



FILTERLESS 2W CLASS-D STEREO AUDIO AMPLIFIER WITH DC VOLUME CONTROL AND HEADPHONE OUTPUT

Description

The PAM8602E is a 2.0W, Class-D audio amplifier with headphone amplifier. Advanced 64-Step DC volume control minimizes external components and allows speaker volume control and headphone volume control. It offers low THD+N, to produce high-quality sound reproduction. The new filterless architecture allows the device to drive the speaker directly, without low-pass output filters which will save 30% system cost and 75% PCB area.

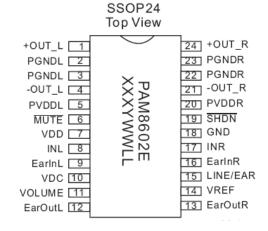
With the same numbers of external components, the efficiency of the PAM8602E is much better than Class-AB cousins. It can extend the battery life thus be ideal for portable applications.

The PAM8602E is available in a SSOP-24 and SOP-24 package.

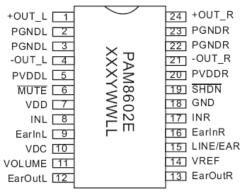
Features

- 2W Output with a 4Ω Load and 5V Power Supply
- Filterless, Low Quiescent Current and Low EMI
- Low THD+N
- 64-Step DC Volume Control
- Headphone Output Function
- Superior Low Noise
- Low Pop Noise
- Efficiency up to 88%
- Short Circuit Protection
- Thermal Shutdown
- Few External Components to Save the Space and Cost
- Pb-Free Package

Pin Assignments



SOP24 Top View



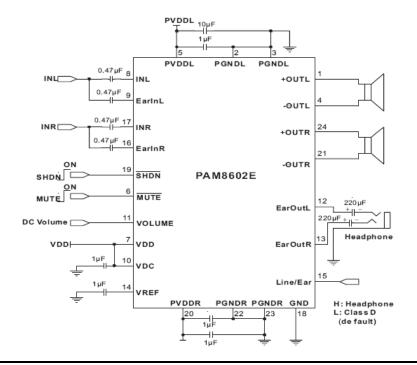
Applications

- LCD Monitors / TV Projectors
- Notebook Computers
- Portable Speakers
- Portable DVD Players, Game Machines
- VoIP/Speakers Phones





Typical Applications Circuit



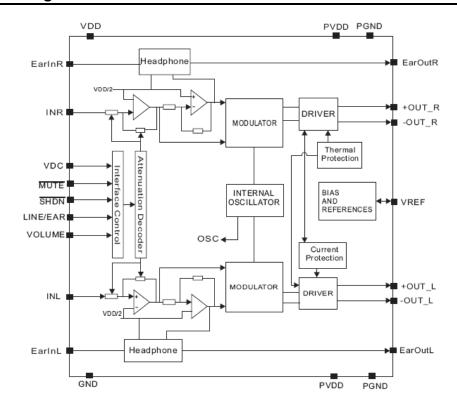
Pin Descriptions

Pin Number	Pin Name	Function
1	+OUT_L	Left Channel Positive Output
2	PGNDL	Left Channel Power GND
3	PGNDL	Left Channel Positive GND
4	-OUT_L	Left Channel Negative Output
5	PVDDL	Left Channel Power Supply
6	MUTE	Mute Control Input (active low)
7	VDD	Analog VDD
8	IN L	Left Channel Input
9	EAR IN L	Left Earphone Input
10	VDC	Analog Reference for Gain Control Section
11	VOLUME	DC Voltage Control to Set the Gain of Class-D
12	EAR OUT L	Left Earphone Output
13	EAR OUT R	Right Earphone Output
14	VREF	Internal Analog Reference, Connect a Bypass Capacitor from VREF to GND
15	LINE/EAR	Line/ Ear Detect
16	EAR IN R	Right Earphone Input
17	INR	Right Channel Input
18	GND	Analog GND
19	SHDN	Shutdown Control Input (active low)
20	PVDDR	Right Channel Power Supply
21	-OUT_R	Right Channel Negaitive Output
22	PGNDR	Right Channel Power GND
23	PGNDR	Right Channel Power GND
24	+OUT_R	Right Channel Positive Output





Functional Block Diagram



Absolute Maximum Ratings (@TA = +25°C, unless otherwise specified.)

