

# S-1214 Series

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# 105°C OPERATION, 36 V INPUT, 1000 mA VOLTAGE REGULATOR

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The S-1214 Series developed by using high-withstand voltage CMOS process technology, is a positive voltage regulator with a high-withstand voltage, low current consumption and high-accuracy output voltage.

The S-1214 Series operates at the maximum operation voltage of 36 V and a low current consumption of  $5.0~\mu A$  typ. and has a built-in low on-resistance output transistor, which provides a very small dropout voltage and a large output current. In addition to the type in which output voltage is set inside the IC, the type for which output voltage can be set via an external resistor is added to a lineup. Also, a built-in overcurrent protection circuit to limit overcurrent of the output transistor and a built-in thermal shutdown circuit to limit heat are included.

#### **■** Features

Output voltage (internally set):
Output voltage (externally set):
1.8 V, 3.0 V, 3.3 V, 5.0 V, 8.0 V, 12.0 V, 15.0 V
1.8 V to 30.0 V, settable via external resistor

Input voltage:
 Output voltage accuracy:
 2.8 V to 36.0 V
 ±1.0% (Ta = +25°C)

• Current consumption: During operation:  $5.0 \mu A \text{ typ.}(\text{Ta} = +25^{\circ}\text{C})$ 

During power-off:  $0.1 \mu A \text{ typ.}(\text{Ta} = +25^{\circ}\text{C})$ 

• Output current: Possible to output 1000 mA (at  $V_{IN} \ge V_{OUT(S)} + 2.0 \text{ V})^{*1}$ • Input and output capacitors: A ceramic capacitor can be used. (1.0  $\mu$ F or more)

• Built-in overcurrent protection circuit: Limits overcurrent of output transistor.

(with a detection function of the difference between input and output voltage)

• Built-in thermal shutdown circuit: Detection temperature 170°C typ.

Built-in ON / OFF circuit:
 Ensures long battery life.

Discharge shunt function is available. Pull-down function is available.

• Operation temperature range: Ta = -40°C to +105°C

• Lead-free (Sn 100%), halogen-free

- \*1. Please make sure that the loss of the IC will not exceed the power dissipation when the output current is large.
- \*2. Contact our sales representatives for details.

# ■ Applications

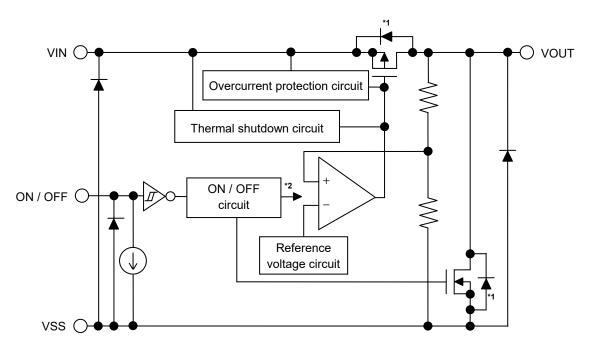
- Constant-voltage power supply for industrial equipment
- Constant-voltage power supply for home electric appliance

#### Packages

- TO-252-5S(A)
- HSOP-8A

# ■ Block Diagrams

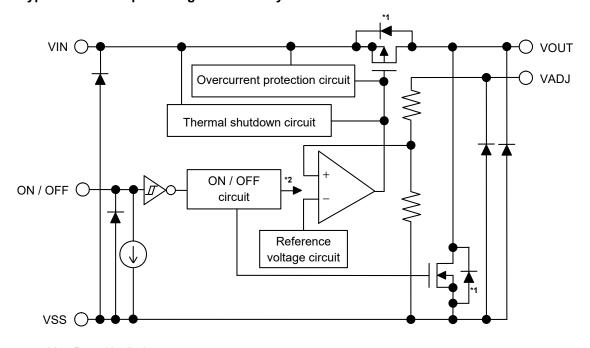
# 1. Type in which output voltage is internally set



- \*1. Parasitic diode
- **\*2.** The ON / OFF circuit controls the internal circuit and the output transistor.

Figure 1

## 2. Type in which output voltage is externally set



- \*1. Parasitic diode
- \*2. The ON / OFF circuit controls the internal circuit and the output transistor.

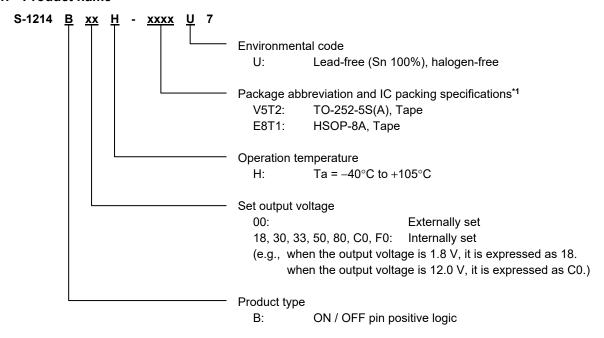
Figure 2

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#### ■ Product Name Structure

Users can select the output voltage and package type for the S-1214 Series. Refer to "1. Product name" regarding the contents of product name, "2. Packages" regarding the package drawings and "3. Product name list" for details of product names.

## 1. Product name



\*1. Refer to the tape drawing.

## 2. Packages

Table 1 Package Drawing Codes

Package Name	Dimension	Tape	Reel	Land
TO-252-5S(A)	VA005-A-P-SD	VA005-A-C-SD	VA005-A-R-SD	VA005-A-L-SD
HSOP-8A	FH008-A-P-SD	FH008-A-C-SD	FH008-A-R-SD	FH008-A-L-SD

## 3. Product name list

Table 2

Output Voltage	TO-252-5S(A)	HSOP-8A
Externally set	S-1214B00H-V5T2U7	S-1214B00H-E8T1U7
1.8 V ± 1.0%	S-1214B18H-V5T2U7	S-1214B18H-E8T1U7
3.0 V ± 1.0%	S-1214B30H-V5T2U7	S-1214B30H-E8T1U7
3.3 V ± 1.0%	S-1214B33H-V5T2U7	S-1214B33H-E8T1U7
5.0 V ± 1.0%	S-1214B50H-V5T2U7	S-1214B50H-E8T1U7
8.0 V ± 1.0%	S-1214B80H-V5T2U7	S-1214B80H-E8T1U7
12.0 V ± 1.0%	S-1214BC0H-V5T2U7	S-1214BC0H-E8T1U7
15.0 V ± 1.0%	S-1214BF0H-V5T2U7	S-1214BF0H-E8T1U7

**Remark** Please contact our sales representatives for products other than the above.

# **■** Pin Configurations

# 1. TO-252-5S(A)

Top view

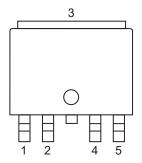


Figure 3

Table 3 Type in Which Output Voltage is Internally Set

Pin No.	Symbol	Description
1	VIN	Input voltage pin
2	ON / OFF	ON / OFF pin
3	VSS	GND pin
4	NC*1	No connection
5	VOUT	Output voltage pin

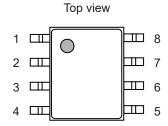
Table 4 Type in Which Output Voltage is Externally Set

Pin No.	Symbol	Description
1	VIN	Input voltage pin
2	ON / OFF	ON / OFF pin
3	VSS	GND pin
4	VADJ	Output voltage adjustment pin
5	VOUT	Output voltage pin

<sup>\*1.</sup> The NC pin is electrically open.

The NC pin can be connected to the VIN pin or the VSS pin.

#### 2. HSOP-8A



Bottom view

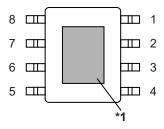


Figure 4

Table 5 Type in Which Output Voltage is Internally Set

Pin No.	Symbol	Description
1	VIN	Input voltage pin
2	NC*2	No connection
3	NC*2	No connection
4	ON / OFF	ON / OFF pin
5	VSS	GND pin
6	NC*2	No connection
7	NC*2	No connection
8	VOUT	Output voltage pin

Table 6 Type in Which Output Voltage is Externally Set

Pin No.	Symbol	Description
1	VIN	Input voltage pin
2	NC*2	No connection
3	NC*2	No connection
4	ON / OFF	ON / OFF pin
5	VSS	GND pin
6	NC*2	No connection
7	VADJ	Output voltage adjustment pin
8	VOUT	Output voltage pin

**<sup>\*1.</sup>** Connect the heat sink of backside at shadowed area to the board, and set electric potential GND. However, do not use it as the function of electrode.

The NC pin can be connected to the VIN pin or the VSS pin.

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**<sup>\*2.</sup>** The NC pin is electrically open.

# ■ Absolute Maximum Ratings

Table 7

(Ta = +25°C unless otherwise specified)

Item	Symbol	Absolute Maximum Rating	Unit
	V <sub>IN</sub>	$V_{SS} - 0.3 \text{ to } V_{SS} + 45$	V
Input voltage	V <sub>ON / OFF</sub>	$V_{SS} - 0.3$ to $V_{SS} + 45$	V
	V <sub>VADJ</sub>	$V_{SS} - 0.3 \text{ to } V_{SS} + 45$	V
Output voltage	Vout	$V_{SS} - 0.3 \text{ to } V_{IN} + 0.3 \le V_{SS} + 45$	V
Output current	I <sub>OUT</sub>	1100	mA
Junction temperature	Tj	-40 to +150	°C
Operation ambient temperature	Topr	-40 to +105	°C
Storage temperature	T <sub>stg</sub>	-40 to +150	°C

Caution The absolute maximum ratings are rated values exceeding which the product could suffer physical damage. These values must therefore not be exceeded under any conditions.

# **■** Thermal Resistance Value

Table 8

Item	Symbol	Condition		Min.	Тур.	Max.	Unit
			Board A	_	86	_	°C/W
			Board B	_	ı	_	°C/W
	θја	TO-252-5S(A)	Board C	_	ı	_	°C/W
			Board D	_	ı	_	°C/W
Junction-to-ambient thermal resistance*1, *2			Board E	_	24	_	°C/W
Junction-to-ambient thermal resistance ", -			Board A	_	104	_	°C/W
			Board B	_	ı	_	°C/W
		HSOP-8A	Board C	_	ı	_	°C/W
			Board D	_	_	_	°C/W
			Board E	_	30	_	°C/W

<sup>\*1.</sup> Test environment: compliance with JEDEC STANDARD JESD51-2A

Remark Refer to "■ Power Dissipation" and "Test Board" for details.

