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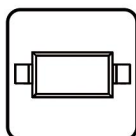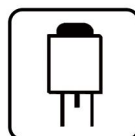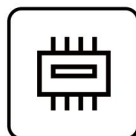
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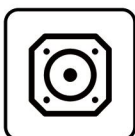
電源管理　顯示驅動　二三極管　LDO穩壓器　觸摸芯片

MOS管　運算放大器　存儲芯片　MCU　串口通信

PCA82C251T-YM-TD

產品規格說明書

**PCA82C251T-YM-TD**
CAN transceiver for 24 V systems

1. DESCRIPTION

The PCA82C251 and PCA82C251 are CAN controller interface chips. As a CAN transceiver, The devices provide transmit and receive capability between the differential CAN bus and a CAN controller, with signaling rates up to 1 Mbps.

2. FEATURES

- Fully compatible with the "ISO 11898-24 V" standard

- Slope control to reduce Radio Frequency Interference (RFI)

- Short-circuit proof to battery and ground in 24 V powered systems

- Low-current Standby mode

- An unpowered node does not disturb the bus lines

- CAN bus communication speed up to 1 Mbps

- High immunity against electromagnetic interference

- Thermally protected

- Package option: PCA82C251 (SOP8), PCA82C251 (DIP8)

3. TYPICAL APPLICATION
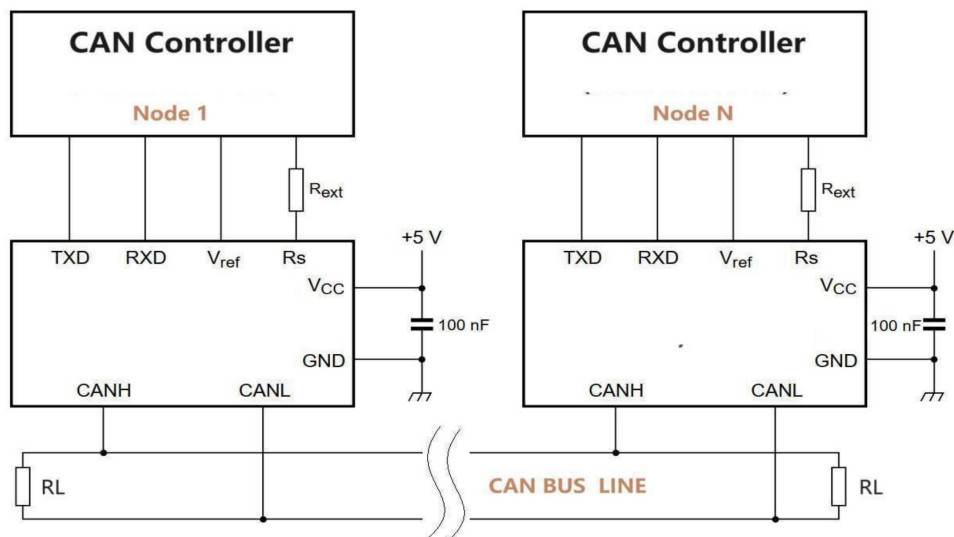
4. PIN CONFIGURATIONS AND FUNCTIONS

PCA82C251 PCA82C251

TXD	1	8	Rs
GND	2	7	CANH
VCC	3	6	CANL
RXD	4	5	Vref

(TOP VIEW)
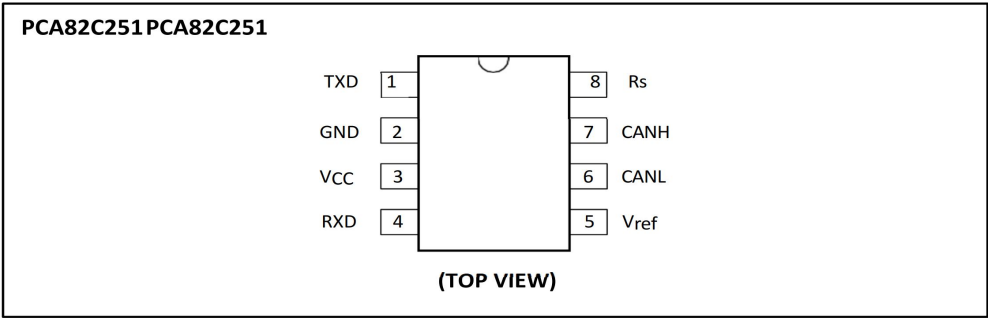
Pin Functions

Symbol	Pin	Description
TXD	1	transmit data input
GND	2	ground
VCC	3	supply voltage
RXD	4	receive data output
Vref	5	reference voltage output
CANL	6	LOW-level CAN voltage input/output
CANH	7	HIGH-level CAN voltage input/output
Rs	8	slope resistor input

