

# 6MHz Rail-to-Rail I/O CMOS Operational Amplifier

### **Features**

Single-Supply Operation from +2.1V ~ +5.5V

Rail-to-Rail Input / Output

• Gain-Bandwidth Product: 6MHz (Typ)

Low Input Bias Current: 1pA (Typ)

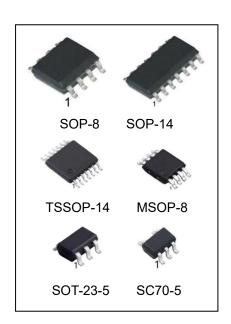
Low Offset Voltage: 3.5mV (Max)

• Quiescent Current: 470µA per Amplifier (Typ)

● Operating Temperature: -40°C ~ +125°C

Small Package:

HGV8631 Available in SOT23-5 and SC70-5,SOP-8 Packages
HGV8632 Available in SOP-8,MSOP-8 Packages
HGV8634 Available in SOP-14 and TSSOP-14 Packages



# **Ordering Information**

DEVICE	Package Type	MARKING	Packing	Packing Qty
HGV8631M5/TR	SOT-23-5	V8631,8631	REEL	3000pcs/reel
HGV8631M7/TR	SC70-5(SOT-353)	8631	REEL	3000pcs/reel
HGV8631M/TR	SOP-8	V8631	REEL	2500pcs/reel
HGV8632M/TR	SOP-8	V8632	REEL	2500pcs/reel
HGV8632MM/TR	MSOP-8	V8632	REEL	3000pcs/reel
HGV8634M/TR	SOP-14	HGV8634	REEL	2500pcs/reel
HGV8634MT/TR	TSSOP-14	V8634	REEL	2500pcs/reel



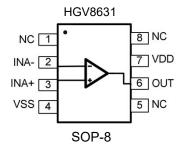
### **General Description**

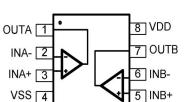
The HGV863X have a high gain-bandwidth product of 6MHz, a slew rate of  $4.2V/\mu s$ , and a quiescent current of  $470\mu A$  per amplifier at 5V. The HGV863X are designed to provide optimal performance in low voltage and low noise systems. They provide rail-to-rail output swing into heavy loads. The input common mode voltage range includes ground, and the maximum input offset voltage is 3.5mV for HGV863X. They are specified over the extended industrial temperature range ( $-40^{\circ}C$  to  $+125^{\circ}C$ ). The operating range is from 2.1V to 5.5V. The HGV8631 single is available in Green SC70-5, SOT23-5 and SOP-8 packages. The HGV8632 dual is available in Green SOP-8 and MSOP-8 packages. The HGV8634 Quad is available in Green SOP-14 and TSSOP-14 packages

## **Applications**

- Sensors
- Active Filters
- Cellular and Cordless Phones
- Laptops and PDAs
- Audio
- Handheld Test Equipment
- Battery-Powered Instrumentation
- A/D Converters

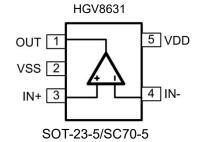
# **Pin Configuration**

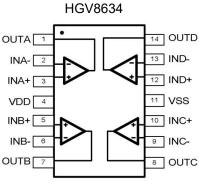




HGV8632

SOP-8/MSOP-8





SOP-14/TSSOP-14



# **Absolute Maximum Ratings**

CONDITIONS	MIN	MAX					
Power Supply Voltage (VDD to Vss)	-0.5V	+7.5V					
Analog Input Voltage (IN+ or IN-)	Vss-0.5V	VDD+0.5V					
PDB Input Voltage	Vss-0.5V	+7V					
Operating Temperature Range	-40°C	+125°C					
Junction Temperature							
Storage Temperature Range	-55°C	+150°C					
Lead Temperature (soldering, 10sec) +260°C							
Package Thermal Resistance (TA=+25℃)							
SOP-8, θ <sub>JA</sub>	125°	C/W					
MSOP-8, θ <sub>JA</sub>	216°	C/W					
SOT23-5, θ <sub>JA</sub>	190°	C/W					
SC70-5, θ <sub>JA</sub>	333°	C/W					
ESD Susceptibility							
HBM 8KV							
MM	400	)V					

Note: Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability



# **Electrical Characteristics**

(At Vs=5V,  $T_A$  = +25°C,  $V_{CM}$  =  $V_S/2$ ,  $R_L$  = 600 $\Omega$ , unless otherwise noted.)