<sup>\*2.</sup> Measurement values when this IC is mounted on each board

# ■ Electrical Characteristics

# 1. Type in which output voltage is internally set

Table 9 (1 / 2)

				(Ta = +2	5°C unle	ss other	wise s	pecified
Item	Symbol	Co	ondition	Min.	Тур.	Max.	Unit	Test Circuit
		V <sub>IN</sub> = 13.5 V, I <sub>OUT</sub> = 10 mA	V <sub>OUT(S)</sub> = 1.8 V, 3.0 V, 3.3 V, 5.0 V, 8.0 V, 12.0 V	V <sub>OUT(S)</sub> × 0.99	V <sub>OUT(S)</sub>	V <sub>OUT(S)</sub> × 1.01	V	1
		$V_{IN} = V_{OUT(S)} + 1.0 V,$ $I_{OUT} = 10 \text{ mA}$	V <sub>OUT(S)</sub> = 15.0 V	V <sub>OUT(S)</sub> × 0.99	V <sub>OUT(S)</sub>	V <sub>OUT(S)</sub> × 1.01	V	1
Output voltage* <sup>1</sup>	V <sub>OUT(E)</sub>	$V_{IN}$ = 13.5 V, $I_{OUT}$ = 10 mA, $-40^{\circ}C \le Ta \le +105^{\circ}C^{*4}$	V <sub>OUT(S)</sub> = 1.8 V, 3.0 V, 3.3 V, 5.0 V, 8.0 V, 12.0 V	V <sub>OUT(S)</sub> × 0.98	V <sub>OUT(S)</sub>	V <sub>OUT(S)</sub> × 1.02	V	1
		$V_{IN} = V_{OUT(S)} + 1.0 \text{ V},$ $I_{OUT} = 10 \text{ mA},$ $-40^{\circ}\text{C} \le \text{Ta} \le +105^{\circ}\text{C}^{*4}$	V <sub>OUT(S)</sub> = 15.0 V	V <sub>OUT(S)</sub> × 0.98	V <sub>OUT(S)</sub>	V <sub>OUT(S)</sub> × 1.02	٧	1
Output current*2	Іоит	$V_{OUT(S)} + 2.0 \text{ V} \leq V_{IN}$		1000*5	_	_	mA	3
		, ,	V <sub>OUT(S)</sub> = 1.8 V	_	*6	_	V	1
			V <sub>OUT(S)</sub> = 3.0 V, 3.3 V	_	0.16	0.32	V	1
Dropout voltage*3	V <sub>drop</sub>	Іоит = 200 mA	V <sub>OUT(S)</sub> = 5.0 V, 8.0 V, 12.0 V, 15.0 V	-	0.13	0.22	٧	1
Line regulation	$\frac{\Delta V_{\text{OUT1}}}{\Delta V_{\text{IN}} \bullet V_{\text{OUT}}}$	$V_{OUT(S)} + 1.0 \text{ V} \le V_{IN} \le 2$		-	0.015	0.03	%/V	1
			$V_{OUT(S)} = 1.8 \text{ V}, 3.0 \text{ V}, 3.3 \text{ V}$	_	15	30	mV	1
		$V_{\text{IN}} = V_{\text{OUT(S)}} + 1.0 \text{ V},$ $1 \text{ mA} \leq I_{\text{OUT}} \leq 300 \text{ mA}$	V <sub>OUT(S)</sub> = 5.0 V	_	20	40	mV	1
Load regulation	$\Delta V_{OUT2}$		V <sub>OUT(S)</sub> = 8.0 V	_	35	70	mV	1
			V <sub>OUT(S)</sub> = 12.0 V, 15.0 V	_	45	90	mV	1
Current consumption during operation	loor	V <sub>IN</sub> = 13.5 V, I <sub>OUT</sub> = 10 μA, ON / OFF pin = ON	V <sub>OUT(S)</sub> = 1.8 V, 3.0 V, 3.3 V, 5.0 V, 8.0 V, 12.0 V	-	5.0	9.8	μΑ	2
	1551	$V_{IN} = V_{OUT(S)} + 1.0 V$ , $I_{OUT} = 10 \mu A$ , ON / OFF pin = ON	V <sub>OUT(S)</sub> = 15.0 V	-	5.0	9.8	μΑ	2
Our and a second star		$V_{IN}$ = 13.5 V, no load, ON / OFF pin = OFF	V <sub>OUT(S)</sub> = 1.8 V, 3.0 V, 3.3 V, 5.0 V, 8.0 V, 12.0 V	_	0.1	0.5	μА	2
Current consumption during power-off	Iss2	$V_{IN} = V_{OUT(S)} + 1.0 V$ , no load, ON / OFF pin = OFF	V <sub>OUT(S)</sub> = 15.0 V	_	0.1	0.5	μΑ	2
Input voltage	VIN		_	2.8	_	36.0	V	_
ON / OFF pin		$V_{IN}$ = 13.5 V, $R_L$ = 1.0 k $\Omega$ , determined by $V_{OUT}$ output level	V <sub>OUT(S)</sub> = 1.8 V, 3.0 V, 3.3 V, 5.0 V, 8.0 V, 12.0 V	2.0	-	_	V	4
input voltage "H"	VsH	$V_{\text{IN}} = V_{\text{OUT(S)}} + 1.0 \text{ V},$ $R_{\text{L}} = 1.0 \text{ k}\Omega,$ determined by $V_{\text{OUT}}$ output level	V <sub>OUT(S)</sub> = 15.0 V	2.0	_	_	٧	4
ON / OFF pin	Vo	$V_{IN}$ = 13.5 V, $R_L$ = 1.0 k $\Omega$ , determined by $V_{OUT}$ output level	V <sub>OUT(S)</sub> = 1.8 V, 3.0 V, 3.3 V, 5.0 V, 8.0 V, 12.0 V	-	-	0.8	٧	4
input voltage "L"	VsL	$V_{\text{IN}} = V_{\text{OUT(S)}} + 1.0 \text{ V},$ $R_{\text{L}} = 1.0 \text{ k}\Omega,$ determined by $V_{\text{OUT}}$ output level	V <sub>OUT(S)</sub> = 15.0 V	-	-	0.8	٧	4

#### Table 9 (2 / 2)

(Ta = +25°C unless otherwise specified)

				(14 - 12	o o unio			,
Item	Symbol	С	condition	Min.	Тур.	Max.	Unit	Test Circuit
ON / OFF pin		V <sub>IN</sub> = 13.5 V, V <sub>ON / OFF</sub> = V <sub>IN</sub>	V <sub>OUT(S)</sub> = 1.8 V, 3.0 V, 3.3 V, 5.0 V, 8.0 V, 12.0 V	-0.1	0.3	1.0	μΑ	4
input current "H"	Isн	$V_{IN} = V_{OUT(S)} + 1.0 V,$ $V_{ON/OFF} = V_{IN}$	V <sub>OUT(S)</sub> = 15.0 V	-0.1	0.3	1.0	μΑ	4
ON / OFF pin	le	V <sub>IN</sub> = 13.5 V, V <sub>ON / OFF</sub> = 0 V	V <sub>OUT(S)</sub> = 1.8 V, 3.0 V, 3.3 V, 5.0 V, 8.0 V, 12.0 V	-0.1	_	0.1	μΑ	4
input current "L"	IsL	$V_{IN} = V_{OUT(S)} + 1.0 \text{ V},$ $V_{ON/OFF} = 0 \text{ V}$	V <sub>OUT(S)</sub> = 15.0 V	-0.1	_	0.1	μΑ	4
		V <sub>IN</sub> = 13.5 V, f = 100 Hz,	V <sub>OUT(S)</sub> = 1.8 V, 3.0 V, 3.3 V, 5.0 V	_	60	_	dB	5
Dipple rejection	IDDI	$\Delta V_{rip} = 0.5 \text{ Vrms},$ $I_{OUT} = 100 \text{ mA}$	V <sub>OUT(S)</sub> = 8.0 V, 12.0 V	_	50	_	dB	5
Ripple rejection  RR	$V_{IN} = V_{OUT(S)} + 1.0 \text{ V},$ f = 100  Hz, $\Delta V_{rip} = 0.5 \text{ Vrms},$ $I_{OUT} = 100 \text{ mA}$	V <sub>OUT(S)</sub> = 15.0 V	_	50	-	dB	5	
Short-circuit current	I <sub>short</sub>	$V_{IN} = V_{OUT(S)} + 1.0 \text{ V}, \text{ C}$ $V_{OUT} = 0 \text{ V}$	ON / OFF pin = ON,	_	250	_	mA	3
Thermal shutdown detection temperature	T <sub>SD</sub>	Junction temperature		-	170	-	°C	-
Thermal shutdown release temperature	T <sub>SR</sub>	Junction temperature		_	135	-	°C	_
Discharge shunt	D	V <sub>IN</sub> = 13.5 V, ON / OFF pin = OFF V <sub>OUT</sub> = 0.1 V	V <sub>OUT(S)</sub> = 1.8 V, 3.0 V, 3.3 V, 5.0 V, 8.0 V, 12.0 V	_	3.0	_	kΩ	6
resistance during power-off	R <sub>LOW</sub>	$V_{IN} = V_{OUT(S)} + 1.0 \text{ V},$ ON / OFF pin = OFF $V_{OUT} = 0.1 \text{ V}$	V <sub>OUT(S)</sub> = 15.0 V	_	3.0	-	kΩ	6

**<sup>\*1.</sup>**  $V_{OUT(S)}$ : Set output voltage

V<sub>OUT(E)</sub>: Actual output voltage

 $V_{\text{IN1}}$  is the input voltage at which the output voltage becomes 98% of  $V_{\text{OUT3}}$  after gradually decreasing the input voltage.  $V_{\text{OUT3}}$  is the output voltage when  $V_{\text{IN}} = V_{\text{OUT(S)}} + 1.0 \text{ V}$  and  $I_{\text{OUT}} = 200 \text{ mA}$ .

- **\*4.** The specification for this temperature range is guaranteed by design, not tested in production.
- **\*5.** Due to limitation of the power dissipation, this value may not be satisfied. Attention should be paid to the power dissipation when the output current is large.
  - This specification is guaranteed by design.
- \*6. The dropout voltage is limited by the difference between the input voltage (min. value) and the set output voltage. In case of 1.8 V  $\leq$  V<sub>OUT(S)</sub> < 2.8 V  $\cdot$  V<sub>OUT(S)</sub> = V<sub>drop</sub>

<sup>\*2.</sup> The output current at which the output voltage becomes 95% of V<sub>OUT(E)</sub> after gradually increasing the output current.

