

Features

- 2.5V to 5.5V Input Voltage Range
- 1.0A Output Current
- 1.5MHz Constant Frequency Operation
- No Schottky Diode Required
- Feedback Voltage as Low as 0.1V
- 100% Duty Cycle in Dropout
- Low Quiescent Current: 50μA
- Short Circuit Protection
- Thermal Fault Protection
- Input over voltage protection (OVP)
- <1μA Shutdown Current
- Compact package: SOT23-5

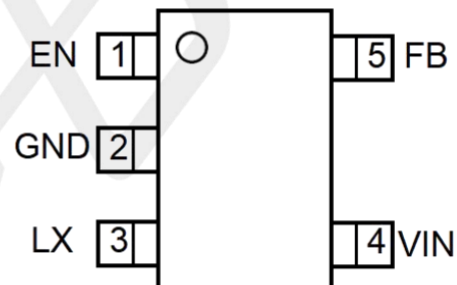
Applications

- LED Drivers
- 3AA or 4AA Batteries Powered Flashlight
- 1 Cell Li-Ion Battery Powered Flashlight

General Description

The is a constant frequency, current mode PWM step-down LED driver with 1A output current and 1.5MHz switch frequency. The device integrates a main switch and a synchronous rectifier for high efficiency without an external Schottky diode. This device offers two operation modes, PWM control and PFM Mode switching control, which allows a high efficiency over the wider range of the LED current. The feedback voltage is 0.1V typically. The device also integrates many protection features included input OVP, cycle-by-cycle current limit, over temperature protection.

Pin out (top view)



Pin Configurations

Pin	Name	Function
1	EN	Chip Enable Pin. Drive EN above 1.5V to turn on the part. Drive EN below 0.4V to turn it off. Do not leave EN floating.
2	GND	Ground
3	LX	Power Switch Output. It is the switch node connection to Inductor.
4	VIN	Power Supply Input. Must be closely decoupled to GND with a 4.7μF or greater ceramic capacitor.
5	FB	Feedback Pin. A resistor connected between FB to GND is used to program LED current, the threshold voltage is 0.1V.

Absolute Maximum Ratings (Note1)

Parameter	Value	Unit
Input Supply Voltages	-0.3~6.5	V
LX Voltages	-0.3~6.5	V
EN, FB Voltage	-0.3~6.5	V
Storage Temperature Range	-65~150	°C
Junction Temperature (Note 2)	-40~150	°C
Power Dissipation	600	mW
Lead Temperature Soldering, 10sec	260	°C

ESD Rating

Items	Description	Value	Unit
V _{ESD_HBM}	Human Body Model for all pins	±2000	V
V _{ESD_CDM}	Charged Device Model for all pins	±500	V

Recommended Operating Conditions

Items	Description	Min	Max	Unit
Voltage Range	IN	2.5	5.5	V
T _J	Operating Junction Temperature	-40	125	°C
T _A	Operating Ambient Temperature	-40	85	°C

Thermal Resistance (Note3)

Items	Description	Value	Unit
θ _{JA}	Junction-to-ambient thermal resistance	200	°C/W
θ _{JC}	Junction-to-case(top) thermal resistance	108	°C/W

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2: T_J is calculated from the ambient temperature T_A and power dissipation P_D according to the following formula: T_J = T_A + (P_D) × θ_{JA}.

Note 3: Measured on JESD51-7, 4-layer PCB.