					HGV863	31/2/4		
DADAMETED	CONDITIONS	TYP		MIN/I	MAX OVI	ER TEMPI	ERATURE	
PARAMETER	CONDITIONS	<b>+25</b> ℃	<b>+25</b> ℃	0℃ to 70℃	-40℃ to 85℃	-40 ℃ to 125℃	UNITS	MIN /MAX
NPUT CHARACTERISTICS								
Input Offset Voltage (Vos)		8.0	3.5	3.9	4.3	4.6	mV	MAX
Input Bias Current (I <sub>B</sub> )		1					pА	TYP
Input Offset Current (Ios)		1					pА	TYP
Input Common Mode Voltage Range (V <sub>CM</sub> )	Vs = 5.5V	-0.1 to +5.6					V	TYP
Common Mode Rejection Ratio (CMRR)	$V_S = 5.5V$ , $V_{CM} = -0.1V$ to 4V	90	73	70	70	65	dB	MIN
	$V_S = 5.5V$ , VCM = -0.1V to 5.6V	83					dB	MIN
Open-Loop Voltage Gain (A <sub>OL</sub> )	$R_L = 600\Omega, VO = 0.15V \text{ to } 4.85V$	97	90	87	86	79	dB	MIN
	$R_L = 10k\Omega, VO = 0.05V \text{ to } 4.95V$	108					dB	MIN
Input Offset Voltage Drift ( $\Delta V_{OS}/\Delta_T$ )		2.4					μV/°C	TYP
OUTPUT CHARACTERISTICS								
Output Voltage Swing from Rail	$R_L = 600\Omega$	0.1					V	TYP
	$R_L = 10k\Omega$	0.015					V	TYP
Output Current (I <sub>OUT</sub> )		53	49	45	40	35	mA	MIN
Closed-Loop Output Impedance	f = 200kHz, G = 1	3					Ω	TYP
POWER-DOWN DISABLE								
Turn-On Time		4					μs	TYP
Turn-Off Time		1.2					μs	TYP
POWER SUPPLY								
Operating Voltage Range			2.1	2.1	2.1	2.1	V	MIN
Power Supply Rejection Ratio (PSRR)	$V_S = +2.5V \text{ to } +5.5V$ $V_{CM} = (-VS) + 0.5V$	91	5.5 74	5.5 72	5.5 72	5.5 68	V dB	MAX MIN
Quiescent Current/Amplifier (IQ)	$I_{OUT} = 0$	470	650	750	750	815	μA	MAX



### **Electrical Characteristics**

(At Vs=5V, TA = +25°C, VCM = VS/2, RL =  $600\Omega$ , unless otherwise noted.)

			HGV8631/2/4								
PARAMETER	CONDITIONS	TYP	MIN/MAX OVER TEMPERATURE								
PARAMETER	CONDITIONS	+25℃		0℃to 70℃	-40℃to 85℃	-40℃to 125℃	UNITS	MIN / MAX			
DYNAMIC PERFORMANCE											
Gain-Bandwidth Product (GBP)	$R_L = 10k\Omega$ , $C_L = 100pF$	6					MHz	TYP			
Phase Margin (φ <sub>0</sub> )	$R_L = 10k\Omega, C_L = 100pF$	53					Degrees	TYP			
Full Power Bandwidth (BWP)	$<$ 1% distortion, R <sub>L</sub> = 600 $\Omega$	250					kHz	TYP			
Slew Rate (SR)	G = +1, 2V Step, R <sub>L</sub> = 10kΩ	4.2					V/µs	TYP			
Settling Time to 0.1% (ts)	G = +1, 2V Step, R <sub>L</sub> = 600Ω	0.4					μs	TYP			
Overload Recovery Time	V <sub>IN</sub> ·Gain = VS, R <sub>L</sub> = 600Ω	2.5					μs	TYP			
NOISE PERFORMANCE											
Voltage Noise Density (en)	f = 1kHz	13					nV /√Hz	TYP			
	f = 10kHz	9.5					nV /√Hz	TYP			