<sup>\*3.</sup>  $V_{drop} = V_{IN1} - (V_{OUT3} \times 0.98)$ 

#### 2. Type in which output voltage is externally set

Table 10

(Ta = +25°C, V<sub>OUT</sub> = V<sub>VADJ</sub> unless otherwise specified)

$(1a = +25^{\circ}C, V_{OUT} = V_{VADJ} \text{ unless otherwise specified})$							
Item	Symbol	Condition	Min.	Тур.	Max.	Unit	Test Circuit
Adjustment pin	.,	V <sub>IN</sub> = 13.5 V, I <sub>OUT</sub> = 10 mA	1.782	1.8	1.818	٧	7
output voltage*1	Vvadj	$V_{IN} = 13.5 \text{ V}, I_{OUT} = 10 \text{ mA}, -40^{\circ}\text{C} \le \text{Ta} \le +105^{\circ}\text{C}^{*4}$	1.764	1.8	1.836	V	7
Output voltage range	V <sub>ROUT</sub>	-	1.8	-	30.0	V	13
Adjustment pin internal resistance	Rvadj	-	_	26	_	МΩ	-
Output current*2	Гоит	$3.8 \text{ V} \leq \text{V}_{\text{IN}}$	1000*5	-	-	mA	9
Dropout voltage*3	$V_{drop}$	I <sub>OUT</sub> = 200 mA	-	*6	-	V	7
Line regulation	$\frac{\Delta V_{\text{OUT1}}}{\Delta V_{\text{IN}} \bullet V_{\text{OUT}}}$	$2.8 \text{ V} \le \text{V}_{\text{IN}} \le 28.0 \text{ V}, \text{I}_{\text{OUT}} = 10 \text{ mA}$	_	0.015	0.03	%/V	7
Load regulation	$\Delta V_{OUT2}$	$V_{IN} = 2.8 \text{ V}, 1 \text{ mA} \le I_{OUT} \le 300 \text{ mA}$	_	15	30	mV	7
Current consumption during operation	I <sub>SS1</sub>	$V_{IN}$ = 13.5 V, ON / OFF pin = ON, $I_{OUT}$ = 10 $\mu$ A	-	5.0	9.8	μА	8
Current consumption during power-off	I <sub>SS2</sub>	V <sub>IN</sub> = 13.5 V, ON / OFF pin = OFF, no load	_	0.1	0.5	μА	8
Input voltage	V <sub>IN</sub>	-	2.8	_	36.0	V	_
ON / OFF pin	V	$V_{IN} = 13.5 \text{ V}, R_L = 1.0 \text{ k}\Omega,$	2.0			V	10
input voltage "H"	V <sub>SH</sub>	determined by Vout output level	2.0	_	_	V	10
ON / OFF pin input voltage "L"	VsL	$V_{\text{IN}}$ = 13.5 V, $R_{\text{L}}$ = 1.0 k $\Omega$ , determined by $V_{\text{OUT}}$ output level	-	_	0.8	V	10
ON / OFF pin input current "H"	Isн	V <sub>IN</sub> = 13.5 V, V <sub>ON / OFF</sub> = V <sub>IN</sub>	-0.1	0.3	1.0	μА	10
ON / OFF pin input current "L"	IsL	V <sub>IN</sub> = 13.5 V, V <sub>ON/OFF</sub> = 0 V	-0.1	-	0.1	μА	10
Ripple rejection	RR	$V_{IN}$ = 13.5 V, f = 100 Hz, $\Delta V_{rip}$ = 0.5 Vrms, $I_{OUT}$ = 100 mA	_	60	_	dB	11
Short-circuit current	I <sub>short</sub>	$V_{IN}$ = 2.8 V, ON / OFF pin = ON, $V_{OUT}$ = 0 V	_	250	_	mA	9
Thermal shutdown detection temperature	T <sub>SD</sub>	Junction temperature	_	170	_	°C	_
Thermal shutdown release temperature	T <sub>SR</sub>	Junction temperature	-	135	-	°C	-
Discharge shunt resistance during power-off	R <sub>LOW</sub>	V <sub>IN</sub> = 13.5 V, ON / OFF pin = OFF, V <sub>OUT</sub> = 0.1 V	_	3.0	_	kΩ	12

- \*1. V<sub>OUT(S)</sub>: Set output voltage = 1.8 V
- \*2. The output current at which the output voltage becomes 95% of VOUT(E) after gradually increasing the output current.
- \*3.  $V_{drop} = V_{IN1} (V_{OUT3} \times 0.98)$ 
  - $V_{IN1}$  is the input voltage at which the output voltage becomes 98% of  $V_{OUT3}$  after gradually decreasing the input voltage.  $V_{OUT3}$  is the output voltage when  $V_{IN} = V_{OUT(S)} + 1.0 \text{ V}$  and  $I_{OUT} = 200 \text{ mA}$ .
- **\*4.** The specification for this temperature range is guaranteed by design, not tested in production.
- \*5. Due to limitation of the power dissipation, this value may not be satisfied. Attention should be paid to the power dissipation when the output current is large.
  - This specification is guaranteed by design.
- **\*6.** The dropout voltage is limited by the difference between the input voltage (min. value) and the set output voltage. In case of 1.8 V  $\leq$  V<sub>OUT(S)</sub> < 2.8 V V<sub>OUT(S)</sub> = V<sub>drop</sub>

# **■** Test Circuits

# 1. Type in which output voltage is internally set

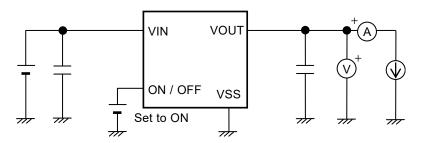


Figure 5 Test Circuit 1

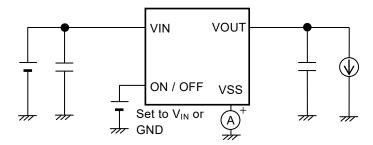


Figure 6 Test Circuit 2

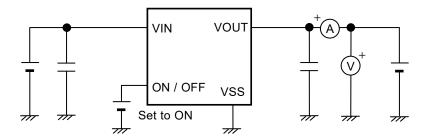


Figure 7 Test Circuit 3

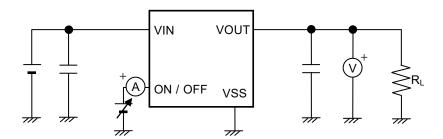


Figure 8 Test Circuit 4

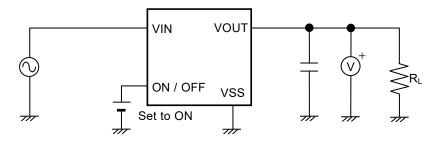


Figure 9 Test Circuit 5

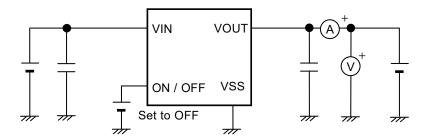


Figure 10 Test Circuit 6

# 2. Type in which output voltage is externally set

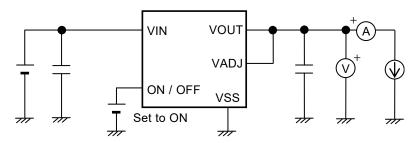


Figure 11 Test Circuit 7

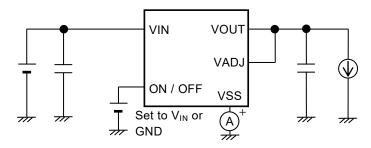


Figure 12 Test Circuit 8

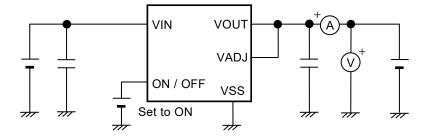


Figure 13 Test Circuit 9

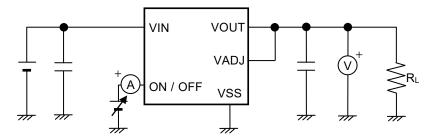


Figure 14 Test Circuit 10

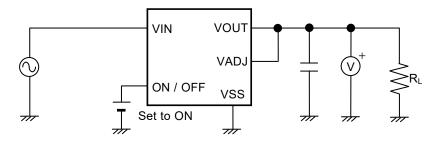


Figure 15 Test Circuit 11

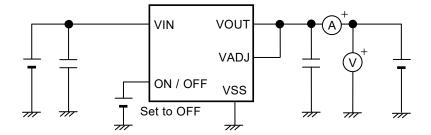


Figure 16 Test Circuit 12

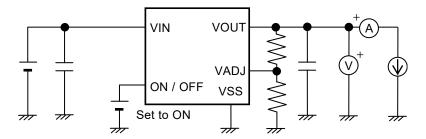
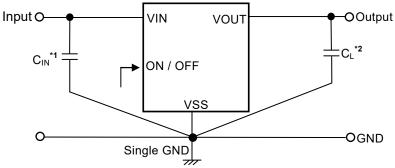


Figure 17 Test Circuit 13

# ■ Standard Circuits

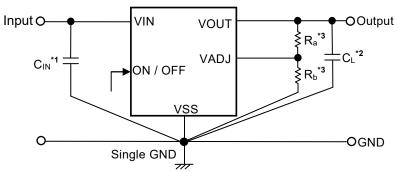
# 1. Type in which output voltage is internally set



- \*1. C<sub>IN</sub> is a capacitor for stabilizing the input.
- \*2. C<sub>L</sub> is a capacitor for stabilizing the output.

Figure 18

## 2. Type in which output voltage is externally set



- \*1. C<sub>IN</sub> is a capacitor for stabilizing the input.
- \*2. C<sub>L</sub> is a capacitor for stabilizing the output.
- \*3.  $R_a$  and  $R_b$  are resistors for output voltage external setting.

Figure 19

Caution The above connection diagrams and constants will not guarantee successful operation. Perform thorough evaluation including the temperature characteristics with an actual application to set the constants.

# ■ Condition of Application

Input capacitor ( $C_{IN}$ ): A ceramic capacitor with capacitance of 1.0  $\mu F$  or more is recommended. Output capacitor ( $C_L$ ): A ceramic capacitor with capacitance of 1.0  $\mu F$  or more is recommended. ESR of output capacitor: A ceramic capacitor with capacitance of 100  $\Omega$  or less is recommended.