These are stress ratings only and functional operation is not implied. Exposure to absolute maximum ratings for prolonged time periods may affect device reliability. All voltages are with respect to ground.

Parameter	Rating	Unit	
Supply Voltage	6.0	V	
Input Voltage	-0.3 to V _{DD} +0.3	\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \	
Operation Junction Temperature	-40 to +125		
Storage Temperature	-65 to +150	°C	
Soldering Temperature	300, 5 sec		

Recommended Operating Conditions (@TA = +25°C, unless otherwise specified.)

Parameter	Rating	Unit
Supply Voltage Range	2.5 to 5.5	V
Ambient Operation Temperature Range	-20 to +85	°C
Junction Temperature Range	-20 to +125	°C

Thermal Information

Parameter	Package	Symbol	Max	Unit
Thermal Resistance (Junction to Ambient)	SSOP-24	θ_{JA}	96	°C/W
Thermal Resistance (Junction to Ambient)	SOP-24	θ_{JA}	79.2	°C/W





Electrical Characteristics (@ T_A = +25°C, V_{DD} = 5V, Gain = Maximum, R_L = 8 Ω , unless otherwise specified.)

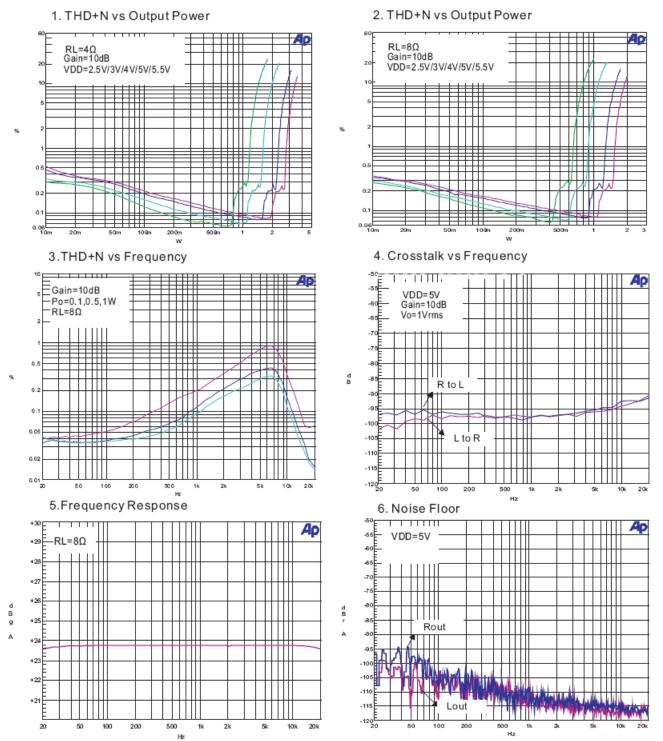
Parameter	Symbol	Test Conditions		Min	Тур	Max	Units	
Class D Stage	-	1			1		ı	.1
Supply Voltage Range	V_{DD}				2.5		5.5	V
Quiescent Current	ΙQ	No Load				10	15	mA
Output Offset Voltage	Vos	No Load				10	50	mV
Drain Sauras On State Begintanes	_	1 -0.54		P MOSFET		0.35		Ω
Drain-Source On-State Resistance	R _{DS(ON)}	I _{DS} = 0.5A		N MOSFET		0.25		
Output Power	Po	THD+N = 10%, f = 1k	кHz		1.55	1.70		W
Total Harmonic Distortion Plus Noise	THD+N	P _O = 0.85W, f = 1KH;	Z			0.08		%
Power Supply Ripple Rejection	PSRR	Input AC-GND, f = 1k	KHz, V _P	_P = 200mV		70		dB
Channel Separation	CS	P _O = 1W, f = 1KHz				-95		dB
Oscillator Frequency	fosc				200	250	300	kHz
Efficiency	η	P _O = 1.7W, f = 1 kHz			85	89		%
Noise	.,	Input AC-GND	A-W	/eighting		220		μV
Noise	V _N		No A	A-Weighting		350		
Signal Noise Ratio	SNR	f = 20 – 20kHz, THD	= 1%			85		dB
Earphone Stage								
Quiescent Current	IQ	No Load			4.5	7.5	mA	
Output Offset Voltage	Vos	No Load			2.5		V	
Output Power	Po	THD+N = 1%, R _L = 3	2Ω, f =	1KHz		60		mW
Total Harmonic Distortion Plus Noise	THD+N	$R_L = 32\Omega, P_O = 10mV$	N, f = 1	kHz		0.02		%
Power Supply Ripple Rejection	PSRR	Input AC-GND, f = 1k	Hz, V _{PI}	⇒ = 200mV		75		dB
Channel Separation	CS	P _O = 1W, f = 1kHz				-85		dB
Noise	\/	Input	A-W	/eighting		70		μV
Noise	V _N	AC-GND	No A	A-Weighting		40		μν
Signal Noise Ratio	SNR	f = 20 - 20kHz, THD =	= 1%			75		dB
Control Section								
Under Voltage Lock-Out	UVLO					2		
Mute Current	I _{MUTE}	V _{MUTE} = 0V			1	3	mA	
Shutdown Current	I _{SHDN}	V _{SHDN} = 0V				1	μΑ	
SHDN Input High	V_{SH}			1.2				
SHDN Input High	V_{SL}					0.5		
MUTE Input High	V_{MH}			1.2				
MUTE Input High	V _{ML}					0.5		
Over Temperature Protection	OTP				140		°C	
Over Temperature Hysteresis	OTH				30		°C	





