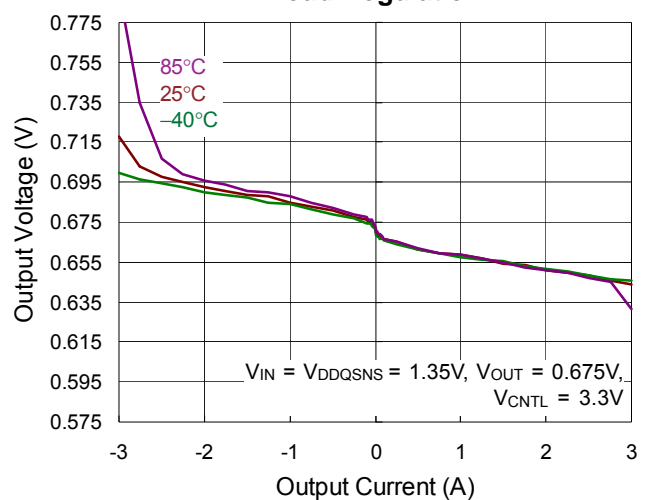
### Load Regulation



### Load Regulation

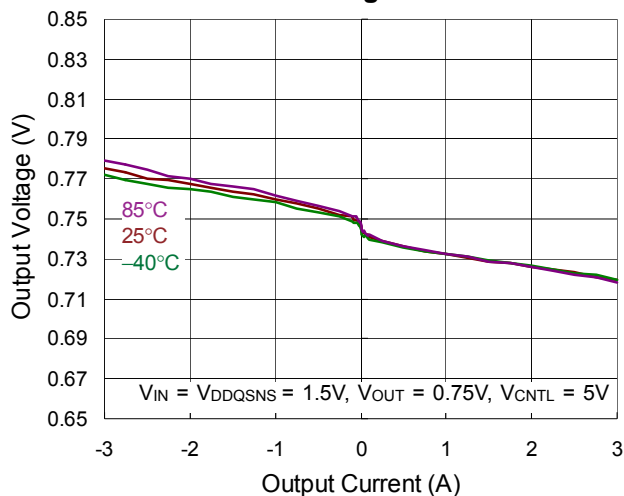


### Load Regulation

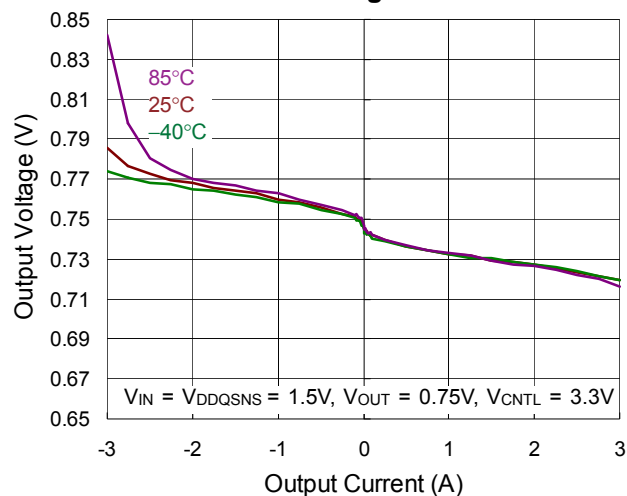




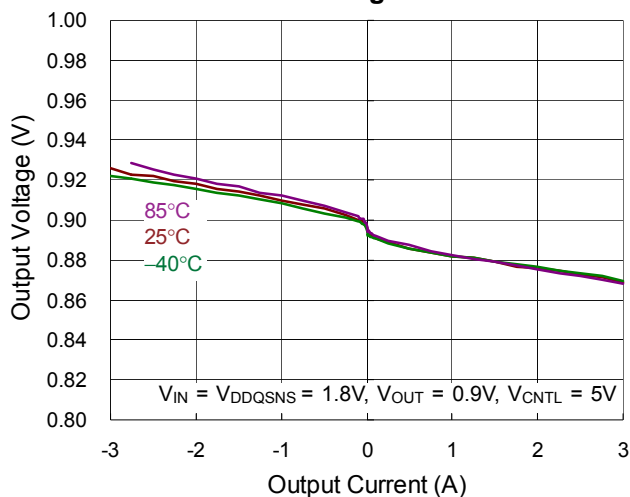
Load Regulation



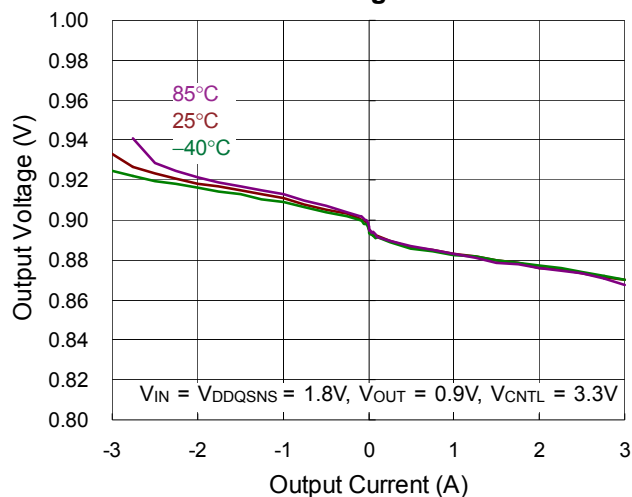
Load Regulation



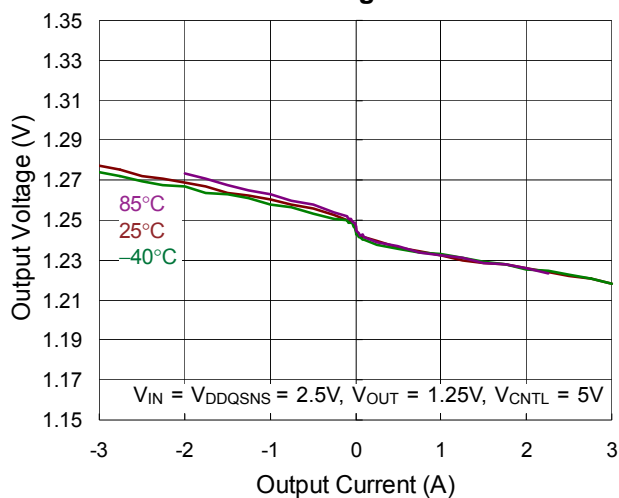
Load Regulation



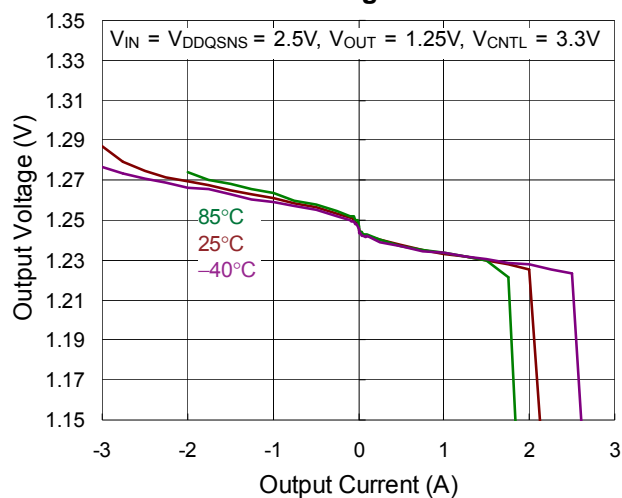
Load Regulation



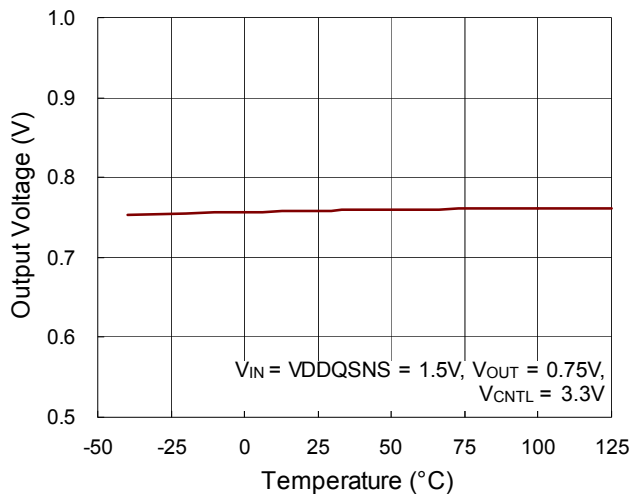
Load Regulation



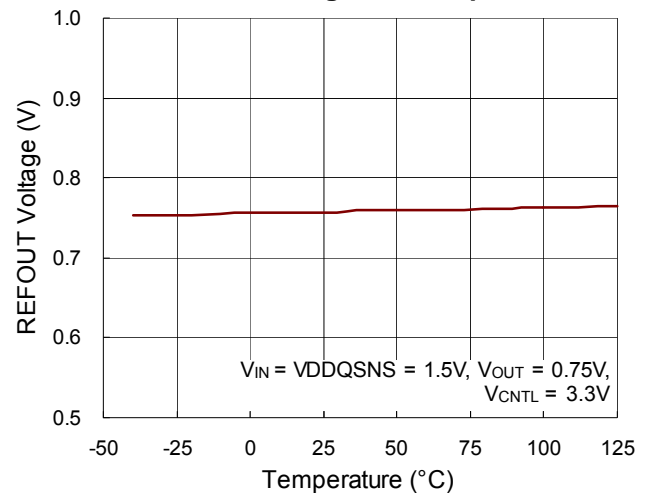
Load Regulation



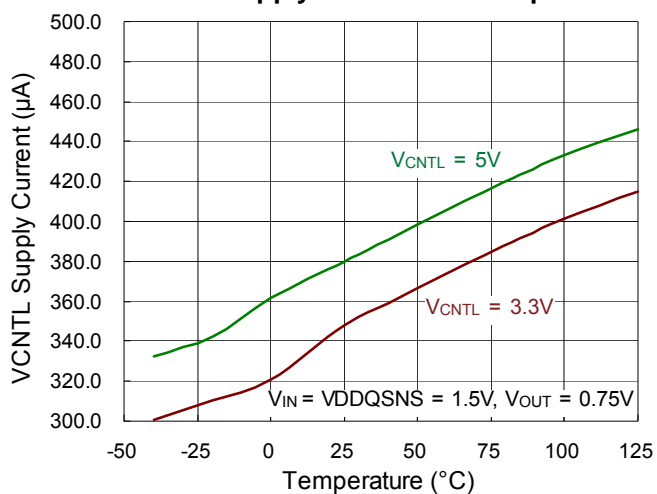
Output Voltage vs. Temperature