Caution Generally, in a voltage regulator, an oscillation may occur depending on the selection of the external parts. Perform thorough evaluation including the temperature characteristics with an actual application using the above capacitors to confirm no oscillation occurs.

# ■ Selection of Input Capacitor (C<sub>IN</sub>) and Output Capacitor (C<sub>L</sub>)

The S-1214 Series requires  $C_L$  between the VOUT pin and the VSS pin for phase compensation. The operation is stabilized by a ceramic capacitor with capacitance of 1.0  $\mu F$  or more. When using an OS capacitor, a tantalum capacitor or an aluminum electrolytic capacitor, the capacitance also must be 1.0  $\mu F$  or more. However, an oscillation may occur depending on the equivalent series resistance (ESR).

Moreover, the S-1214 Series requires C<sub>IN</sub> between the VIN pin and the VSS pin for a stable operation.

Generally, an oscillaiton may occur when a voltage regulator is used under the conditon that the impedance of the power supply is high.

Note that the output voltage transient characteristics varies depending on the capacitance of C<sub>IN</sub> and C<sub>L</sub> and the value of FSR.

Caution Perform thorough evaluation including the temperature characteristics with an actual application to select  $C_{\text{IN}}$  and  $C_{\text{L}}$ .

## ■ Selection of Resistors (Ra, Rb) for Output Voltage External Setting

The S-1214 Series provides the type in which output voltage can be set via the external resistor. The output voltage can be set by connecting a resistor ( $R_a$ ) between the VOUT pin and the VADJ pin, and a resistor ( $R_b$ ) between the VADJ pin and the VSS pin.

Depending on the intended output voltage, select  $R_a$  and  $R_b$  from the range shown in **Table 11**.

Caution Since the VADJ pin impedance is comparatively high and is easily affected by noise, pay adequate attention to the wiring pattern.

Table 11

Vout	Ra	R♭
1.8 V	Connect to VOUT pin	Unnecessary
1.85 V to 30.0 V	$0.25~\text{k}\Omega$ to $2.6~\text{M}\Omega$	10 k $\Omega$ to 200 k $\Omega$

# **■** Explanation of Terms

#### 1. Low dropout voltage regulator

This is a voltage regulator which made dropout voltage small by its built-in low on-resistance output transistor.

## 2. Output voltage (Vout)

This voltage is output at an accuracy of  $\pm 1.0\%$  when the input voltage, the output current and the temperature are in a certain condition\*1.

\*1. Differs depending on the product.

Caution If the certain condition is not satisfied, the output voltage may exceed the accuracy range of ±1.0%. Refer to "■ Electrical Characteristics" and "■ Characteristics (Typical Data)" for details.

3. Line regulation 
$$\left(\frac{\Delta V_{\text{OUT1}}}{\Delta V_{\text{IN}} \bullet V_{\text{OUT}}}\right)$$

Indicates the dependency of the output voltage against the input voltage. That is, the value shows how much the output voltage changes due to a change in the input voltage after fixing output current constant.

# 4. Load regulation (ΔV<sub>OUT2</sub>)

Indicates the dependency of the output voltage against the output current. That is, the value shows how much the output voltage changes due to a change in the output current after fixing input voltage constant.

#### 5. Dropout voltage (V<sub>drop</sub>)

Indicates the difference between input voltage ( $V_{IN1}$ ) and the output voltage when the output voltage becomes 98% of the output voltage value ( $V_{OUT3}$ ) at  $V_{IN} = V_{OUT(S)} + 1.0$  V after the input voltage ( $V_{IN}$ ) is decreased gradually.

$$V_{drop} = V_{IN1} - (V_{OUT3} \times 0.98)$$

## Operation

#### 1. Basic operation

Figure 20 shows the block diagram of the S-1214 Series to describe the basic operation.

The error amplifier compares the feedback voltage  $(V_{fb})$  whose output voltage  $(V_{OUT})$  is divided by the feedback resistors  $(R_s$  and  $R_f)$  with the reference voltage  $(V_{ref})$ . The error amplifier controls the output transistor, consequently, the regulator starts the operation that keeps  $V_{OUT}$  constant without the influence of the input voltage  $(V_{IN})$ .

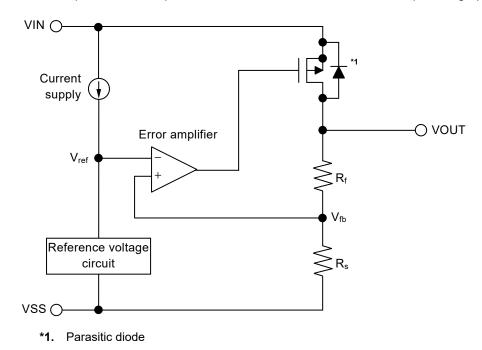


Figure 20

## 2. Output transistor

In the S-1214 Series, a low on-resistance P-channel MOS FET is used between the VIN pin and the VOUT pin as the output transistor. In order to keep  $V_{OUT}$  constant, the on-resistance of the output transistor varies appropriately according to the output current ( $I_{OUT}$ ).

Caution Since a parasitic diode exists between the VIN pin and the VOUT pin due to the structure of the transistor, the IC may be damaged by a reverse current if  $V_{\text{OUT}}$  becomes higher than  $V_{\text{IN}}$ . Therefore, be sure that  $V_{\text{OUT}}$  does not exceed  $V_{\text{IN}} + 0.3 \text{ V}$ .

#### 3. ON / OFF pin

The ON / OFF pin controls the internal circuit and the output transistor in order to start and stop the regulator. When the ON / OFF pin is set to OFF, the internal circuit stops operating and the output transistor between the VIN pin and the VOUT pin is turned off, reducing current consumption significantly.

Note that the current consumption increases when a voltage of 0.8 V to  $V_{\text{IN}}$  – 0.3 V is applied to the ON / OFF pin. The ON / OFF pin is configured as shown in **Figure 21**.

Since the ON / OFF pin is internally pulled down to the VSS pin in the floating status, the VOUT pin is set to the  $V_{SS}$  level.

Table 12

Product Type	ON / OFF Pin	Internal Circuit	VOUT Pin Voltage	Current Consumption
В	"H" : ON	Operate	Constant value*1	Iss <sub>1</sub>
В	"L" : OFF	Stop	Pulled down to Vss*2	Iss2

- \*1. The constant value is output due to the regulating based on the set output voltage value.
- \*2. The VOUT pin voltage is pulled down to  $V_{SS}$  due to combined resistance ( $R_{LOW}$  = 3.0 k $\Omega$  typ.) of the discharge shunt circuit and the feedback resistors, and a load.

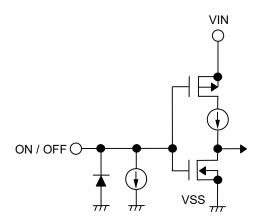
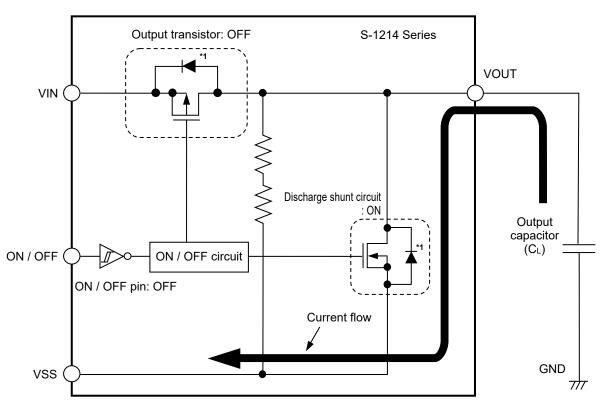


Figure 21

#### 4. Discharge shunt function

The S-1214 Series has a built-in discharge shunt circuit to discharge the output capacitance. The output capacitance is discharged as follows so that the VOUT pin reaches the  $V_{SS}$  level.

- (1) The ON / OFF pin is set to OFF level.
- (2) The output transistor is turned off.
- (3) The discharge shunt circuit is turned on.
- (4) The output capacitor discharges.



\*1. Parasitic diode

Figure 22

# 5. Constant current source pull-down

Note that the IC's current consumption increases as much as current flows into the constant current of 0.3  $\mu$ A typ. when the ON / OFF pin is connected to the VIN pin and the S-1214 Series is operating.

Since the ON / OFF pin is internally pulled down to the VSS pin in the floating status, the VOUT pin is set to the  $V_{SS}$  level.

#### 6. Overcurrent protection circuit

The S-1214 Series has a built-in overcurrent protection circuit to limit the overcurrent of the output transistor. When the VOUT pin is shorted to the VSS pin, that is, at the time of the output short-circuit, the output current is limited to 250 mA typ. due to the overcurrent protection circuit operation. The S-1214 Series restarts regulating when the output transistor is released from the overcurrent status.

In addition, the overcurrent protection circuit is equipped with a detection function of the difference between input and output voltage. When the output current is large and the difference between input voltage and output voltage is large, the output current is limited by the detection function of the difference between input and output voltage. This prevents the output transistor which drives the load from being damaged by excessive electric power. The S-1214 Series restarts regulating when the output transistor is released from the excessive electric power.

For details on output current limit values in input voltage, refer to "10. Input voltage vs. Output current (When load current increases) (Ta = +25°C)" in "

Characteristics (Typical Data)".

#### Caution

This overcurrent protection circuit does not work as for thermal protection. For example, when the output transistor keeps the overcurrent status long at the time of output short-circuit or due to other reasons, pay attention to the conditions of the input voltage and the load current so as not to exceed the power dissipation.

#### 7. Thermal shutdown circuit

The S-1214 Series has a built-in thermal shutdown circuit to limit overheating. When the junction temperature increases to 170°C typ., the thermal shutdown circuit becomes the detection status, and the regulating is stopped. When the junction temperature decreases to 135°C typ., the thermal shutdown circuit becomes the release status, and the regulator is restarted.

If the thermal shutdown circuit becomes the detection status due to self-heating, the regulating is stopped and  $V_{\text{OUT}}$  decreases. For this reason, the self-heating is limited and the temperature of the IC decreases. The thermal shutdown circuit becomes release status when the temperature of the IC decreases, and the regulating is restarted thus the self-heating is generated again. Repeating this procedure makes the waveform of  $V_{\text{OUT}}$  into a pulse-like form. This phenomenon continues unless decreasing either or both of the input voltage and the output current in order to reduce the internal power consumption, or decreasing the ambient temperature. Note that the product may suffer physical damage such as deterioration if the above phenomenon occurs continuously.