Typical Performance Characteristics (@T_A = +25°C, unless otherwise specified.)

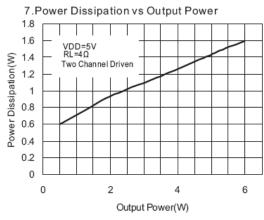
Class-D Output



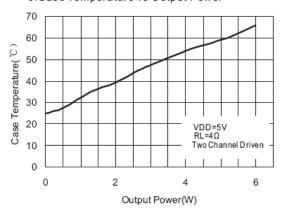




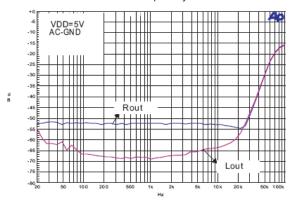
Class-D Output

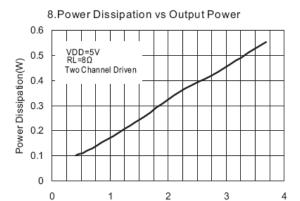






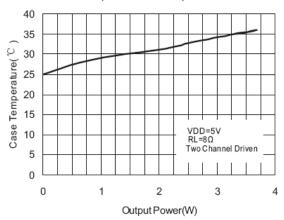
11. PSRR vs Frequency





10. Case Temperature vs Output Power

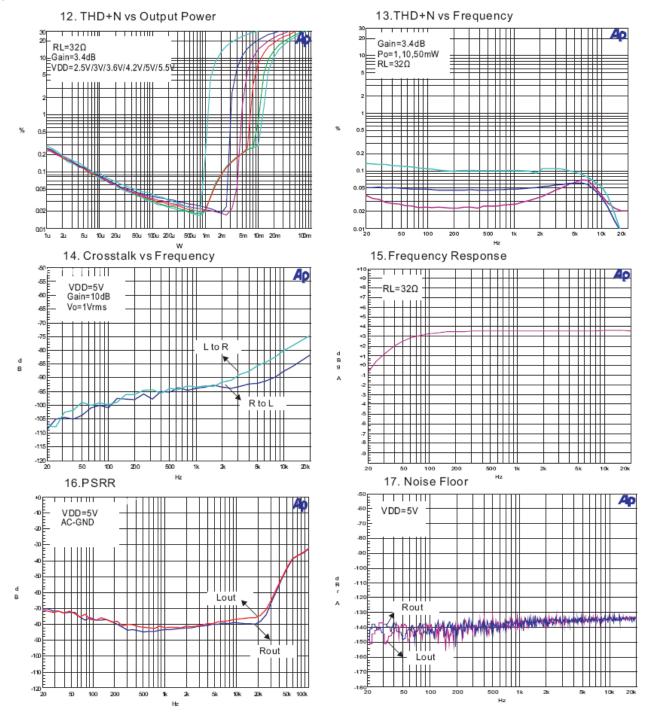
Output Power (W)