4.1. Functional description

The PCA82C251 and PCA82C251 are the interface between a CAN protocol controller and the physical bus. It is primarily intended for applications up to 1 MBd in trucks and buses. The device provides differential transmit capability to the bus and differential receive capability to the CAN controller. It is fully compatible with the "ISO 11898-24 V" standard.

A current-limiting circuit protects the transmitter output stage against short-circuits to positive and negative battery voltage. Although power dissipation will increase as a result of a short circuit fault condition, this feature will prevent destruction of the transmitter output stage.

If the junction temperature exceeds approximately 150 $^\circ$C, the limiting current of both transmitter outputs is decreased. Because the transmitter is responsible for most of the power dissipated, this will result in reduced power dissipation and hence a lower chip temperature. All other parts of the IC will remain operational. The thermal protection is needed, in particular, when a bus line is short-circuited.

The CANH and CANL lines are also protected against electrical transients which may occur in an automotive environment.

Pin 8 (Rs) allows three different modes of operation to be selected: High-speed, Slope control and Standby.

For high-speed operation, the transmitter output transistors are simply switched on and off as fast as possible. In this mode, no measures are taken to limit the rise and fall slopes. A shielded cable is recommended to avoid RFI problems. High-speed mode is selected by connecting pin 8 to ground.

Slope control mode allows the use of an unshielded twisted pair or a parallel pair of wires as bus lines. To reduce RFI, the rise and fall slopes should be limited. The rise and fall slopes can be programmed with a resistor connected from pin 8 to ground. The slope is proportional to the current output at pin 8.

If a HIGH level is applied to pin 8, the circuit enters a low-current Standby mode. In this mode, the transmitter is switched off and the receiver is switched to a low current. If dominant bits are detected (differential bus voltage >0.9 V), RXD will be switched to a LOW level. The microcontroller should react to this condition by switching the transceiver back to normal operation (via pin 8). Because the receiver is slower in Standby mode, the first message will be lost at higher bit rates.

Table 4-1. Truth table of the CAN transceiver

Supply	TXD	CANH	CANL	Bus state	RXD
4.5 V to 5.5 V	0	HIGH	LOW	dominant	0
4.5 V to 5.5 V	1 (or floating)	floating	floating	recessive	1
<2V (not powered)	X[1]	floating	floating	recessive	X[1]
2 V < Vcc < 4.5 V	>0.75 V_{CC}	floating	floating	recessive	X[1]
2 V < Vcc < 4.5 V	X[1]	floating if $V_{Rs} > 0.75V_{CC}$	Floating If $V_{Rs} > 0.75V_{CC}$	recessive	X[1]

[1]   X = don't care.