#### Caution

If a large load current flows during the restart process of regulating after the thermal shutdown circuit changes to the release status from the detection status, the thermal shutdown circuit becomes the detection status again due to self-heating, and a problem may happen in the restart of regulating. A large load current, for example, occurs when charging to the  $C_L$  whose capacitance is large.

Perform thorough evaluation including the temperature characteristics with an actual application to select  $\mathbf{C}_{L}$ .

Table 13

Thermal Shutdown Circuit	VOUT Pin Voltage
Release: 135°C typ.*1	Constant value*2
Detection: 170°C typ.*1	Pulled down to Vss*3

- \*1. Junction temperature
- \*2. The constant value is output due to the regulating based on the set output voltage value.
- \*3. The VOUT pin voltage is pulled down to  $V_{SS}$  due to the feedback resistors ( $R_s$  and  $R_f$ ) and a load.

## 8. Inrush current limit circuit

The S-1214 Series has a built-in inrush current limit circuit to limit the inrush current generated at power-on or at the time when the ON / OFF pin is set to ON. The inrush current is limited to 400 mA typ. immediately after power-on or for the internally set, predetermined time of 100  $\mu$ s min. from the time when ON / OFF pin is set to ON.

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#### 9. Type in which output voltage is externally set

The S-1214 Series provides the type in which output voltage can be set via the external resistor. The output voltage can be set by connecting a resistor ( $R_a$ ) between the VOUT pin and the VADJ pin, and a resistor ( $R_b$ ) between the VADJ pin and the VSS pin.

The output voltage is determined by the following formulas.

$$\begin{aligned} &V_{OUT} = 1.8 + R_a \times I_a & \cdots \\ &S_b = 1.8 + R_a \times I_b \end{aligned} (1)$$
 By substituting  $I_a = I_{VADJ} + 1.8 / R_b$  to above formula (1), 
$$V_{OUT} = 1.8 + R_a \times (I_{VADJ} + 1.8 / R_b) = 1.8 \times (1.0 + R_a / R_b) + R_a \times I_{VADJ} \cdots (2)$$

In above formula (2),  $R_a \times I_{VADJ}$  is a factor for the output voltage error.

Whether the output voltage error is minute is judged depending on the following (3) formula.

By substituting 
$$I_{VADJ} = 1.8 / R_{VADJ}$$
 to  $R_a \times I_{VADJ}$   
 $V_{OUT} = 1.8 \times (1.0 + R_a / R_b) + 1.8 \times R_a / R_{VADJ}$  .....(3)

If R<sub>VADJ</sub> is sufficiently larger than R<sub>a</sub>, the error is judged as minute.

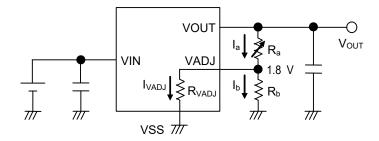


Figure 23

The following expression is in order to determine V<sub>OUT</sub> = 20.0 V. If R<sub>b</sub> = 10 k $\Omega$ , substitute R<sub>VADJ</sub> = 26 M $\Omega$  typ. into (3), R<sub>a</sub> = (20.0 / 1.8 – 1) × ((10 k × 26 M) / (10 k + 26 M))  $\cong$  101 k $\Omega$ 

Caution The above connection diagram and constants will not guarantee successful operation. Perform thorough evaluation using the actual application to set the constants.

#### ■ Precautions

- Generally, when a voltage regulator is used under the condition that the load current value is small (1 mA or less), the output voltage may increase due to the leakage current of an output transistor.
- Generally, when a voltage regulator is used under the condition that the temperature is high, the output voltage may increase due to the leakage current of an output transistor.
- Generally, when the ON / OFF pin is used under the condition of OFF, the output voltage may increase due to the leakage current of an output transistor.
- Generally, when a voltage regulator is used under the condition that the impedance of the power supply is high, an oscillation may occur. Perform thorough evaluation including the temperature characteristics with an actual application to select C<sub>IN</sub>.
- Generally, in a voltage regulator, an oscillation may occur depending on the selection of the external parts. The following use conditions are recommended in the S-1214 Series; however, perform thorough evaluation including the temperature characteristics with an actual application to select C<sub>IN</sub> and C<sub>L</sub>.

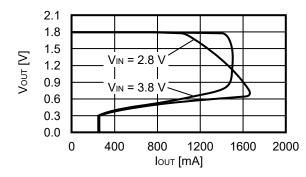
Input capacitor ( $C_{IN}$ ): A ceramic capacitor with capacitance of 1.0  $\mu F$  or more is recommended. Output capacitor ( $C_L$ ): A ceramic capacitor with capacitance of 1.0  $\mu F$  or more is recommended.

- Generally, in a voltage regulator, the values of an overshoot and an undershoot in the output voltage vary depending
  on the variation factors of input voltage start-up, input voltage fluctuation, load fluctuation etc., or the capacitance of C<sub>IN</sub>
  or C<sub>L</sub> and the value of the equivalent series resistance (ESR), which may cause a problem to the stable operation.
  Perform thorough evaluation including the temperature characteristics with an actual application to select C<sub>IN</sub> and C<sub>L</sub>.
- Generally, in a voltage regulator, an overshoot may occur in the output voltage momentarily if the input voltage steeply
  changes when the input voltage is started up, the input voltage fluctuates, etc. Perform thorough evaluation including
  the temperature characteristics with an actual application to confirm no problems happen.
- Generally, in a voltage regulator, if the VOUT pin is steeply shorted with GND, a negative voltage exceeding the absolute maximum ratings may occur in the VOUT pin due to resonance phenomenon of the inductance and the capacitance including C<sub>L</sub> on the application. The resonance phenomenon is expected to be weakened by inserting a series resistor into the resonance path, and the negative voltage is expected to be limited by inserting a protection diode between the VOUT pin and the VSS pin.
- If the input voltage is started up steeply under the condition that the capacitance of C<sub>L</sub> is large, the thermal shutdown circuit may be in the detection status by self-heating due to the charge current to C<sub>L</sub>.
- Make sure of the conditions for the input voltage, output voltage and the load current so that the internal loss does not exceed the power dissipation.
- Do not apply an electrostatic discharge to this IC that exceeds the performance ratings of the built-in electrostatic protection circuit.
- When considering the output current value that the IC is able to output, make sure of the output current value specified in **Table 9** and **Table 10** in  **Electrical Characteristics**" and footnote \*4 of the table.
- Wiring patterns on the application related to the VIN pin, the VOUT pin and the VSS pin should be designed so that the
  impedance is low. When mounting C<sub>IN</sub> between the VIN pin and the VSS pin and C<sub>L</sub> between the VOUT pin and the
  VSS pin, connect the capacitors as close as possible to the respective destination pins of the IC.
- When setting the output voltage by using an external resistor, connect a resistor (R<sub>a</sub>) between the VOUT pin and the VADJ pin and a resistor (R<sub>b</sub>) between the VADJ pin and the VSS pin close to the respective pins.
- In the package equipped with heat sink of backside, mount the heat sink firmly. Since the heat radiation differs according to the condition of the application, perform thorough evaluation with an actual application to confirm no problems happen.
- ABLIC Inc. claims no responsibility for any disputes arising out of or in connection with any infringement by products including this IC of patents owned by a third party.

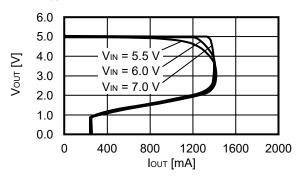
# ■ Characteristics (Typical Data)

1. Output voltage vs. Output current (When load current increases) (Ta = +25°C)

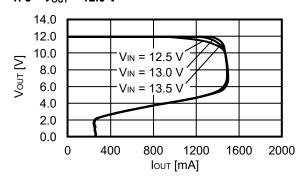
#### 1. 1 Vout = 1.8 V



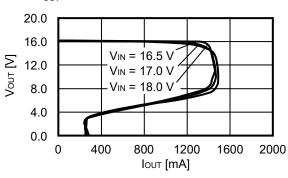
#### 1. 2 V<sub>OUT</sub> = 5.0 V



1. 3 V<sub>OUT</sub> = 12.0 V



1. 4  $V_{OUT} = 16.0 V$ 

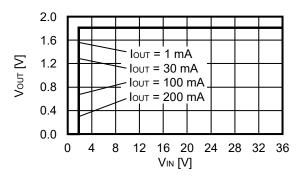


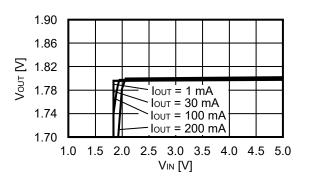
**Remark** In determining the output current, attention should be paid to the following.