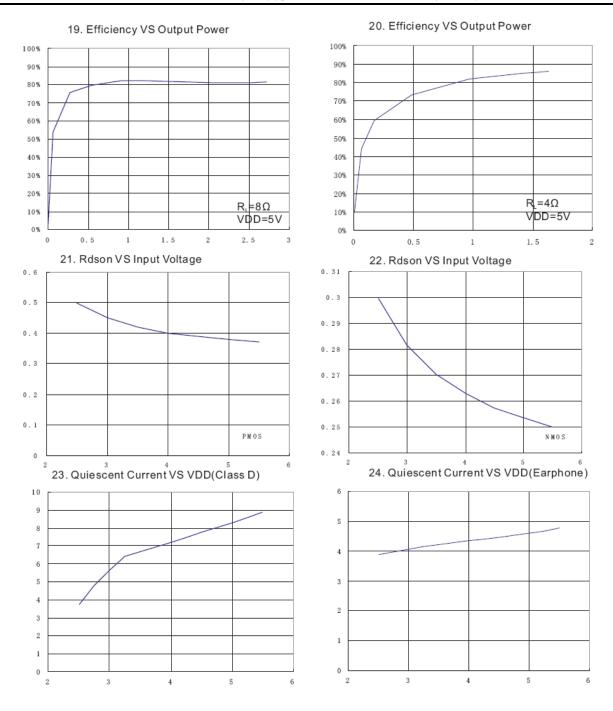


Earphone Output



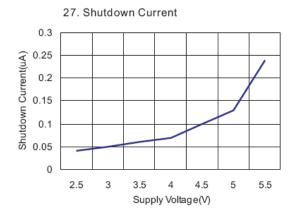


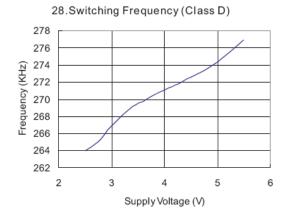


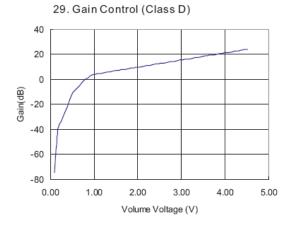












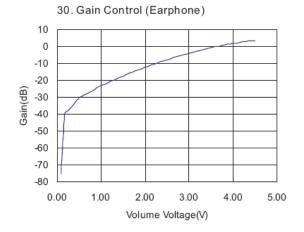






Table 1. DC Volume Control

STEP	Gain (dB) Class D	Gain (dB) Earphone	STEP	Gain (dB) Class D	Gain (dB) Earphone
0	-75	-75	32	11.6	-9.2
1	-40	-40	33	12.0	-8.6
2	-34	-38	34	12.4	-8.0
3	-28	-36	35	12.8	-7.4
4	-22	-34	36	13.2	-6.8
5	-16	-32	37	13.6	-6.2
6	-10	-30	38	14.0	-5.7
7	-7.5	-29	39	14.4	-5.2
8	-5.0	-28	40	14.8	-4.7
9	-2.5	-27	41	15.2	-4.2
10	0	-26	42	15.6	-3.7
11	1.5	-25	43	16.0	-3.2
12	3.0	-24	44	16.4	-2.7
13	4.0	-23	45	16.8	-2.2
14	4.4	-22.2	46	17.2	-1.8
15	4.8	-21.4	47	17.6	-1.4
16	5.2	-20.6	48	18.0	-1.0
17	5.6	-19.8	49	18.4	-0.6
18	6.0	-19.0	50	18.8	-0.2
19	6.4	-18.2	51	19.2	0.2
20	6.8	-17.4	52	19.6	0.6
21	7.2	-16.6	53	20.0	0.9
22	7.6	-15.9	54	20.4	1.2
23	8.0	-15.2	55	20.8	1.5
24	8.4	-14.5	56	21.2	1.8
25	8.8	-13.8	57	21.6	2.1
26	9.2	-13.1	58	22.0	2.4
27	9.6	-12.4	59	22.4	2.7
28	10.0	-11.7	60	22.8	2.9
29	10.4	-11.0	61	23.2	3.1
30	10.8	-10.4	62	23.6	3.3
31	11.2	-9.8	63	24.0	3.5