# **Typical Performance characteristics**

(At Vs=5V,  $T_A = +25^{\circ}\text{C}$ ,  $V_{CM} = V_S/2$ ,  $R_L = 600\Omega$ , unless otherwise noted.) Large-Signal Step Response Large-Signal Step Response Voltage (500mV/div) Vs=5V Voltage (1V/div) Vs=2.5V G=+1 G=+1 CL=100pF CL=100pF RL=10KΩ RL=10 KΩ Time (1µs/div) Time (1µs/div) Small-Signal Step Response Small-Signal Step Response Voltage (50mV/div) Voltage (50mV/div) Vs=5V Vs=2.5V G=+1 CL=100pF G=+1 CL=100pF RL=10 KΩ RL=10 KΩ Time (1µs/div) Time (1µs/div) Negative Overload Recov Positive Overload Recovery 50mV 50mV VIN VIN /div /div  $Vs = \pm 2.5V$ VIN=-50mVp-p RL=600Ω G=-100 RL=600Ω 1V Vout G=-100  $Vs = \pm 2.5V$ 500mV /div VIN=50mVp-p /div

http://www.hgsemi.com.cn

Time (2µs/div)

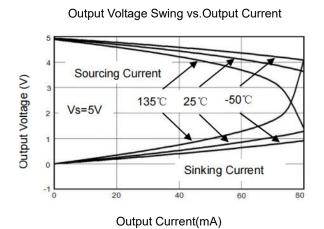
Vout

Time (2µs/div)



# **Typical Performance characteristics**

(At Vs=5V,  $T_A$  = +25°C,  $V_{CM}$  =  $V_S/2$ ,  $R_L$  = 600 $\Omega$ , unless otherwise noted.)



550 Supply Current (µA) 500 450 Vs=2.5 400 Vs=3

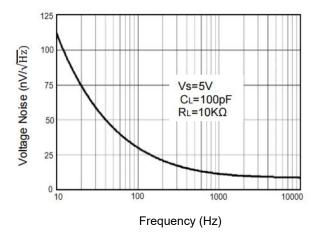
Supply Current vs. Temperature

25 Temperature (°C)

350

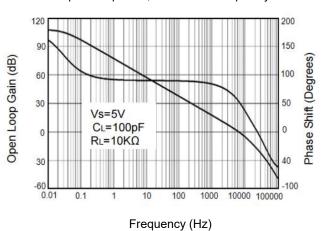
300

Input Voltage Noise Spectral Density vs. Frequency

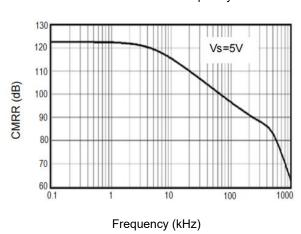


Open Loop Gain, Phase shift Frequency

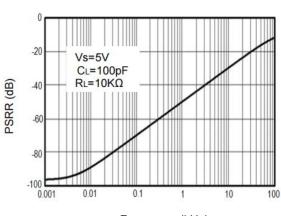
100



CMRR vs. Frequency



PSRR vs. Frequency



Frequency (kHz)



### **Application Note**

#### Size

HGV863X series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The small footprints of the HGV863X series packages save space on printed circuit boards and enable the design of smaller electronic products.

### Power Supply Bypassing and Board Layout

HGV863X series operates from a single 2.1V to 5.5V supply or dual  $\pm 1.05$ V to  $\pm 2.75$ V supplies. For best performance, a  $0.1\mu$ F ceramic capacitor should be placed close to the VDD pin in single supply operation. For dual supply operation, both VDD and VSS supplies should be bypassed to ground with separate  $0.1\mu$ F ceramic capacitors.

### Low Supply Current

The low supply current (typical 470µA per channel) of HGV863X series will help to maximize battery life. They are ideal for battery powered systems.

### **Operating Voltage**

HGV863X series operate under wide input supply voltage (2.1V to 5.5V). In addition, all temperature specifications apply from -40°C to +125°C. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-lon battery lifetime.