- The minimum output current value and footnote \*4 of Table 9 and Table 10 in "■ Electrical Characteristics"
- 2. Power dissipation

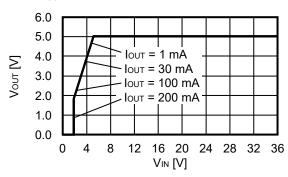
#### 2. Output voltage vs. Input voltage ( $Ta = +25^{\circ}C$ )

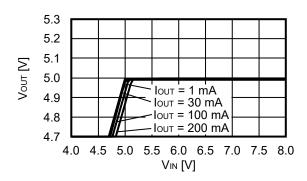
#### 2. 1 Vout = 1.8 V



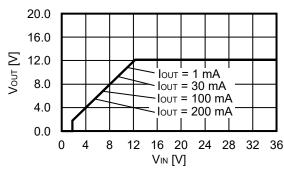


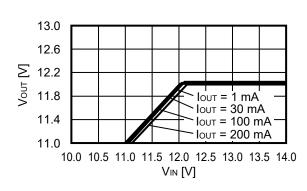
#### 2. 2 Vout = 5.0 V



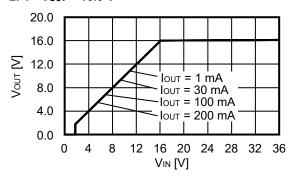


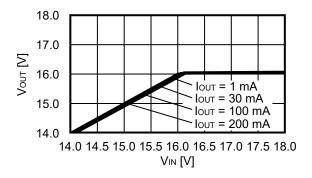
# 2. 3 V<sub>OUT</sub> = 12.0 V





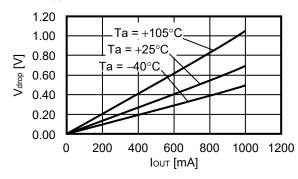
#### 2. 4 V<sub>OUT</sub> = 16.0 V



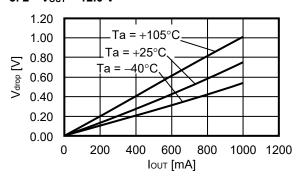


#### 3. Dropout voltage vs. Output current

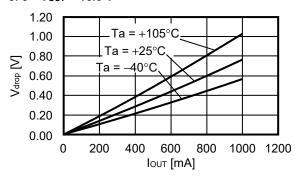
#### 3. 1 Vout = 5.0 V



3. 2 V<sub>OUT</sub> = 12.0 V

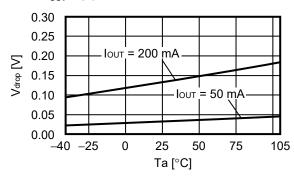


3. 3 Vout = 16.0 V

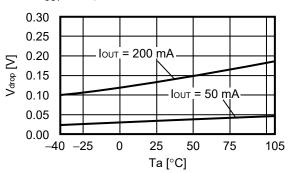


# 4. Dropout voltage vs. Operation ambient temperature

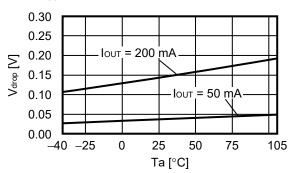
4. 1 Vout = 5.0 V



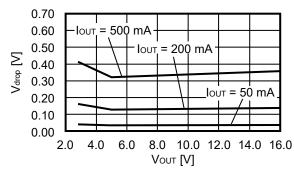
4. 2 Vout = 12.0 V



4. 3 Vout = 16.0 V

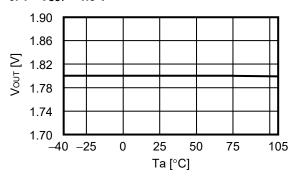


5. Dropout voltage vs. Set output voltage (Ta = +25°C)

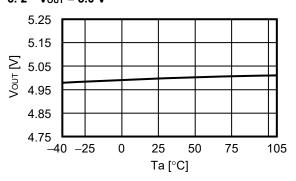


# 6. Output voltage vs. Operation ambient temperature

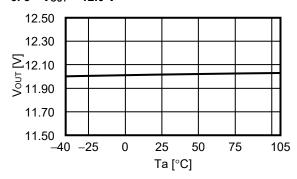
6. 1 Vout = 1.8 V



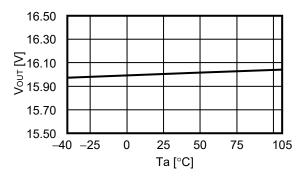
6. 2 V<sub>OUT</sub> = 5.0 V



6. 3 V<sub>OUT</sub> = 12.0 V

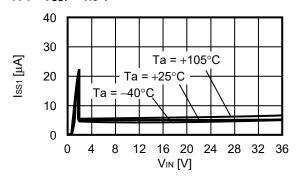


6. 4 V<sub>OUT</sub> = 16.0 V

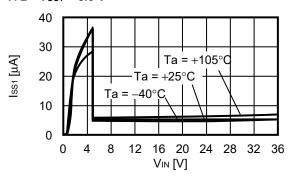


## 7. Current consumption during operation vs. Input voltage (When ON / OFF pin is ON, no load)

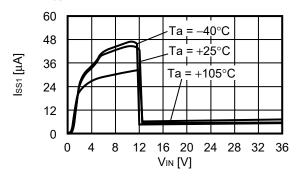
### 7. 1 V<sub>OUT</sub> = 1.8 V



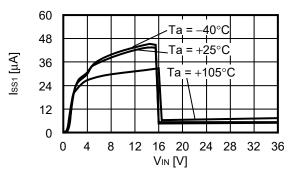
7. 2 V<sub>OUT</sub> = 5.0 V



7. 3  $V_{OUT} = 12.0 V$ 

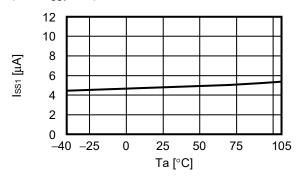


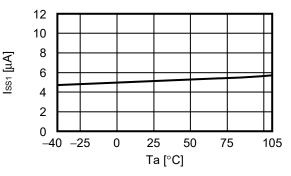
7. 4 V<sub>OUT</sub> = 16.0 V



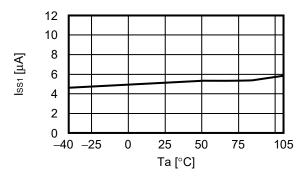
## 8. Current consumption during operation vs. Operation ambient temperature

8. 1 Vout = 1.8 V

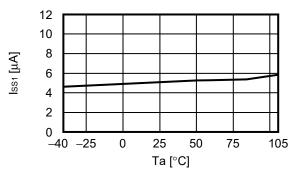




8. 3  $V_{OUT} = 12.0 V$ 

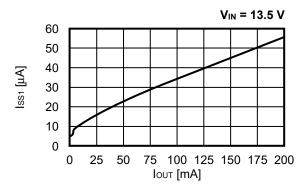


8. 4  $V_{OUT} = 16.0 V$ 

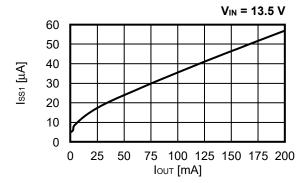


## 9. Current consumption during operation vs. Output current (Ta = +25°C)

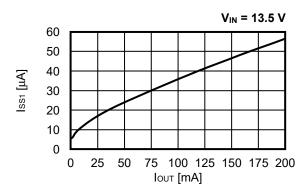
#### 9. 1 Vout = 1.8 V



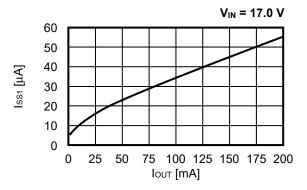
### 9. 2 V<sub>OUT</sub> = 5.0 V



#### 9. 3 V<sub>OUT</sub> = 12.0 V

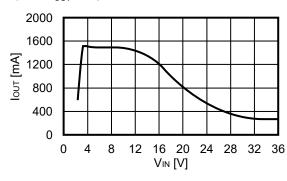


9. 4 V<sub>OUT</sub> = 16.0 V

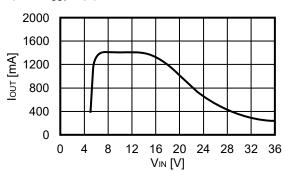


## 10. Input voltage vs. Output current (When load current increases) (Ta = +25°C)

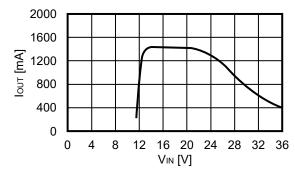
10. 1 Vout = 1.8 V



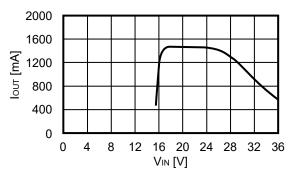
10. 2 V<sub>OUT</sub> = 5.0 V



10. 3 V<sub>OUT</sub> = 12.0 V

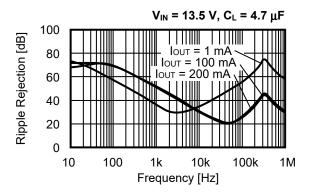


10. 4 V<sub>OUT</sub> = 16.0 V

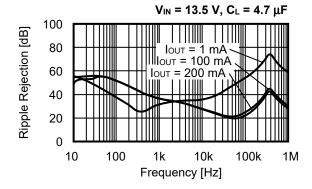


# 11. Ripple rejection (Ta = +25°C)

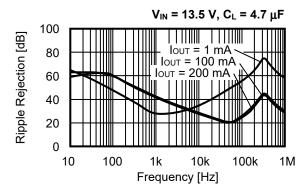
#### 11. 1 Vout = 1.8 V



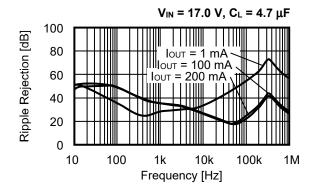
# 11. 3 V<sub>OUT</sub> = 12.0 V



#### 11. 2 Vout = 5.0 V



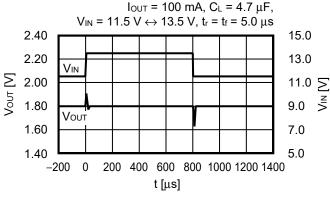
11. 4 V<sub>OUT</sub> = 16.0 V



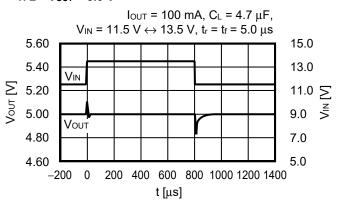
#### ■ Reference Data

#### 1. Characteristics of input transient response ( $Ta = +25^{\circ}C$ )

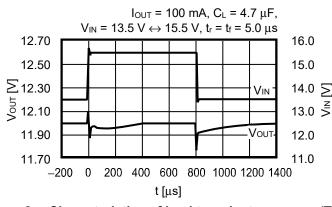
#### 1. 1 Vout = 1.8 V



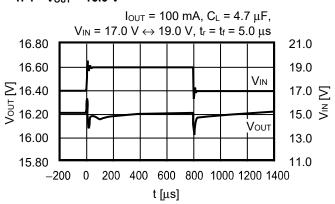
#### 1. 2 Vout = 5.0 V



#### 1. 3 $V_{OUT} = 12.0 V$

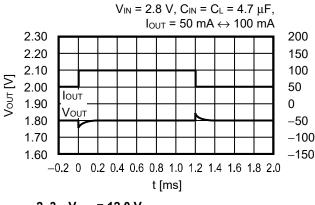


## 1. 4 $V_{OUT} = 16.0 V$

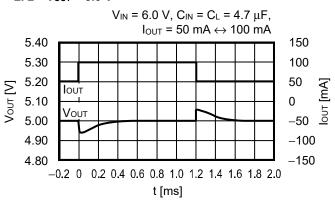


## 2. Characteristics of load transient response (Ta = +25°C)

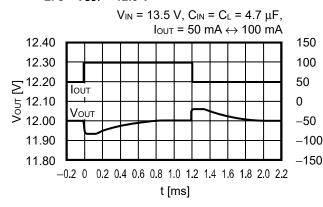
#### 2. 1 Vout = 1.8 V



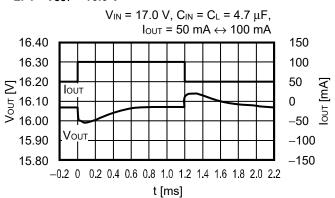
#### 2. 2 Vout = 5.0 V



#### 2. 3 Vout = 12.0 V



#### 2. 4 Vout = 16.0 V

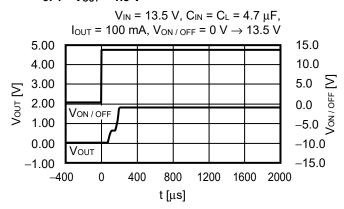


[mA]