Table 4-2. Pin Rs summary

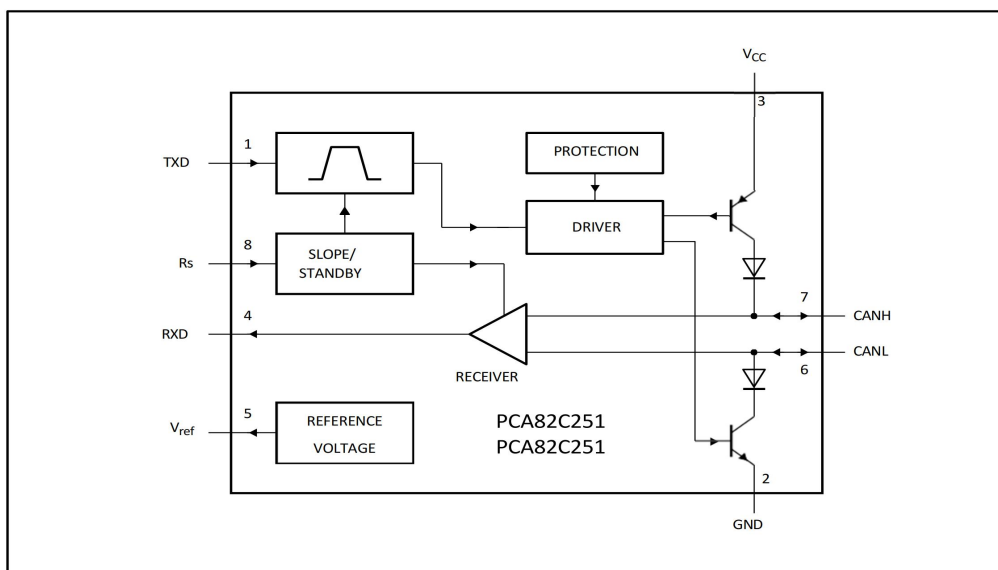| Condition forced at pin Rs	Mode	Resulting voltage or current at pin Rs
$V_{Rs} > 0.75V_{CC}$	Standby	$I_{Rs} < \lvert 10 \mu A \rvert$
-10 µA< I_{Rs} < -200 µA	Slope control	$0.4V_{CC} < V_{Rs} < 0.6V_{CC}$
$V_{Rs} < 0.3V_{CC}$	High-speed	$I_{Rs} < -500 \mu A$

5. BLOCK DIAGRAM

Fig 5-1. Block Diagram

6.   SPECIFICATIONS

6.1 Absolute Maximum Ratings

In accordance with the Absolute Maximum Rating System (IEC 60134). All voltages are referenced to pin 2（GND）; positive input current.

Symbol	Parameter	Conditions	Min	Max	Unit
VCC	supply voltage		- 0.3	+7.0	V
Vn	DC voltage at pins 1, 4, 5 and 8		- 0.3	V_{CC} + 0.3	V
V6,7	DC voltage at pin 6 and 7	0 V < V_{CC} < 5.5 V; TXD HIGH or floating	-32	+32	V
Vtrt	transient voltage at pins 6 and 7	see Figure 6-6	- 180	+180	V
Tstg	storage temperature		- 50	+150	℃
Tamb	ambient temperature		- 40	+85	℃
Tvj	virtual junction temperature	[2]	- 40	+150	℃
VESD	electrostatic discharge voltage	[3]	- 2000	+2000	V
		[4]	- 150	+150	V

[1]   Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

[2]   An alternative definition of virtual junction temperature is: Tvj = Tamb + Pd × Rth(vj-a), where Rth(j-a) is a fixed value to be used for the calculation of Tvj. The rating for Tvj limits the allowable combinations of power dissipation (Pd) and ambient temperature (Tamb).

[3]   Classification A: human body model; C = 100 pF; R = 1500 Ω ; V = ±2000 V.

[4]   Classification B: machine model; C = 200 pF; R = 25 Ω ; V = ±150 V.

6.2 Recommended Operating Conditions

Symbol	Parameter	Conditions	Min	Max	Unit
VCC	supply voltage		4.5	5.5	V
ICC	supply current	Standby mode	-	220	μA
1/tbit	maximum transmission speed	non-return-to-zero	-	1	MBd
VCAN	CANH, CANL input/output voltage		-24	+24	V
Vdiff	differential bus voltage		1.5	3.0	V
tPD	propagation delay	High-speed mode	-	60	ns
Tamb	ambient temperature		-40	+85	° C

6.3 Thermal Data

Symbol	Parameter	Conditions	Typ	Unit
Rth(j-a)	thermal resistance from junction to ambient	in free air	165	K/W

6.4 Characteristics

VCC = 4.5 V to 5.5 V; Tamb = -40° C to +85° C; RL = 60 Ω; I8 > -10 µA; unless otherwise specified; all voltages referenced to ground (pin 2); positive input current; all parameters are guaranteed over the ambient temperature range by design, but only 100 % tested at +25 ° C.

