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電源管理 顯示驅動 二三極管 LDO穩壓器 觸摸芯片











MOS管 運算放大器 存儲芯片

MCU

串口通信

LM2576T-3.3

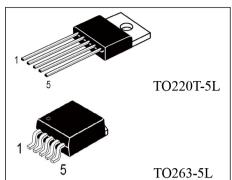
產品規格說明書



General Description

The

LM2576series of regulators are monolithic integrated circuit th atprovides all the active functions for a step-down (buck) switching regulator, capable of driving 3A load with excellent line and load regulation. The LM2576available in fixed output voltages of 3.3V, 5V, 12Vand an adjustable output version.



Requiring a minimum number of external components, these 5 TO2
regulators are simple to use and include internal frequency compensation and a fixed-frequency oscillator.

The LM2576series offers a high-efficiency replacement for popular three-terminal linear regulators. It substantially reduces the size of the heat sink, and in some cases no heat sink is required. A standard series of

LM2576are available from several different manufacturers.

inductors optimized for use with the

This feature greatly simplifies the design of switch-mode power supplies.

Other features include a guaranteed $\pm 4\%$ tolerance on output voltage within specified input voltages and output load conditions, and $\pm 10\%$ on the oscillator frequency. External shutdown is included, featuring 50 μ A (typical) standby current. The output switch includes cycle-by-cycle current limiting, as well as thermal shutdown for full protection under fault conditions.

The LM2576is available in TO220T-5L and TO263-5L package.

Features

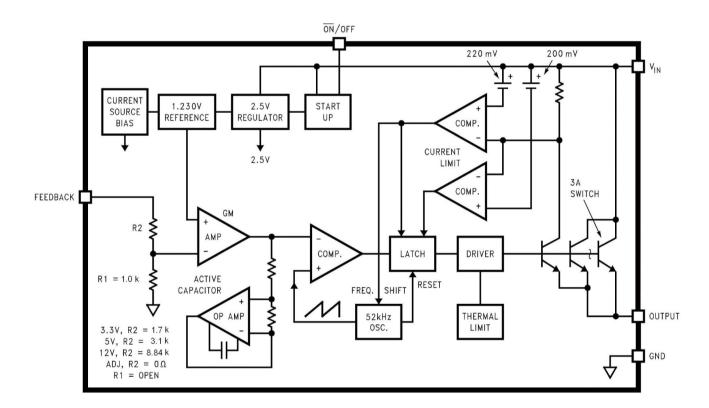
- 3.3V, 5V, 12V and adjustable output versions
- High efficiency
- Guaranteed 3A output current
- Requires only 4 external components
- 52 kHz fixed frequency internal oscillator
- TTL shutdown capability, low power standby mode
- Uses readily available standard inductors
- Thermal shutdown and current limit protection
- Adjustable version output voltage range, 1.23V to 37V ±4% max over line and load conditions



Applications

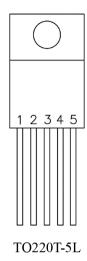
- Simple high-efficiency step-down (buck) regulator
- Efficient pre-regulator for linear regulators
- On-card switching regulators
- Positive to negative converter (Buck-Boost)

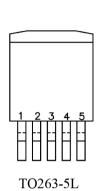
Functional Block Diagram





Pin Configuration





Pin Description

Pin Number	Pin Name	Function Description
1	VIN	This is the positive input supply for the IC switching regulator. A suitable input bypass capacitor must be present at this pin to minimize voltage transients and to supply the switching currents needed by the regulator.
2	OUTPUT	Internal switch, the voltage at this pin switches between (+V _{IN} V _{SAT}) and approximately -0.5V. To minimize coupling to sensitive circuitry, the PC board copper area connected to this pin should be kept to a minimum.
3	GND	Circuit Ground
4	FEEDBACK	Senses the regulated output voltage to complete the feedback loop
5	ON / OFF	Allows the switching regulator circuit to be shut down using logic level signals.