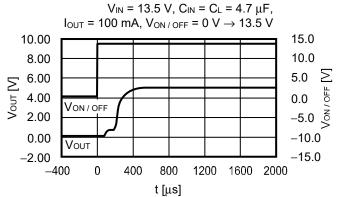
-50

#### 3. Transient response characteristics of ON / OFF pin (Ta = +25°C)

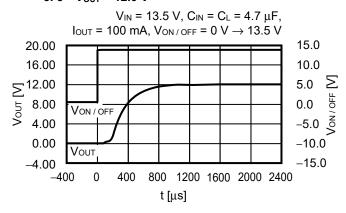
#### 3. 1 Vout = 1.8 V



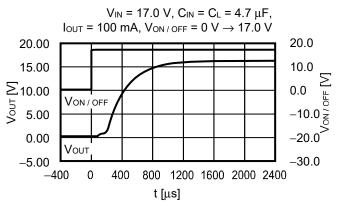
#### 3. 2 Vout = 5.0 V



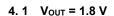
## 3. 3 $V_{OUT} = 12.0 V$

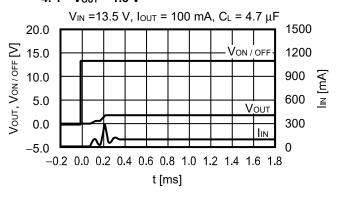


#### 3. 4 V<sub>OUT</sub> = 16.0 V

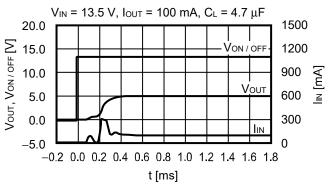


#### 4. Inrush current characteristics (Ta = +25°C)

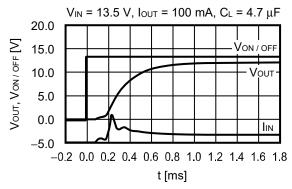




#### 4. 2 V<sub>OUT</sub> = 5.0 V



#### 4. 3 V<sub>OUT</sub> = 12.0 V



4. 4 V<sub>OUT</sub> = 16.0 V

1500

1200

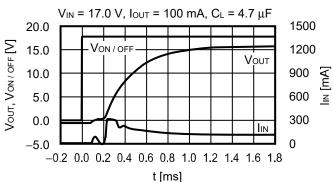
900

600

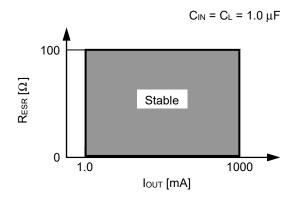
300

n

ll<sub>N</sub> [mA]



# 5. Example of equivalent series resistance vs. Output current characteristics (Ta = +25°C)



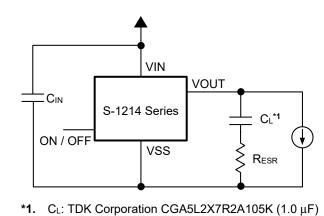
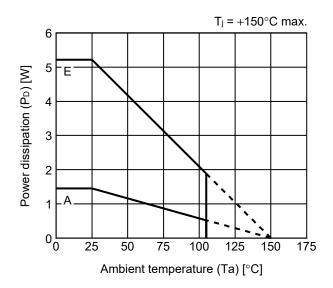


Figure 24

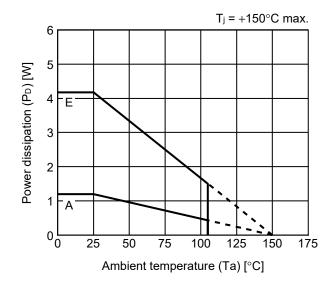
Figure 25

# **■** Power Dissipation

# TO-252-5S(A)



#### **HSOP-8A**



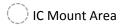
Board	Power Dissipation (P <sub>D</sub> )*1
Α	1.45 W
В	_
С	_
D	_
Е	5.21 W

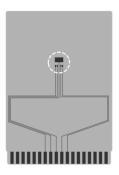
Board	Power Dissipation (P <sub>D</sub> )*1
Α	1.20 W
В	_
С	_
D	_
Е	4.17 W

<sup>\*1.</sup> Measurement values when this IC is mounted on each board

# **TO-252-5S** Test Board

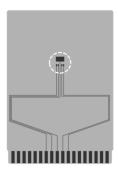
# (1) Board A





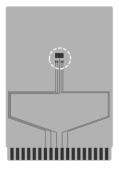
Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		2
	1	Land pattern and wiring for testing: t0.070
Coppor foil layer [mm]	2	-
Copper foil layer [mm]	3	-
	4	74.2 x 74.2 x t0.070
Thermal via		-

# (2) Board B



Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		4
	1	Land pattern and wiring for testing: t0.070
Copper foil layer [mm]	2	74.2 x 74.2 x t0.035
Copper foli layer [min]	3	74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		-

# (3) Board C



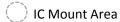
Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		4
	1	Land pattern and wiring for testing: t0.070
Copper foil layer [mm]	2	74.2 x 74.2 x t0.035
Copper foil layer [min]	3	74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		Number: 4 Diameter: 0.3 mm

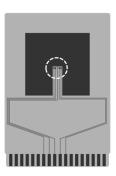


No. TO252-5S-A-Board-SD-1.0

# TO-252-5S Test Board

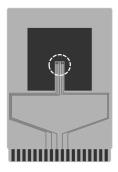
# (4) Board D





Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil Is	ayer	4
	1	Pattern for heat radiation: 2000mm <sup>2</sup> t0.070
Copper foil layer [mm]	2	74.2 x 74.2 x t0.035
Copper foil layer [min]	3	74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		-

# (5) Board E



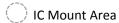
Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil I	ayer	4
	1	Pattern for heat radiation: 2000mm <sup>2</sup> t0.070
Copper foil layer [mm]	2	74.2 x 74.2 x t0.035
Copper foil layer [min]	3	74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		Number: 4 Diameter: 0.3 mm



No. TO252-5S-A-Board-SD-1.0

# **HSOP-8A** Test Board

# (1) Board A





Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		2
	1	Land pattern and wiring for testing: t0.070
Copper foil layer [mm]	2	-
Copper foil layer [min]	3	-
	4	74.2 x 74.2 x t0.070
Thermal via		-

# (2) Board B



Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		4
	1	Land pattern and wiring for testing: t0.070
Connor foil lover [mm]	2	74.2 x 74.2 x t0.035
Copper foil layer [mm]	3	74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		-

# (3) Board C



Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil I	ayer	4
	1	Land pattern and wiring for testing: t0.070
Copper foil layer [mm]	2	74.2 x 74.2 x t0.035
Copper foil layer [min]	3	74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		Number: 4 Diameter: 0.3 mm



enlarged view

No. HSOP8A-A-Board-SD-1.0

# **HSOP-8A** Test Board

# (4) Board D





Item		Specification	
Size [mm]		114.3 x 76.2 x t1.6	
Material		FR-4	
Number of copper foil layer		4	
Copper foil layer [mm]	1	Pattern for heat radiation: 2000mm2 t0.070	
	2	74.2 x 74.2 x t0.035	
	3	74.2 x 74.2 x t0.035	
	4	74.2 x 74.2 x t0.070	
Thermal via		-	

# (5) Board E

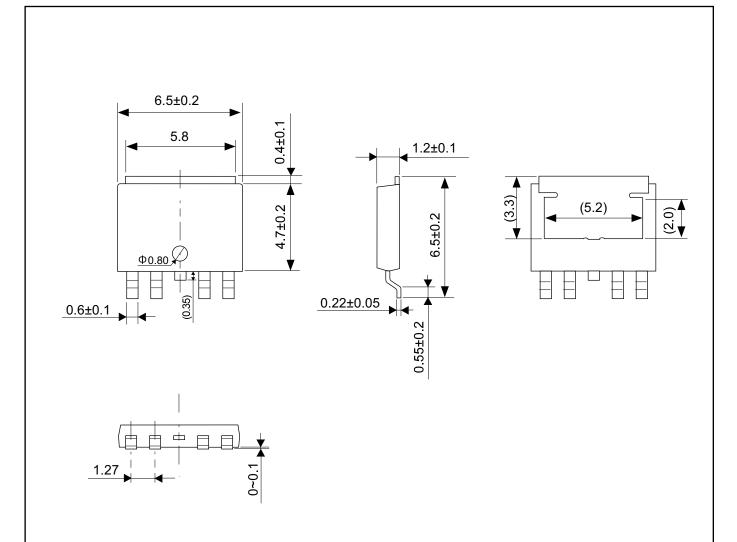


Item		Specification	
Size [mm]		114.3 x 76.2 x t1.6	
Material		FR-4	
Number of copper foil layer		4	
Copper foil layer [mm]	1	Pattern for heat radiation: 2000mm <sup>2</sup> t0.070	
	2	74.2 x 74.2 x t0.035	
	3	74.2 x 74.2 x t0.035	
	4	74.2 x 74.2 x t0.070	
Thermal via		Number: 4 Diameter: 0.3 mm	