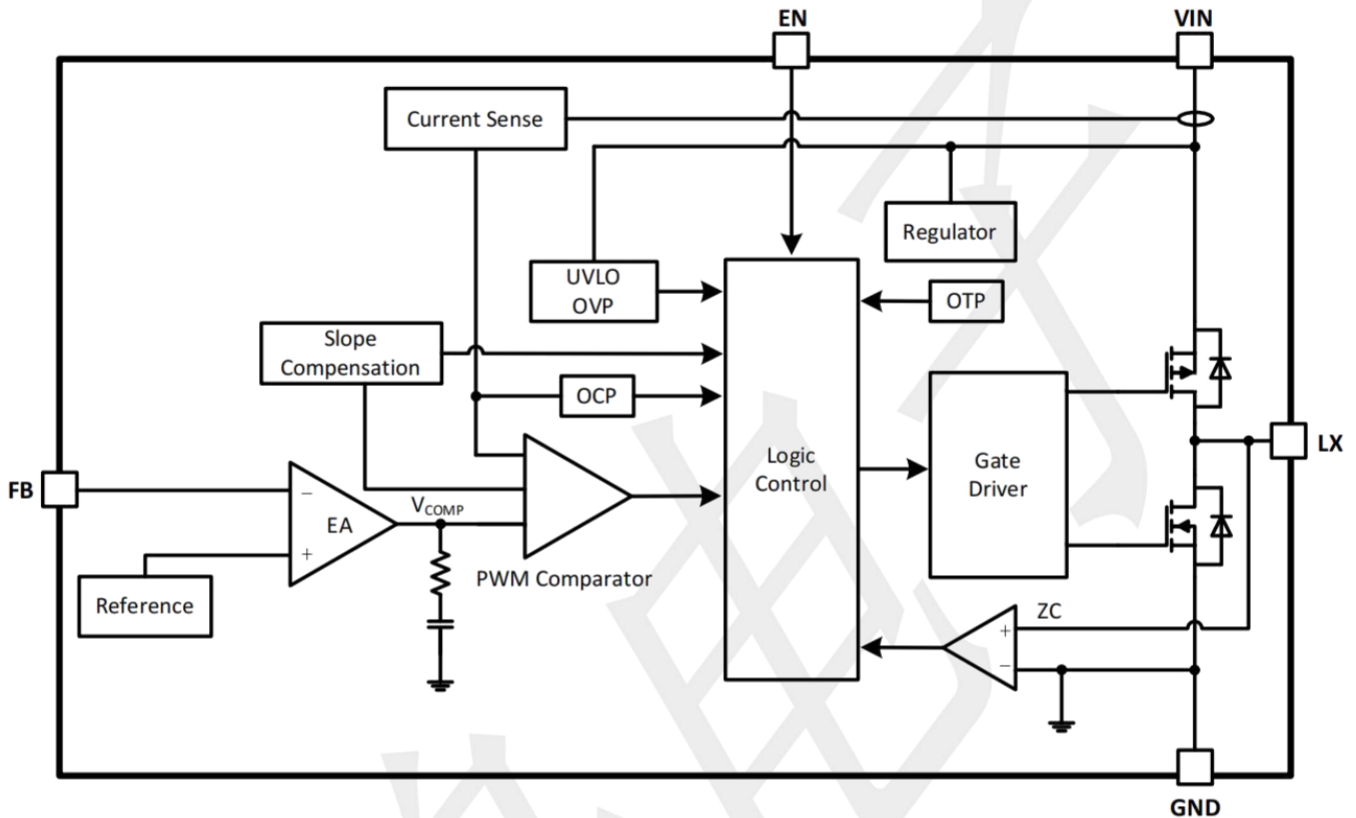
Electrical Characteristics

($V_{IN}=V_{EN}=5V$, $T_A = 25^\circ C$, unless otherwise noted.)