Absolute Maximum Ratings

Parameter Name	Symbol	Value	Unit
Maximum Supply Voltage	$V_{ m IN}$	45	V
<u>ON</u> / OFF pin Input Voltage	\overline{ON} / OFF	-0.3V≤V≤+V _{IN}	V
Output Voltage to Ground (steady state)	V _{OUT}	-1	V
Power Dissipation	P_{DMAX}	Internally Limited	
Storage Temperature Range	Tstg	-65~+150	$^{\circ}$
Maximum Junction Temperature	T_{JA}	150	${\mathbb C}$
ESD Susceptibility (Human Body Model)	ESD	2	kV
Lead Temperature (Soldering, 10 Seconds)	$T_{ m L}$	260	$^{\circ}$



Recommended Operating Conditions

Parameter Name	Symbol	Value	Unit
Supply Voltage	V _{IN}	6~40	V
Operating temperature range	Topr	-40~+125	$^{\circ}$

Electrical Characteristics

(Unless otherwise specified: $T_J = 25^{\circ}C$)

Parameter Name	Symbol	Test Conditions	Min	Тур	Max	Units	
Device Parameters							
Feedback Bias Current	Ib	Adjustable version only, V _{OUT} =5V		50	100	nA	
Oscillator Frequency	fo	(Note 1)	47	52	58	kHz	
V _{SAT} Saturation Voltage	Vsat	I _{OUT} =3A		1.4	1.8	V	
Max. Duty Cycle(ON)	DC		93	98		%	
Current Limit	I_{CL}	(Note 1)	4.2	5.8	6.9	A	
O 45 4 I 1 C	$I_{\rm L}$	Output=0V			2	mA	
Output Leakage Current		Output=-1V		7.5	30	mA	
Quiescent Current	I_Q			5	10	mA	
Standby Quiescent Current	I _{STBY}	$\overline{ON} / OFF $ pin=5V(OFF)		50	200	μА	
ON / OFF Control							
ON / OFF Pin Logic	$V_{ m IH}$	V _{OUT} =0V	2.0			V	
Input Level	$V_{\mathrm{I\!L}}$	Vout=nominal output voltage			0.8	V	
ON COEFFE In the Comment	$ m I_{IH}$	$\overline{ON} / OFF $ pin=5V(OFF)		12	30	μА	
ON / OFF Pin Input Current	$ m I_{IL}$	<u>ON</u> / OFF pin=0V(ON)		0	10	μА	

Note 1: The oscillator frequency reduces to approximately 11 kHz in the event of an output short or an overload which causes the regulated output voltage to drop approximately 40% from the nominal output voltage. This self protection feature lowers the average power dissipation of the IC by lowering the minimum duty cycle from 5% down to approximately 2%.

Output pin sourcing current. No diode, inductor or capacitor connected to output.



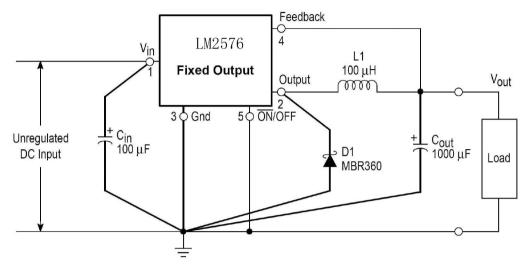
Parameter Name	Symbol	Test Conditions	Min	Тур	Max	Units		
LM2576-3.3V								
		V _{IN} =12V, I _O =500mA	3.234	3.3	3.366	V		
Output Voltage	V _{OUT}	6V≤V _{IN} ≤40V 0.5A≤I _{LOAD} ≤3A	3.168	3.3	3.450	V		
Efficiency	η	$V_{IN}=12V$, $I_{LOAD}=3A$		75		%		
LM2576- 5.0V	LM2576-5.0V							
	V _{OUT}	V _{IN} =12V, I _O =500mA	4.90	5.00	5.10	V		
Output Voltage		$\begin{array}{c} 8V \leq V_{IN} \leq 40V \\ 0.5A \leq I_{LOAD} \leq 3A \end{array}$	4.80	5.00	5.20	V		
Efficiency	η	$V_{IN}=12V$, $I_{LOAD}=3A$		77		%		
LM2576-12V								
		V _{IN} =25V, I _O =500mA	11.76	12.00	12.24	V		
Output Voltage	V _{OUT}	15V≤V _{IN} ≤40V 0.5A≤I _{LOAD} ≤3A	11.52	12.00	12.48	V		
Efficiency	η	$V_{IN}=25V$, $I_{LOAD}=3A$		88		%		

Parameter Name	Symbol	Test Conditions	Min	Тур	Max	Units	
LM2576-ADJ							
0	Vout	V _{IN} =12V,I _O =500mA, V _{OUT} =5V	1.217	1.230	1.243	V	
Output Voltage		$\begin{array}{c} 8V \leq V_{IN} \leq 40V, V_{OUT} = 5V \\ 0.5A \leq I_{LOAD} \leq 3A \end{array}$	1.193	1.230	1.267	V	
Efficiency	η	V _{IN} =12V,I _{LOAD} =3A, V _{OUT} =5V		77		%	