### Rail-to-Rail Input

The input common-mode range of HGV863X series extends 100mV beyond the supply rails ( $V_{SS}$ -0.1V to  $V_{DD}$ +0.1V). This is achieved by using complementary input stage. For normal operation, inputs should be limited to this range.

#### Rail-to-Rail Output

Rail-to-Rail output swing provides maximum possible dynamic range at the output. This is particularly important when operating in low supply voltages. The output voltage of HGV863X series can typically swing to less than 2mV from supply rail in light resistive loads (>100k $\Omega$ ), and 60mV of supply rail in moderate resistive loads (10k $\Omega$ ).

### **Capacitive Load Tolerance**

The HGV863x family is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain. Figure 2. shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.



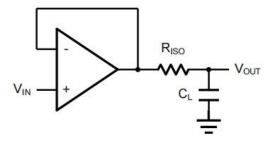


Figure 2. Indirectly Driving a Capacitive Load Using Isolation Resistor

The bigger the  $R_{\rm ISO}$  resistor value, the more stable  $V_{\rm OUT}$  will be. However, if there is a resistive load  $R_{\rm L}$  in parallel with the capacitive load, a voltage divider (proportional to  $R_{\rm ISO}/R_{\rm L}$ ) is formed, this will result in a gain error.

The circuit in Figure 3 is an improvement to the one in Figure 2.  $R_F$  provides the DC accuracy by feed-forward the  $V_{IN}$  to  $R_L$ .  $C_F$  and  $R_{ISO}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of  $C_F$ . This in turn will slow down the pulse response.

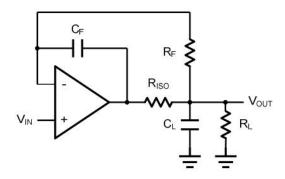


Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy



### **Typical Application Circuits**

### Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4. shown the differential amplifier using HGV863X.

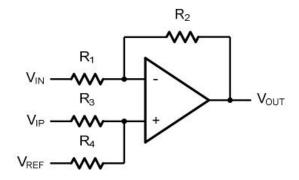


Figure 4. Differential Amplifier

$$V_{OUT} \text{--} (\frac{R_1 + R_2}{R_3 + R_4}) \frac{R_4}{R_1} V_{IN} \text{--} \frac{R2}{R1} V_{IP} \text{--} (\frac{R_1 + R_2}{R_3 + R_4}) \frac{R_3}{R_1} V_{REF}$$

If the resistor ratios are equal  $(i.e.R_1 = R_3 \text{ and } R_{2=}R_4)$ , then

$$V_{OUT} = \frac{R2}{R1} (V_{IP} - V_{IN}) + V_{REF}$$

### Low Pass Active Filter

The low pass active filter is shown in Figure 5. The DC gain is defined by -R2/R1. The filter has a -20dB/decade roll-off after its corner frequency  $fC=1/(2\pi R3C1)$ 

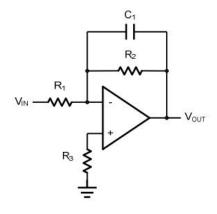


Figure 5. Low Pass Active Filter



### **Instrumentation Amplifier**

The triple HGV863X can be used to build a three-op-amp instrumentation amplifier as shown in Figure 6. The amplifier in Figure 6 is a high input impedance differential amplifier with gain of  $R_2/R_1$ . The two differential voltage followers assure the high input impedance of the amplifier.

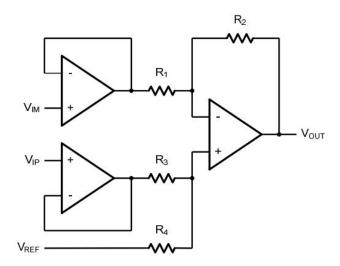
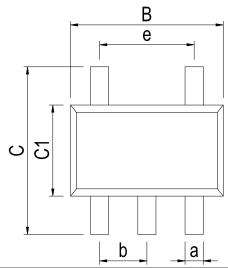