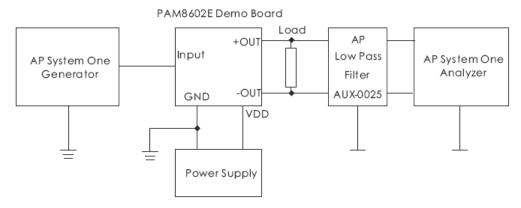




Application Information

Test Setup for Performance Testing (Class D)

- 1. When the PAM8602E works with LC filters, it should be connected with the speaker before it's powered on, otherwise it will be damaged easily.
- 2. When the PAM8602E works without LC filters, it's better to add a ferrite chip bead at the outgoing line of speaker for suppressing the possible electromagnetic interference.
- 3. The absolute maximum rating of the PAM8602E operation voltage is 6V. When the PAM8602E is powered with four battery cells, it should be noted that the voltage of four new dry or alkaline batteries is over 6V, higher than its maximum operation voltage, which probably make the device damaged. Therefore, it's recommended to use either four Ni-MH (Nickel Metal Hydride) rechargeable batteries or three dry or alkaline batteries.
- 4. The input signal should not be too high, if too high, it will cause the clipping of output signal when increasing the volume. Because the DC volume control of the PAM8602E has big gain, it will make the device damaged.
- 5. When testing the PAM8602E without LC filters by using resistor instead of speaker as the output load, the test results, e.g. THD or efficiency, will be worse than those using speaker as load.



Notes:

- 1. The Audio Precision (AP) AUX-0025 low pass filter is necessary for class-D amplifier measurement with AP analyzer.
- 2. Two 22µH inductors are used in series with load resistor to emulate the small speaker for efficiency measurement.

Mute Operation

The MUTE pin is an input for controlling the output state of the PAM8602E. A logic low on this pin disables the outputs, and a logic high enables the outputs. This pin may be used as a quick disable or enable of the outputs without a volume fade. Quiescent current is listed in the electrical characteristic table. The MUTE pin can be left floating due to the internal pull-up.

Shutdown Operation

In order to reduce power consumption while not in use, the PAM8602E contains shutdown circuitry to turn off the amplifier's bias circuitry. The amplifier is turned off when logic low is placed on the SHDN pin. By switching the SHDN pin connected to GND, the PAM8602E supply current draw will be minimized in idle mode. The SHDN pin can be left floating due to the pull-up.

For the best power on/off pop performance, the amplifier should be placed in the Mute mode prior to turning on/off the power supply.

Power Supply Decoupling

The PAM8602E is a high performance CMOS audio amplifier that requires adequate power supply decoupling to ensure the output THD and PSRR are as low as possible. Power supply decoupling affects low frequency on the power supply leads for higher frey response. Optimum decoupling is achieved by using two capacitors of different types that target different types of noise frequency transients, spike, or digital hash on the line, a good low equivalent-series-resisitance (ESR) ceramic capacitor, typically $1.0\mu\text{F}$, placed as close as possible to the device V_{DD} terminal works best. For filtering lower-frequency noise signals, a large capacitor of $10\mu\text{F}$ (ceramic) or greater placed near the audio power amplifier is recommended.