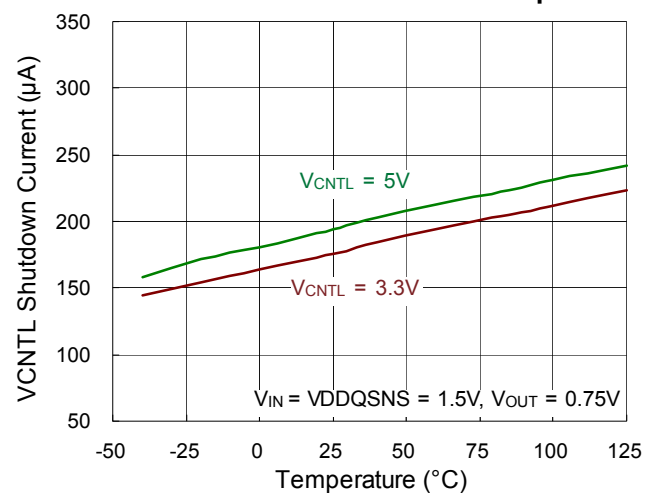
REFOUT Voltage vs. Temperature



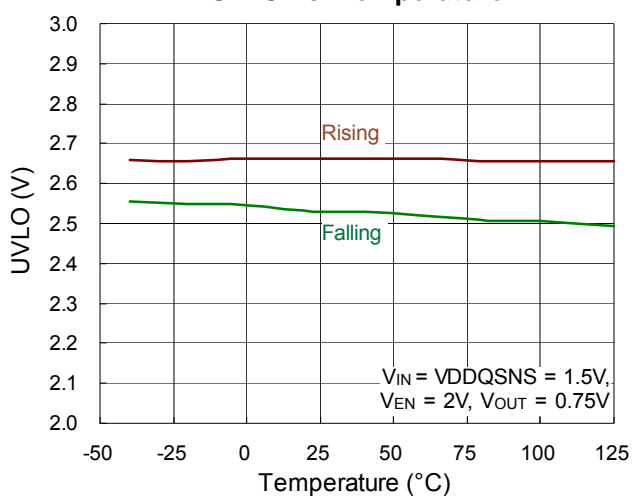
VCNTL Supply Current vs. Temperature



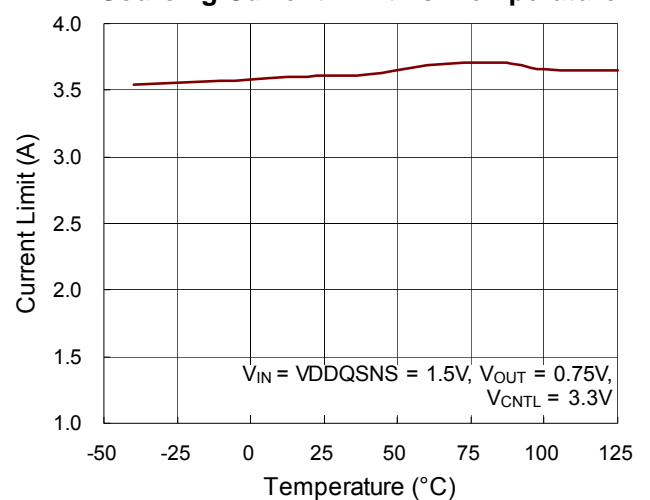
VCNTL Shutdown Current vs. Temperature



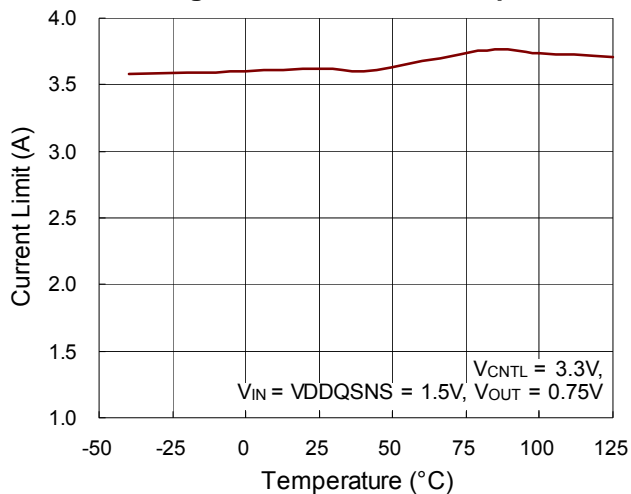
UVLO vs. Temperature



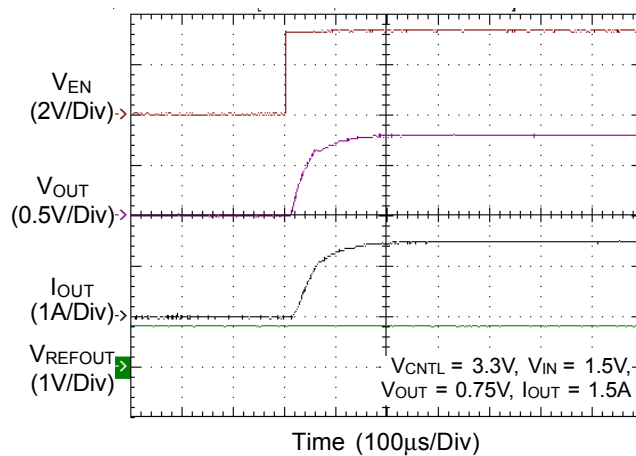
Sourcing Current Limit vs. Temperature



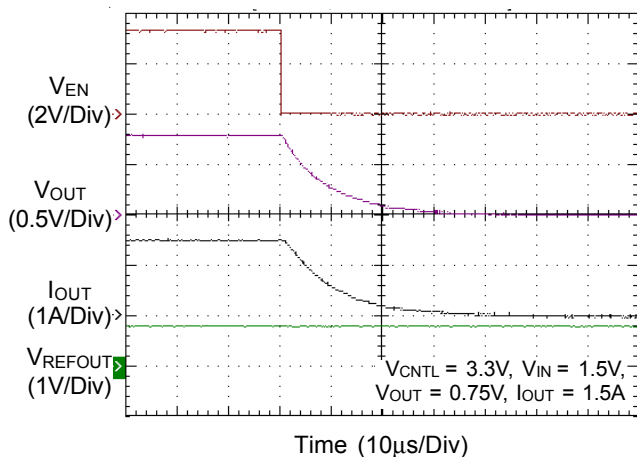
Sinking Current Limit vs. Temperature



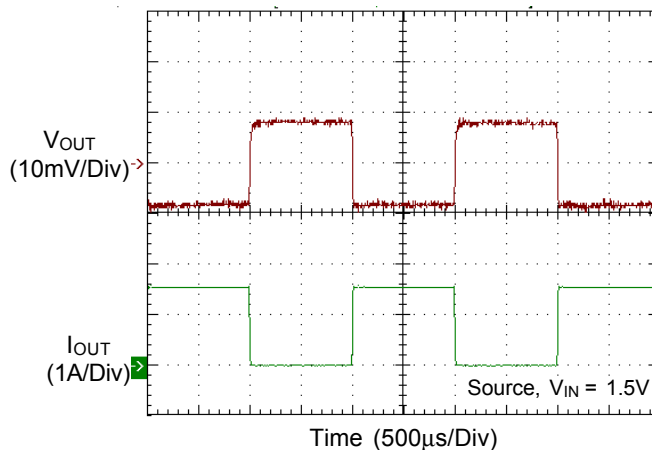
Power On from EN



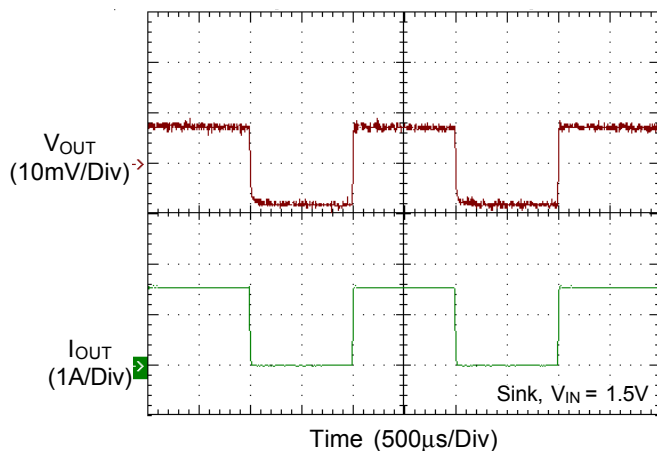
Power Off from EN



0.75V<sub>OUT</sub> @ 1.5A Transient Response