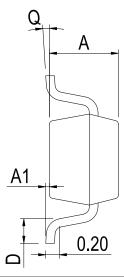
Figure 6. Instrument Amplifier



# **Physical Dimensions**

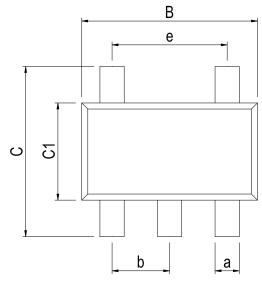
### SOT-23-5

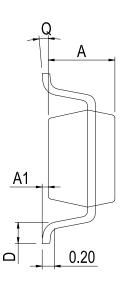




Dimensions In Millimeters(SOT-23-5)										
Symbol:	Α	A1	В	С	C1	D	Q	а	b	е
Min:	1.00	0.00	2.82	2.65	1.50	0.30	0°	0.30	0.95 BSC	1.90 BSC
Max:	1.15	0.15	3.02	2.95	1.70	0.60	8°	0.50	0.95 650	

# SC70-5



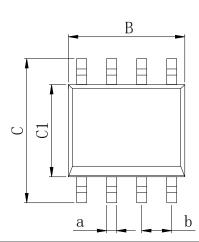


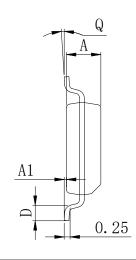
Dimensions In Millimeters(SC70-5)										
Symbol:	А	A1	В	С	C1	D	Q	а	b	е
Min:	0.90	0.00	2.00	2.15	1.15	0.26	0°	0.15	0.65	4 20 DCC
Max:	1.00	0.15	2.20	2.45	1.35	0.46	8°	0.35	BSC	1.30 BSC



# **Physical Dimensions**

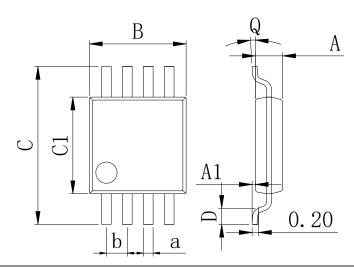
SOP-8 (150mil)





Dimensions In Millimeters(SOP-8)										
Symbol:	Α	A1	В	С	C1	D	Q	а	b	
Min:	1.35	0.05	4.90	5.80	3.80	0.40	0°	0.35	1 27 DCC	
Max:	1.55	0.20	5.10	6.20	4.00	0.80	8°	0.45	1.27 BSC	

## MSOP-8

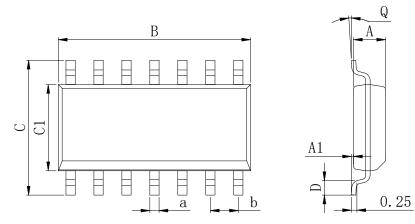


Dimensions In Millimeters(MSOP-8)										
Symbol:	Α	A1	В	С	C1	D	Q	а	b	
Min:	0.80	0.05	2.90	4.75	2.90	0.35	0°	0.25	0.65.000	
Max:	0.90	0.20	3.10	5.05	3.10	0.75	8°	0.35	0.65 BSC	



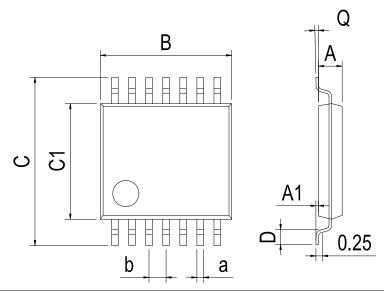
# **Physical Dimensions**

### SOP-14



Dimensions In Millimeters(SOP-14)										
Symbol:	Α	A1	В	С	C1	D	Q	а	b	
Min:	1.35	0.05	8.55	5.80	3.80	0.40	0°	0.35	4.07.BSC	
Max:	1.55	0.20	8.75	6.20	4.00	0.80	8°	0.45	1.27 BSC	

TSSOP-14



Dimensions In Millimeters(TSSOP-14)										
Symbol:	Α	A1	В	С	C1	D	Q	а	b	
Min:	0.85	0.05	4.90	6.20	4.30	0.40	0°	0.20	0.65 BSC	
Max:	0.95	0.20	5.10	6.60	4.50	0.80	8°	0.25		



# **Revision History**

DATE	REVISION	PAGE
2018-6-5	New	1-16
2023-10-31	Update encapsulation type、Update SC70-5 Physical Dimensions	1、12
2024-10-23	Update SOT-23-5 Physical dimension、Update Lead Temperature	12、3



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