Test Circuit

Fixed Output Voltage Versions



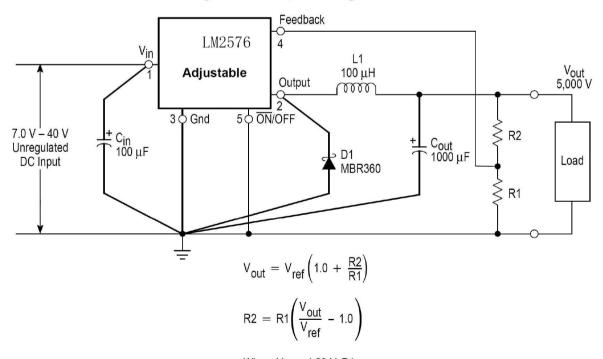
 $C_{in} - 100 \,\mu\text{F}$, 75 V, Aluminium Electrolytic $C_{out} - 1000 \,\mu\text{F}$, 25 V, Aluminium Electrolytic

D1 - Schottky, MBR360

L1 - 100 μ H, Pulse Eng. PE-92108

R1 - 2.0 k, 0.1% R2 - 6.12 k, 0.1%

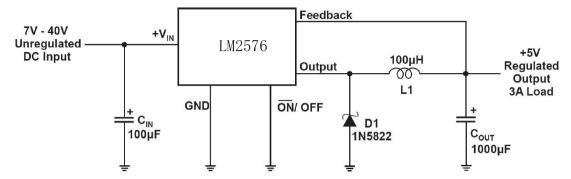
Adjustable Output Voltage Versions



Where $V_{ref} = 1.23 \text{ V}$, R1 between 1.0 k and 5.0 k



Typical Application



Application Information

Input Capacitor (C_{IN})

To maintain stability, the regulator input pin must be bypassed with at least a 100 µF electrolytic capacitor. The capacitor's leads must be kept short, and located near the regulator. If the operating temperature range includes temperatures below -25°C, the input capacitor value may need to be larger. With most electrolytic capacitors, the capacitance value decreases and the ESR increases with lower temperatures and age. Paralleling a ceramic or solid tantalum capacitor will increase the regulator stability at cold temperatures. For maximum capacitor operating lifetime, the capacitor's RMS ripple current rating should be greater than

$$1.2 \times \left(\frac{t_{ON}}{T}\right) \times I_{LOAD}$$

Where
$$\frac{t_{ON}}{T} = \frac{V_{OUT}}{V_{IN}}$$
 for a buck regulator

and
$$\frac{t_{ON}}{T} = \frac{|V_{OUT}|}{|V_{OUT}| + V_{IN}}$$
 for a buck-boost regulator.

Inductor Selection

All switching regulators have two basic modes of operation: continuous and discontinuous. The difference between the two types relates to the inductor current, whether it is flowing continuously, or if it drops to zero for a period of time in the normal switching cycle. Each mode has distinctively different operating characteristics, which can affect the regulator performance and requirements. The LM2576can be used for both continuous and discontinuous modes of operation. When using inductor values shown in the inductor selection guide, the peak-to-peak inductor ripple current will be approximately 20% to 30% of the maximum DC current. With relatively heavy load currents, the circuit operates in the continuous mode (inductor current always flowing), but under light load conditions, the circuit will be forced to the discontinuous mode (inductor current falls to zero for



a period of time). This discontinuous mode of operation is perfectly acceptable. For light loads(less than approximately 300mA) it may be desirable to operate the regulator in the discontinuous mode, primarily because of the lower inductor values required for the discontinuous mode. The selection guide chooses inductor values suitable for continuous mode operation, but if the inductor value chosen is prohibitively high, the designer should investigate the possibility of discontinuous operation.

Inductors are available in different styles such as pot core, toriod, E-frame, bobbin core, etc., as well as different core materials, such as ferrites and powdered iron. The least expensive, the bobbin core type, consists of wire wrapped on a ferrite rod core. This type of construction makes for an inexpensive inductor, but since the magnetic flux is not completely contained within the core, it generates more electromagnetic interference (EMI).

This EMI can cause problems in sensitive circuits, or can give incorrect scope readings because of induced voltages in the scope probe. The inductors listed in the selection chart include ferrite pot core construction for AIE, powdered iron toroid for Pulse Engineering, and ferrite bobbin core for Renco.

An inductor should not be operated beyond its maximum rated current because it may saturate. When an inductor begins to saturate, the inductance decreases rapidly and the inductor begins to look mainly resistive (the DC resistance of the winding) This will cause the switch current to rise very rapidly. Different inductor types have different saturation characteristics, and this should be kept in mind when selecting an inductor.