0.75V<sub>OUT</sub> @ 1.5A Transient Response



## Application Information

The RTQ2508 is a 3A sink/source tracking termination regulator. It is specifically designed for low-cost and low-external component count system such as notebook PC applications. The RTQ2508 possesses a high speed operating amplifier that provides fast load transient response.

### Capacitor Selection

Good bypassing is recommended from VIN to GND to help improve AC performance. A 10 $\mu$ F or greater input capacitor located as close as possible to the IC is recommended. The input capacitor must be located at a distance of less than 0.5 inches from the VIN pin of the IC.

Adding a 1 $\mu$ F ceramic capacitor close to the VCNTL pin and it should be kept away from any parasitic impedance from the supply power. For stable operation, the total capacitance of the ceramic capacitor at the output terminal must be larger than 30 $\mu$ F. The RTQ2508 is designed specifically to work with low ESR ceramic output capacitor in space saving and performance consideration. Larger output capacitance can reduce the noise and improve load transient response, stability and PSRR. The output capacitor should be located near the VOUT pin as close as possible.

### Thermal Considerations

The junction temperature should never exceed the absolute maximum junction temperature  $T_{J(MAX)}$ , listed under Absolute Maximum Ratings, to avoid permanent damage to the device. The maximum allowable power dissipation depends on the thermal resistance of the IC package, the PCB layout, the rate of surrounding airflow, and the difference between the junction and ambient temperatures. The maximum power dissipation can be calculated using the following formula :

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

where  $T_{J(MAX)}$  is the maximum junction temperature,  $T_A$  is the ambient temperature, and  $\theta_{JA}$  is the junction-to-ambient thermal resistance.