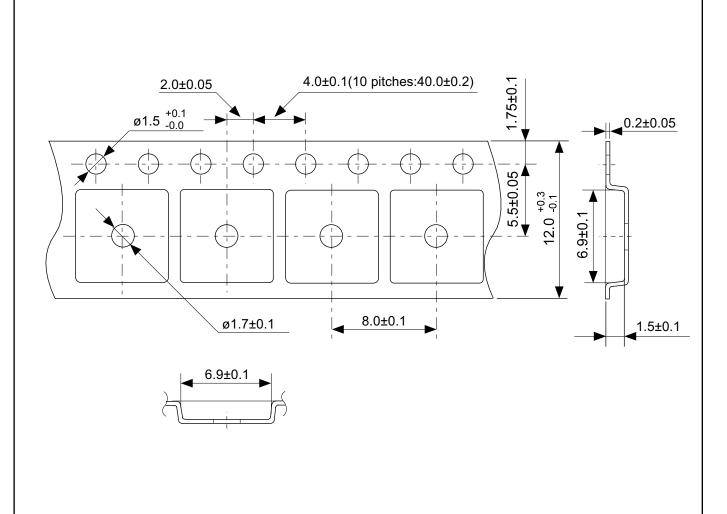
enlarged view

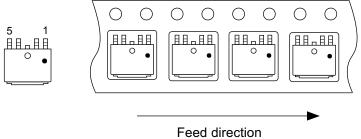
No. HSOP8A-A-Board-SD-1.0



# No. VA005-A-P-SD-2.0

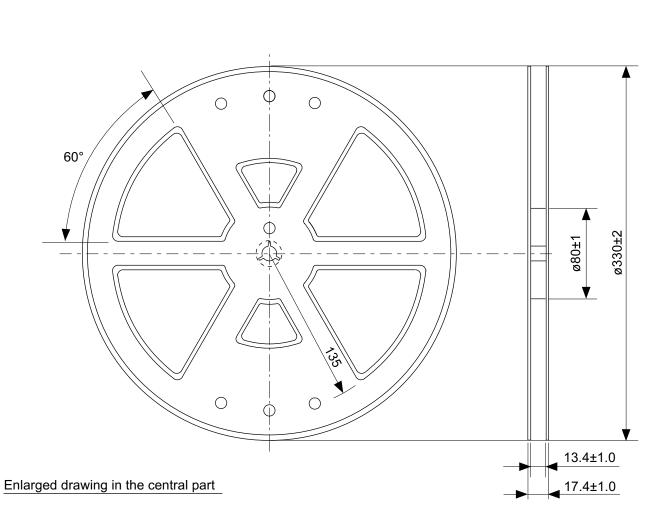
TITLE	TO-252-5S-A-PKG Dimensions
No.	VA005-A-P-SD-2.0
ANGLE	$\oplus$
UNIT	mm
ABLIC Inc.	

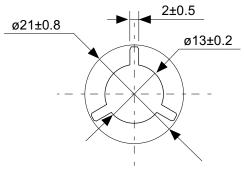




No. VA005-A-C-SD-1.0

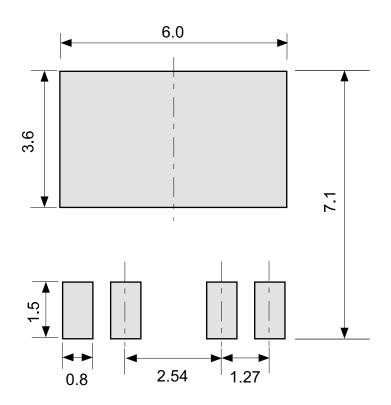
TITLE	TO-252-5S-A-Carrier Tape
No.	VA005-A-C-SD-1.0
ANGLE	
UNIT	mm
ABLIC Inc.	





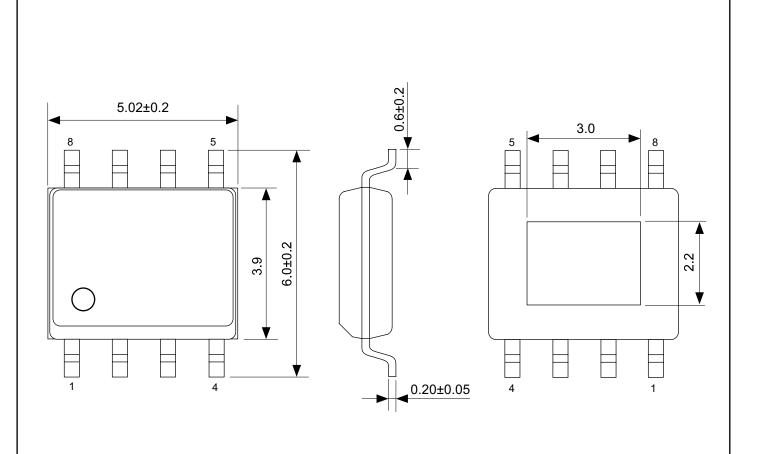
No. VA005-A-R-SD-1.0

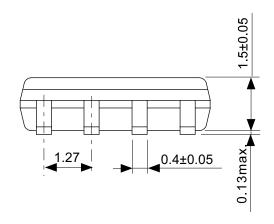
TITLE	TO-252	-5S-A-R	امما
1111	10-232	-30-7-11	661
No.	VA005	VA005-A-R-SD-1.0	
ANGLE		QTY.	4,000
UNIT	mm		
ABLIC Inc.			



# No. VA005-A-L-SD-1.0

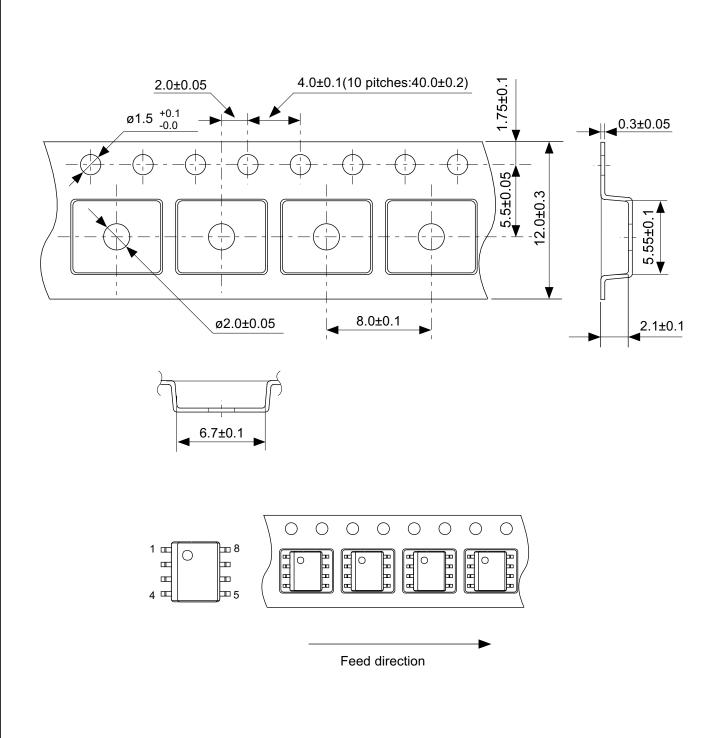
TITLE	TO-252-5S-A -Land Recommendation
No.	VA005-A-L-SD-1.0
ANGLE	
UNIT	mm
ABLIC Inc.	





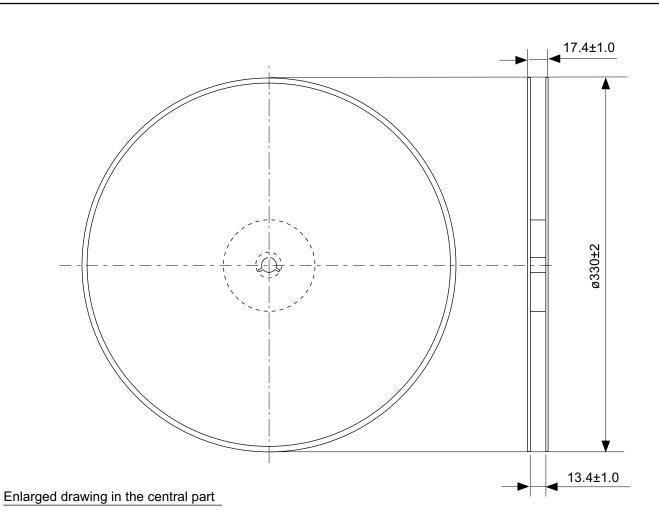
# No. FH008-A-P-SD-2.0

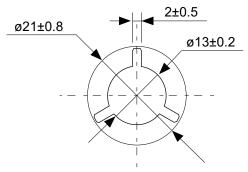
TITLE	HSOP8A-A-PKG Dimensions
No.	FH008-A-P-SD-2.0
ANGLE	<b>\$</b> =1
UNIT	mm
ABLIC Inc.	



# No. FH008-A-C-SD-1.0

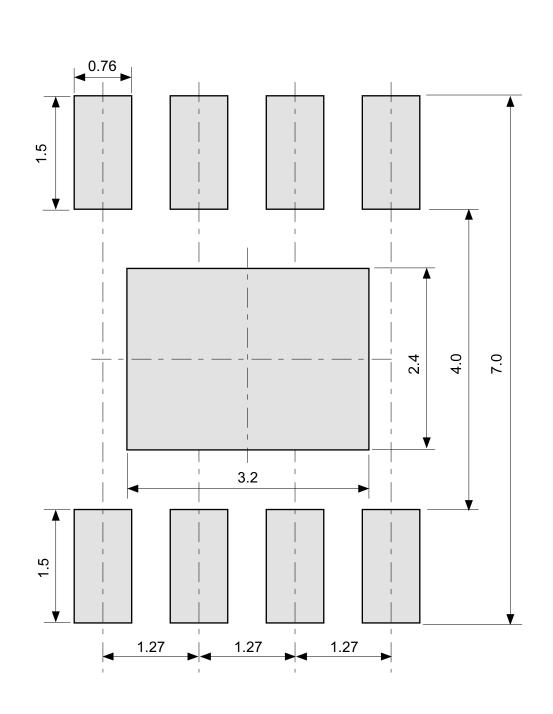
TITLE	HSOP8A-A-Carrier Tape	
No.	FH008-A-C-SD-1.0	
ANGLE		
UNIT	mm	
ABLIC Inc.		





# No. FH008-A-R-SD-1.0

TITLE	HSOP	8A-A-Re	el
No.	FH008	-A-R-SD-1	.0
ANGLE		QTY.	4,000
UNIT	mm		
ABLIC Inc.			



# No. FH008-A-L-SD-1.0

TITLE	HSOP8A-A -Land Recommendation	
No.	FH008-A-L-SD-1.0	
ANGLE		
UNIT	mm	
ABLIC Inc.		

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