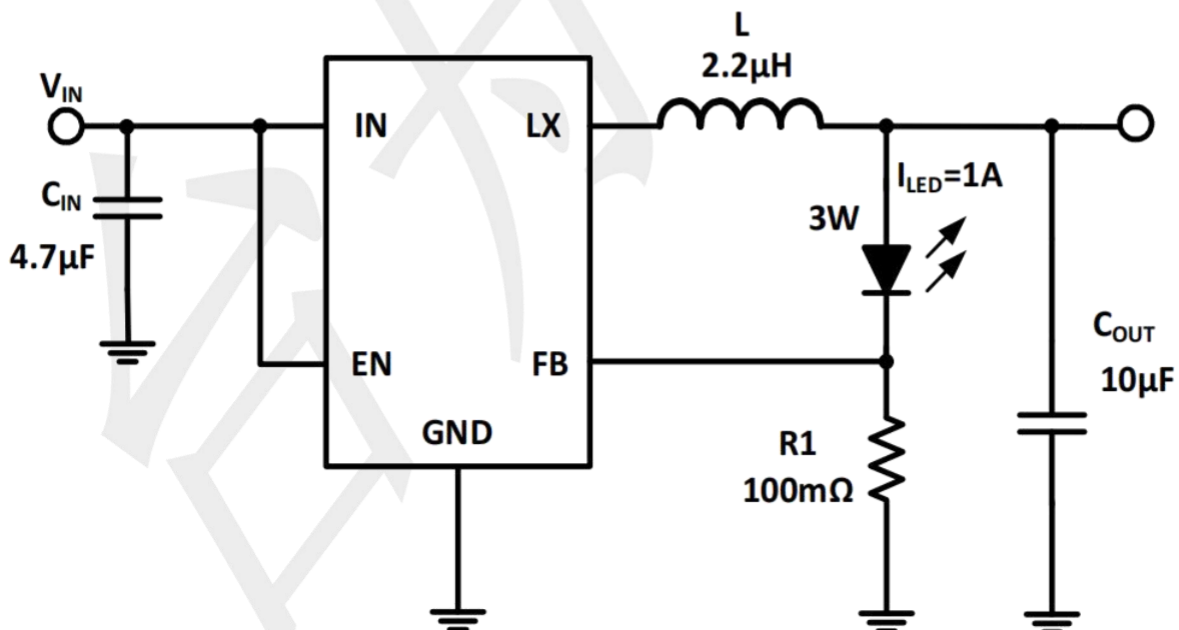
Parameter	Test Conditions	Min	Typ	Max	Unit
Input Voltage Range		2.5	--	5.5	V
OVP Threshold		--	6.0	--	V
UVLO Threshold		--	2.1	--	V
Quiescent Current	$V_{EN}=2.0V$, $I_{OUT}=0$, $V_{FB}=V_{REF} \times 105\%$	--	50	65	μA
Shutdown Current	$V_{EN}=0V$	--	0.1	10	μA
Regulated Feedback Voltage	$T_A = 25^\circ C$	95	100	105	mV
Oscillation Frequency	$V_{FB}=100mV$	--	1.5	--	MHz
	$V_{FB}=0V$	--	400	--	kHz
On Resistance of PMOS	$I_{LX}=100mA$	--	0.29	--	Ω
On Resistance of NMOS	$I_{LX}=-100mA$	--	0.18	--	Ω
Peak Current Limit		1.5	--	--	A
EN Input Low Level Voltage		--	--	0.4	V
EN Input High Level Voltage		1.5	--	--	V
EN Leakage Current		--	± 0.01	± 1.0	μA
LX Leakage Current	$V_{EN}=0V$, $V_{IN}=V_{LX}=5V$	--	± 0.01	± 1.0	μA
Thermal Shutdown Threshold (Note 4)		--	155	--	$^\circ C$
Thermal Shutdown Hysteresis (Note 4)		--	20	--	$^\circ C$

Note 4: Thermal shutdown threshold and hysteresis are guaranteed by design.

BLOCK DIAGRAM



Typical Application Circuits



FUNCTION DESCRIPTION

The is a high performance 1.0A, 1.5MHz monolithic step-down LED driver. The requires only three external power components (Cin, Cout and L). The LED driver current can be programmed with external feedback resistance between FB and GND pin up to 1.0A.

At dropout, the converter duty cycle increases to 100% and the output voltage tracks the input voltage minus the RDS(ON) drop of the high-side MOSFET.

The internal error amplifier and compensation provides excellent load, and line regulation. he device also integrates many protection features such as input OVP, cycle-by-cycle current limit, over temperature protection.

APPLICATION INFORMATION

Setting the Output Voltage

LED drive current of the can be externally programmed by resistor R1. The external resistor sets the LED current according to the following equation:

$$I_{LED} = \frac{V_{FB}}{R1}$$

Inductor Selection

For most designs, 2.2μH inductance can satisfy most application conditions. Inductance value is related to inductor ripple current value, input voltage, output voltage and switching frequency. The inductor value can be derived from the following equation:

$$L = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times \Delta I_L \times f_{OSC}}$$

Where ΔI_L is inductor ripple current. Large value inductors result in lower ripple current and small value inductors result in high ripple current, so inductor value has effect on output voltage ripple value. DC resistance of inductor which has impact on efficiency of DC/DC converter should be taken into account when selecting the inductor.

Input Capacitor Selection

The input capacitor reduces the surge current drawn from the input and switching noise from the device. The input capacitor impedance at the switching frequency should be less than input source impedance to prevent high frequency switching current passing to the input.