For continuous operation, the maximum operating junction temperature indicated under Recommended Operating

Conditions is 125°C. The junction-to-ambient thermal resistance,  $\theta_{JA}$ , is highly package dependent. For a WDFN-10L 3x3 package, the thermal resistance,  $\theta_{JA}$ , is 40°C/W on a standard JEDEC 51-7 high effective-thermal-conductivity four-layer test board. The maximum power dissipation at  $T_A = 25^\circ\text{C}$  can be calculated as below :

$$P_{D(MAX)} = (125^\circ\text{C} - 25^\circ\text{C}) / (40^\circ\text{C/W}) = 2.5\text{W for a WDFN-10L 3x3 package.}$$

The maximum power dissipation depends on the operating ambient temperature for the fixed  $T_{J(MAX)}$  and the thermal resistance,  $\theta_{JA}$ . The derating curves in Figure 1 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.

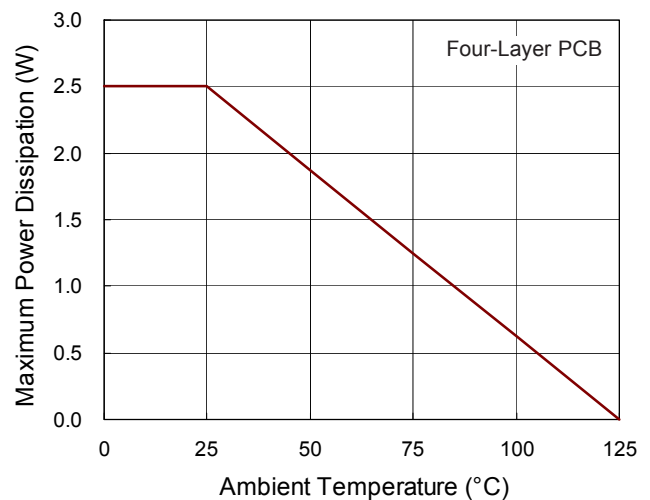


Figure 1. Derating Curve of Maximum Power Dissipation

### Layout Considerations

For best performance of the RTQ2508, the PCB layout suggestions below are highly recommend :

- ▶ With wide and short connection plane between capacitors and pins for trace impedance minimization.
- ▶ The ground plane connected by a wide copper surface for good thermal dissipation, add via connection also helps reduce the GND loop trace.
- ▶ Connect the SENSE pin to the positive node of output capacitor at VOUT terminal for output target level remote sensing.

- ▶ Since the output voltage  $V_{OUT}$  setting is follow the REFIN pin input voltage level  $V_{REFIN}$  ( $V_{OUT} = V_{REFIN}$ ), the REFIN pin can connected with independent voltage source for stable input signal and good  $V_{OUT}$  target accuracy. For the application which  $V_{REFIN}$  sinks the voltage source divided from  $V_{IN}$  power trace, with separate connection trace between R1 and  $V_{IN}$  terminal side for good  $V_{REFIN}$  signal stability also avoid the reference voltage level shrink down caused by  $V_{IN}$  trace loss at high load operation.

Figure 2 shows an example for the layout reference that reduce conduction trace loop, helping inductive parasitic minimize, load transient reduction and good circuit stability.