November 2012

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Application Information

Input Capacitor (C_I)

Large input capacitors are both expensive and space hungry for portable designs. Clearly, a certain sized capacitor is needed to capacitor is needed to couple in low frequencies without severe attenuation. But in many cases the speakers used in portable systems, whether internal or external, have little ability to reproduce signals below 100Hz to 150Hz. Thus, using a large input capacitor may not increase actual system performance. In this case, input capacitor (C_I) and input resisitance (R_I) of the amplifier form a high-pass filter with the corner frequency determined equation below.

$$f_C = \frac{1}{2\Pi R_l C_l}$$

In addition to system cost and size, click and pop perfomance is affected by the size of the input coupling capacitor, $C_{\rm L}$. A larger input coupling capacitor requires more charge to reach its quiescent DC voltage (nominally $^{1}\!\!/_{2}$ V_{DD}). This charge comes from the internal circuit via the feedback and is apt to creat pops upon device enable. Thus, by minimizing the capacitor size based on necessary low frequency response, turn-on pops can be minimized.

Analog Refernce Bypass Capacitor (CBYP)

Analog Reference Bypass Capacitor (C_{BYP}) is the most critical capacitor and serves several important functions. During start-up or recovery from shutdown mode, C_{BYP} determines the rate at which the amplifier starts up. The second function is to reduce noise produced by the power supply caused by coupling into the output drive signal. This noise is from the internal analog reference to the amplifier, which appears as degraded PSRR and THD+N.

A ceramic bypass capacitior (C_{BYP}) of $0.47\mu\text{F}$ to $1.0\mu\text{F}$ is recommended for the best THD and noise performance. Increasing the bypass capacitor reduces clicking and popping noise from power on/off and entering and leaving shutdown.

Under Voltage Lock-Out (UVLO)

The PAM8602E incorporates circuitry designed to detect when the supply voltage is low. When the supply voltage drops to 1.8V or below, the PAM8602E outputs are disable, are the device comes out of this state and states to normal functional once $V_{DD} \ge 2.0V$.

Short Circuit Protection (SCP)

The PAM8602E has short circuit protection circuitry on the outputs that prevents the device from damage when output-to-output and output-to-GND short. When a short circuit is detected on the outputs, the outputs are disabled immediately. If the short was removed, the device activates again.

Over Temperature Protection

Thermal protection on the PAM8602E prevents the device from damage when the internal die temperature exceeds +135°C. There is a 15 degree tolerance on this point from device to device. Once the die temperature exceeds the thermal set point, the device outputs are disabled. This is not a latched fault. The thermal fault is cleared once the temperature of the die is reduced by 30°C. This large hysteresis will prevent motor boating sound well. The device begins normal operation at this point without external system interaction.

How to Reduce EMI (Electro Magnetic Interference)

A simple solution is to put an additional capacitor $1000\mu F$ at power supply terminal for power line coupling if the traces from amplifier to speakers are short (< 20CM).

Most applications require a ferrite bead filter as shown at Figure 1. The ferrite filter reduces EMI around 1MHz and higher. When selecting a ferrite bead, choose one with high impedance at high frequencies, and low impedance at low frequencies (MH2012HM221-T).

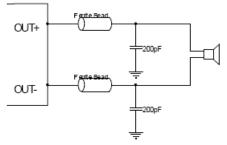


Figure 1. Ferrite Bead Filter to Reduce EMI



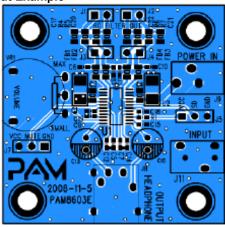


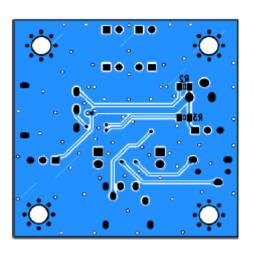
Application Information (cont.)