The inductor manufacturer's data sheets include current and energy limits to avoid inductor saturation.

Inductor Ripple Current

When the switcher is operating in the continuous mode, the inductor current waveform ranges from a triangular to a sawtooth type of waveform (depending on the input voltage). For a given input voltage and output voltage, the peak-to-peak amplitude of this inductor current waveform remains constant. As the load current rises or falls, the entire sawtooth current waveform also rises or falls. The average DC value of this waveform is equal to the DC load current (in the buck regulator configuration). If the load current drops to a low enough level, the bottom of the sawtooth current waveform will reach zero, and the switcher will change to a discontinuous mode of operation. This is a perfectly acceptable mode of operation. Any buck switching regulator (no matter how large the inductor value is) will be forced to run discontinuous if the load current is light enough.

Catch Diode

Buck regulators require a diode to provide a return path for the inductor current when the switch is off. This diode should be located close to the LM2576using short leads and short printed circuit traces.

Because of their fastswitching speed and low forward voltage drop, Schottky diodes provide the best efficiency, especially in low

output voltage switching regulators (less than 5V). Fast-Recovery, High-Efficiency, or Ultra-Fast Recovery diodes are also suitable, but some types with an abrupt turn-off characteristic may cause instability and EMI problems. A fast-recovery diode with soft recovery characteristics is a better choice. Standard 60 Hz diodes (e.g.,



1N4001 or 1N5400, etc.) are also not suitable.

Output Capacitor

An output capacitor is required to filter the output voltage and is needed for loop stability. The capacitor should be located near the LM2576using short pc board traces. Standard aluminum electrolytics are usually adequate, but low ESR types are recommended for low output ripple voltage and good stability. The ESR of a capacitor depends on many factors, some which are: the value, the voltage rating, physical size and the type of construction. In general, low value or low voltage (less than 12V) electrolytic capacitors usually have higher ESR numbers.

The amount of output ripple voltage is primarily a function of the ESR (Equivalent Series Resistance) of the output capacitor and the amplitude of the inductor ripple current (ΔI_{IND}). See the section on inductor ripple current in Application Hints. The lower capacitor values ($220~\mu F$ – $1000~\mu F$) will allow typically 50 mV to 150 mV of output ripple voltage, while larger-value capacitors will reduce the ripple to approximately 20 mV to 50mV. Output Ripple Voltage = (ΔI_{IND}) (ESR of C_{OUT}). To further reduce the output ripple voltage, several standard electrolytic capacitors may be paralleled, or a higher-grade capacitor may be used. Such capacitors are often called "high-frequency" "low-inductance" or "low-ESR" These will reduce the output ripple to 10 mV or 20 mV. However, when operating in the continuous mode, reducing the ESR below 0.03Ω can cause instability in the regulator. Tantalum capacitors can have a very low ESR, and should be carefully evaluated if it is the only output capacitor. Because of their good low temperature characteristics, a tantalum can be used in parallel with aluminum electrolytics, with the tantalum making up 10% or 20% of the total capacitance. The capacitor's ripple current rating at 52 kHz should be at least 50% higher than the peak-to-peak inductor ripple current.

Output Voltage Ripple and Transients

The output voltage of a switching power supply will contain a sawtooth ripple voltage at the switcher frequency, typically about 1% of the output voltage, and may also contain short voltage spikes at the peaks of the sawtooth waveform. The output ripple voltage is due mainly to the inductor sawtooth ripple current multiplied by the ESR of the output capacitor. The voltage spikes are present because of the fast switching action of the output switch, and the parasitic inductance of the output filter capacitor. To minimize these voltage spikes, special low inductance capacitors can be used, and their lead lengths must be kept short. Wiring inductance, stray capacitance, as well as the scope probe used to evaluate these transients, all contribute to the amplitude of these spikes. An additional small LC filter ($20~\mu\text{H} & 100~\mu\text{F}$) can be added to the output to further reduce the amount of output ripple and transients. A 10~x reduction in output ripple voltage and transients is possible with this filter.



Feedback Connection

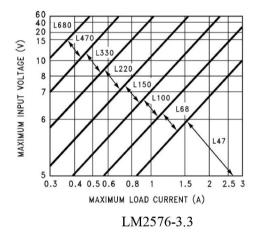
The LM2576(fixed voltage versions)

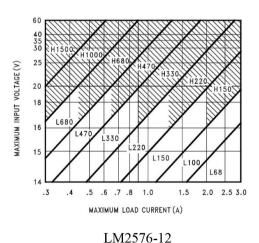
feedback pin must be wired to the output voltage point of the switching power supply.