A low ESR input capacitor sized for maximum RMS current must be used. Ceramic capacitors with X5R or X7R dielectrics are highly recommended because of their low ESR and small temperature coefficients. A 4.7μF ceramic capacitor for most applications is sufficient. A large value may be used for improved input voltage filtering.

Output Capacitor Selection

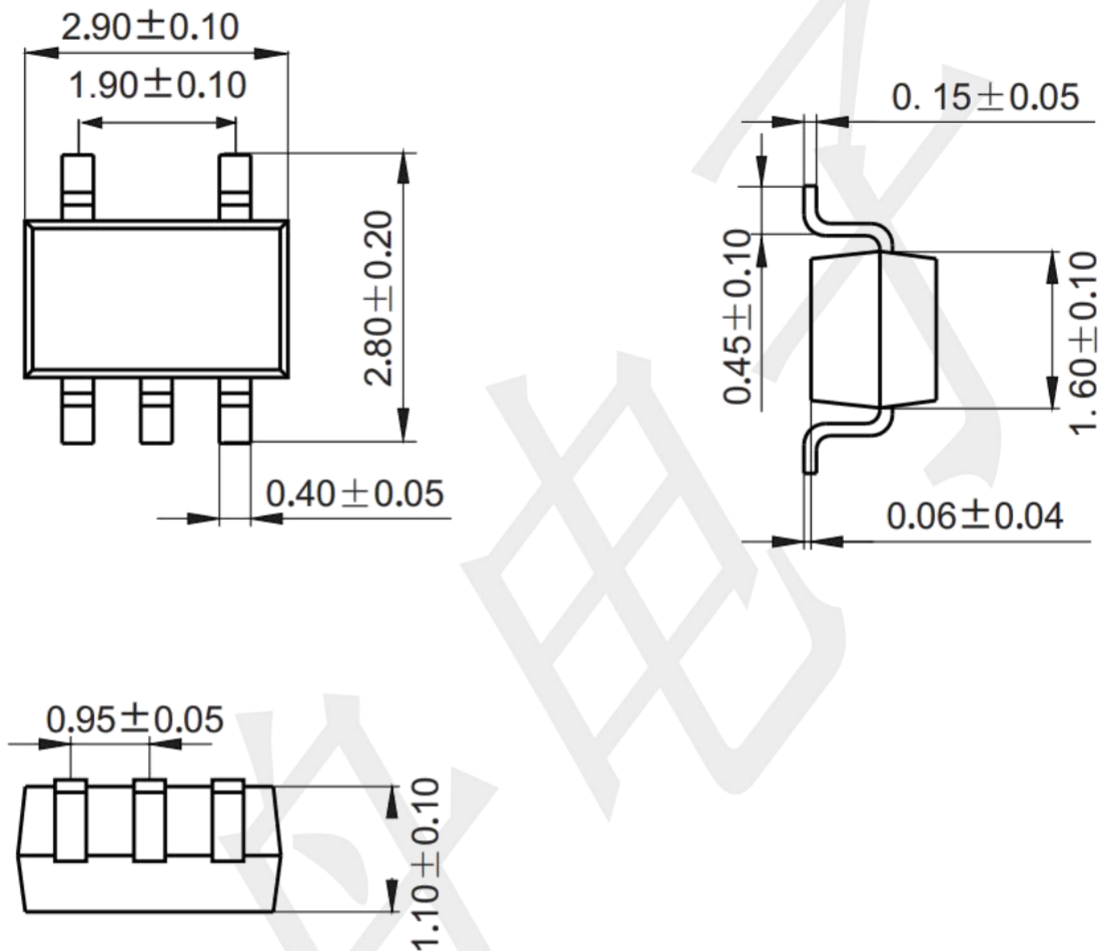
The output capacitor is required to keep the output voltage ripple small and to ensure regulation loop stability. The output capacitor must have low impedance at the switching frequency. Ceramic capacitors with X5R or X7R dielectrics are recommended due to their low ESR and high ripple current ratings. The output ripple VOUT is determined by:

$$\Delta V_{OUT} = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times L \times f_{OSC}} \times \left(ESR + \frac{1}{8 \times f_{osc} \times C_{OUT}} \right)$$

A 10μF ceramic can satisfy most applications.

Package Outline Dimensions (unit: mm)

SOT23-5



Mounting Pad Layout (unit: mm)