Sym.	Parameter	Conditions	Min	Typ	Max	Unit
Power Supply consumption						
I3	Supply current	Dominant; $V_1 = 1$ V; Vcc=5.1V	-	-	80	mA
		Dominant; $V_1 = 1$ V; Vcc=5.25V	-	-	85	mA
		Dominant; $V_1 = 1$ V; Vcc=5.5V			90	mA
		Recessive; $V_1 = 4$ V; $R_8 = 47$ kΩ	-	-	12	mA
		Standby; Tamb < 80° C [1]	-	170	220	µA
DC bus transmitter						
V_{IH}	HIGH-level input voltage	output recessive	$0.7V_{CC}$	-	Vcc + 0.3	V
V_{IL}	LOW-level input voltage	output dominant	-0.3	-	$0.3V_{CC}$	V
I_{IH}	HIGH-level input current	$V_1 = 4$ V	-220	-	+40	µA
I_{IL}	LOW-level input current	$V_1 = 1$ V	-110	-	-650	µA
$V_{6,7}$	recessive bus voltage	$V_1 = 4$ V; no load	2.0	-	3.0	V
I_{LO}	off-state output leakage current	$2 V < (V_6, V_7) < 7 V$	-2	-	+3	mA
		$5 V < (V_6, V_7) < 18V$	-5	-	+12	mA
V_7	CANH output voltage	$V_1 = 1$ V	2.75	-	4.5	V
V_6	CANL output voltage	$V_1 = 1$ V	0.5	--	2.25	V
$\triangle V_{6,7}$	difference between output voltage at pins 6 and 7	$V_1 = 1$ V	1.5	-	3.0	V
		$V_1 = 1$ V; $R_L = 45Ω$; $VCC \geqslant 4.5V$	1.5	-	-	V
		$V_1 = 4$ V; no load	-500	-	+50	mV
I_{sc7}	short-circuit CANH current	$V_7 = -5V$	-	-	-220	mA
		$V_7 = -32V$	-	-100	-	mA
I_{sc6}	short-circuit CANL current	$V_6 = 32$ V	-	-	220	mA
DC bus receiver: V1 = 4 V; pins 6 and 7 externally driven; - 2V< (V6, V7) < 7 V; unless otherwise specified						
$V_{diff(r)}$	differential input voltage (recessive)		-1.0	-	+0.5	V
		$7 V < (V_6, V_7) <12 V$, non-Standy mode	-1.0	-	+0.4	V
$V_{diff(d)}$	differential input voltage (dominant)		0.9	-	5.0	V
		$7 V < (V_6, V_7) <12 V$, non-Standy mode	1.0	-	5.0	V
Vdiff (hys)	differential input hysteresis	see Figure 6-3	-	160	-	mV
VOH	HIGH-level output voltage	pin 4; I4 = -100 µA	0.8VCC	-	VCC	V
VOL	LOW-level output voltage	pin 4; I4 = 1 mA	0	-	0.2VCC	V
		I4 = 10 mA	0	-	1.5	V
Ri	input resistance	CANH, CANL	4.7	-	30	kΩ
Rdiff	differential input resistance		19.2	-	120	kΩ
Ci	input capacitance	CANH, CANL	-	-	20	pF
Cdiff	differential input capacitance		-	-	10	pF
Reference output						
Vref	reference output voltage	$V_8 = 1$ V; $|I5| < 50$ µA	0.45VCC	-	0.55VCC	V
		$V_8 = 4$ V; $|I5| < 5$ µA	0.4VCC	-	0.6VCC	V
Timing (CL = 100 pF; see Figure 6-1, Figure 6-2, Figure 6-4 and Figure 6-5)						
tbit	minimum bit time	Rs = 0 Ω	-	-	1	µs
tonTXD	delay TXD to bus active	Rs = 0 Ω	-	-	60	ns