When using the adjustable version, physically locate both output voltage programming resistors near the LM2576to avoid picking up unwanted noise. Avoid using resistors greater than 100k Ω because of the increased chance of noise pickup. ON/OFF Input

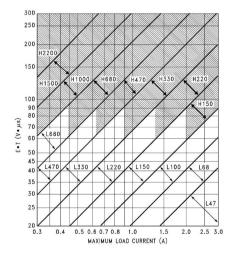
For normal operation, the \overline{ON}/OFF pin should be grounded or driven with a low-level TTL voltage (typically below 1.6V). To put the regulator into standby mode, drive this pin with a high-level TTL or CMOS signal. The \overline{ON}/OFF pin can be safely pulled up to $+V_{IN}$ without a resistor in series with it. The \overline{ON}/OFF pin should not be left open.

Inductor Value Selection Guides (For Continuous Mode Operation)





LM2576-5.0



LM2576-ADJ



Inductor Code	Inductor Value	Schott	Pulse Eng.	Renco
L47	47 µH	671 26980	PE-53112	RL2442
L68	68 µH	671 26990	PE-92114	RL2443
L100	100 µH	671 27000	PE-92108	RL2444
L150	150 µH	671 27010	PE-53113	RL1954
L220	220 µH	671 27020	PE-52626	RL1953
L330	330 µH	671 27030	PE-52627	RL1952
L470	470 µH	671 27040	PE-53114	RL1951
L680	680 µH	671 27050	PE-52629	RL1950
H150	150 µH	671 27060	PE-53115	RL2445
H220	220 µH	671 27070	PE-53116	RL2446
H330	330 µH	671 27080	PE-53117	RL2447
H470	470 µH	671 27090	PE-53118	RL1961
H680	680 µH	671 27100	PE-53119	RL1960
H1000	1000 µH	671 27110	PE-53120	RL1959
H1500	1500 µH	671 27120	PE-53121	RL1958
H2200	2200 µH	671 27130	PE-53122	RL2448

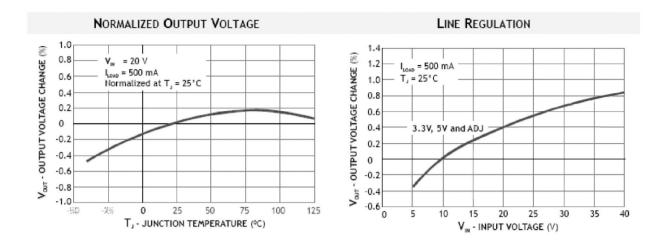
Inductor Selection Guide

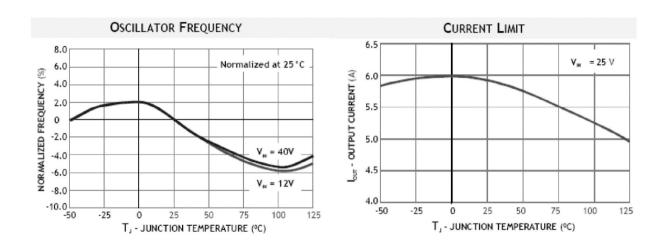
V _R	Sch	ottky	Fast Re	ecovery
	3A	4A-6A	3A	4A-6A
20V	1N5820	1N5823		
	MBR320P			
	SR302			
30V	1N5821	50WQ03		
	MBR330	1N5824		
	31DQ03			The following
	SR303		The following	diodes are all
40V	1N5822	MBR340	diodes are all	rated to 100V
	MBR340	50WQ04	rated to 100V	
	31DQ04	1N5825	31DF1	50WF10
	SR304		HER302	MUR410
50V	MBR350	50WQ05		HER602
	31DQ05			
	SR305			
60V	MBR360	50WR06		
	DQ06	50SQ060		
	SR306		2	

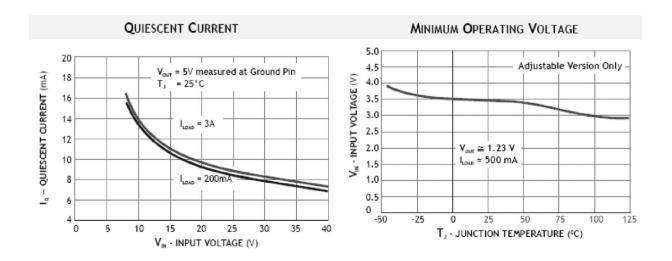
Diode Selection Guide



Characteristic Curves









Outline Dimensions