PCB Layout Guidelines Grounding

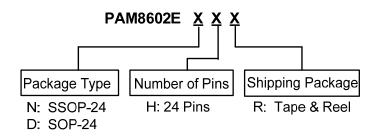
At this stage it is paramount to notice the necessity of separate grounds. Noise currents in the output power stage need to be returned to output noise ground and nowhere else. Were these currents to circulate elsewhere, they may get into the power supply, the signal ground, etc, worse yet, they may form a loop and radiate noise. Any of these cases results in degraded amplifier performance. The logical returns for the output noise currents associated with Class-D switching are the respective PGND pins for each channel. The switch state diagram illustrates that PGND is instrumental in nearly every switch state. This is the perfect point to which the output noise ground trace should return. Also note that output noise ground is channel specific. A two channel amplifier has two seperate channels and consequently must have two seperate output noise ground traces. The layout of the PAM8602E offers separate PGND connections for each channel and in some cases each side of the bridge. Output noise grounds must be tied to system ground at the power in exclusively. Signal currents for the inputs, reference, etc need to be returned to quite ground. This ground is only tied to the signal components and the GND pin, and GND then tied to system ground.

PCB Layout Example





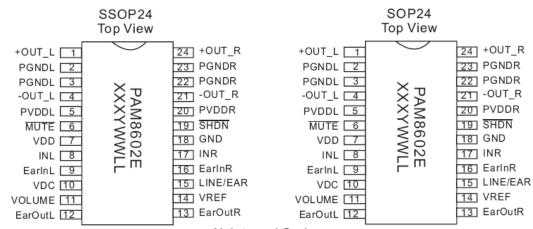
Ordering Information



Part Number	Package Type	Standard Package	
PAM8602ENHR	SSOP-24	2500 Units/Tape&Reel	
PAM8602EDHR	SOP-24	100 Units/Tape&Reel	



Marking Information



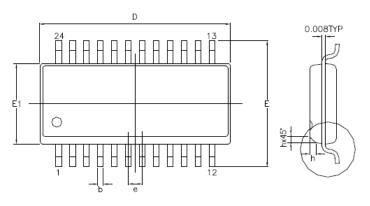
X: Internal Code

Y: Year WW: Week

LL: Internal Code

Package Outline Dimensions (All dimensions in mm.)

SSOP-24



SYMBOLS	MIN.	NOM.	MAX.		
A	0.053	0.061	0.069		
A1	0.004	-	0.010		
A2	0.049	0.057	0.065		
ь	0.008	0.010	0.012		
D	0.335	0.341	0.347		
Ε	0.228	0.236	0.244		
E1	0.150	0.154	0.158		
e	-	0.025	-		
L	0.016	0.033	0.050		
L1	0.041 REF				
R	0.003	-	-		
R1	0.003	-	-		
h	0.010	0.015	0.020		
θ	0.	4.	8*		
θ1	5*	10°	15*		
θ2	o	-	-		

UNIT : INCH

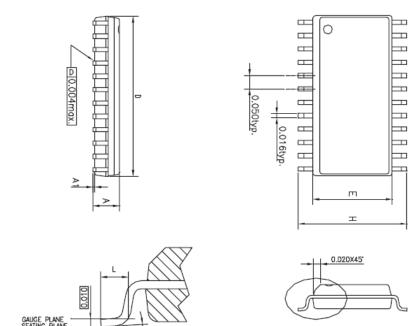
A2 A11	4×θ1 θ2 R1 4×θ1
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Package Outline Dimensions (cont.) (All dimensions in mm.)

SOP-24



SYMBOLS	MIN.	NOM	MAX.
Α	0.093	0.099	0.104
A1	0.004	_	0.012
D	0.599	0.600	0.614
E	0.291	0.295	0.299
Н	0.394	0.406	0.419
L	0.016	0.035	0.050
θ°	0	_	8
			UNIT : INCH

- NOTES:
 1.JEDEC OUTLINE: MS-013 AD
 2.DIMENSIONS "D" DOES NOT INCLUDE MOLD FLASH,
 PROTRUSIONS OR GATE BURRS.MOLD FLASH, PROTRUSIONS
 AND GATE BURRS SHALL NOT EXCEED .15mm (.006in)
- 3.DIMENSIONS "E" DOES NOT INCLUDE INTER-LEAD FLASH, OR PROTRUSIONS. INTER-LEAD FLASH AND PROTRUSIONS SHALL NOT EXCEED .25mm (.010in) PER SIDE.





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