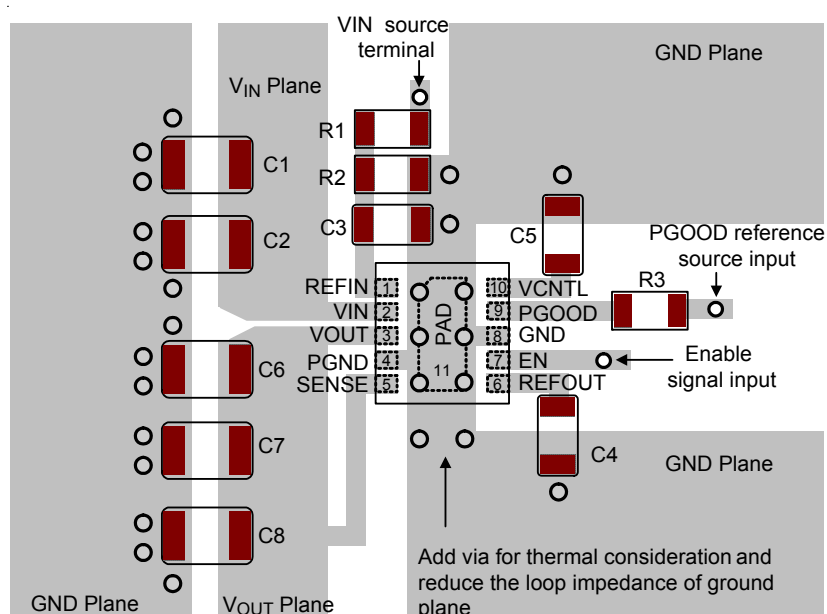
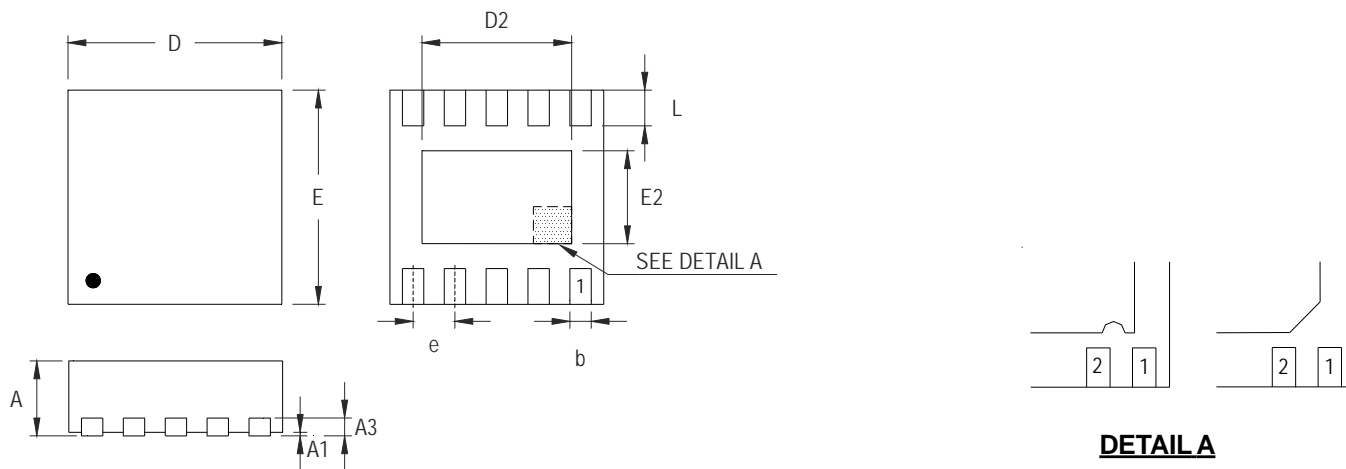


Figure 2. PCB Layout Guide

## Outline Dimension



### DETAIL A

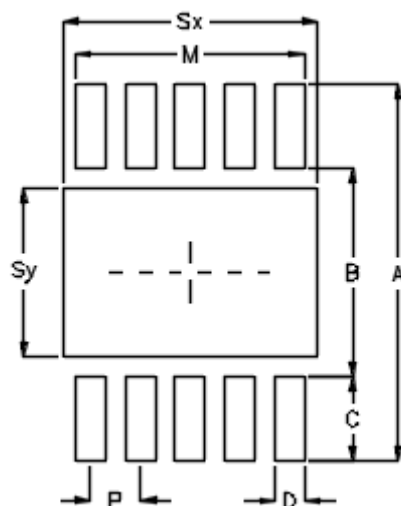
Pin #1 ID and Tie Bar Mark Options

Note : The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	0.700	0.800	0.028	0.031
A1	0.000	0.050	0.000	0.002
A3	0.175	0.250	0.007	0.010
b	0.180	0.300	0.007	0.012
D	2.950	3.050	0.116	0.120
D2	2.300	2.650	0.091	0.104
E	2.950	3.050	0.116	0.120
E2	1.500	1.750	0.059	0.069
e	0.500		0.020	
L	0.350	0.450	0.014	0.018

W-Type 10L DFN 3x3 Package

## Footprint Information



Package	Number of Pin	Footprint Dimension (mm)								Tolerance
		P	A	B	C	D	Sx	Sy	M	
V/W/U/X/ZDFN3*3-10	10	0.50	3.80	2.10	0.85	0.30	2.55	1.70	2.30	±0.05

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