[1] I1 = I4 = I5 = 0 mA; 0 V < V6 < VCC; 0 V < V7 < VCC; V8 = VCC

Sym.	Parameter	Conditions	Min	Typ	Max	Unit
Timing (CL = 100 pF; see Figure 6-1, Figure 6-2, Figure 6-4 and Figure 6-5)						
toffTXD	delay TXD to bus inactive	Rs = 0 Ω	-	45	90	ns
tonRXD	delay TXD to receiver active	Rs = 0 Ω	-	65	130	ns
toffRXD	delay TXD to receiver inactive	Rs = 0 Ω; VCC= 4.5v to 5.1V	-	90	200	ns
		Rs=0 Ω; VCC= 4.5v to 5.1V;Tamb < 85 °C	-	90	170	ns
		Rs = 0 Ω	-	100	200	ns
		Rs = 47 kΩ	-	300	450	ns
tonRXD	delay TXD to receiver active	Rs = 47 kΩ	-	450	580	ns
		Rs = 24 kΩ	-	280	350	ns
toffRXD	delay TXD to receiver inactive	Rs = 47 kΩ	-	280	500	ns
		Rs = 24 kΩ	-	230	350	ns
|SR|	differential output voltage slew rate	Rs = 47 kΩ	-	10	-	V/us
tWAKE	wake-up time from Standby	via pin 8	-	-	25	us
tdRXDL	bus dominant to RXD LOW	V_8 = 4 V; Standby mode	-	-	8	us
Standby/Slope Control (pin 8)						
V_8	input voltage for high-speed		-	-	0.3VCC	V
I_8	input current for high-speed	V_8 = 0V	-	-	- 500	uA
Vstb	input voltage for Standby mode		0.75VCC	-	-	V
Islope	slope control mode current		-10	-	-200	uA
Vslope	slope control mode voltage		0.4VCC	-	0.6VCC	V

[1] I1 = I4 = I5 = 0 mA; 0 V < V6 < VCC; 0 V < V7 < VCC; V8 = VCC;
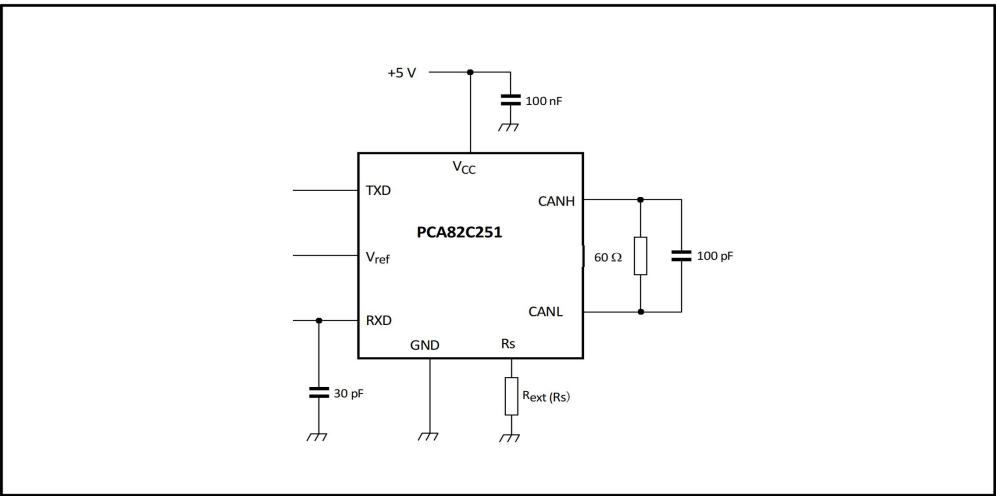[2] This is valid for the receiver in all modes: High-speed, Slope control and Standby.
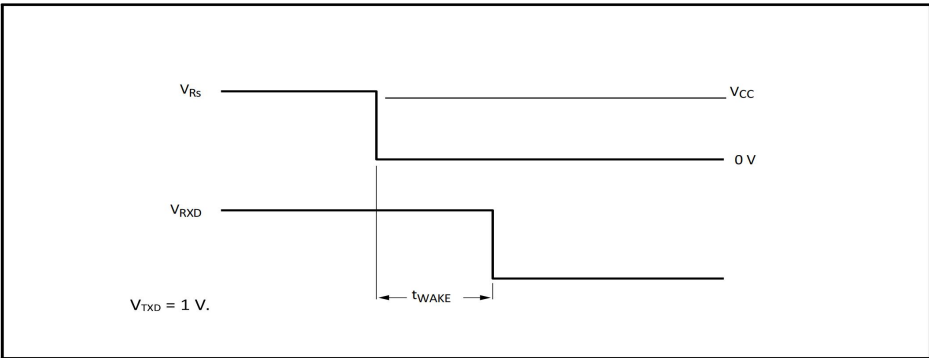

Figure 6-1 : Test circuit for dynamic characteristics

Figure 6-2 : Switching Times Test Circuits.

Figure 6-3. Sink Current Delay Times vs. Input 0 V Enable Switching

Figure 6-4. Bidirectional DC Motor Control

$V_{Rs} = 4$ V; $V_{TXD} = 4$ V.
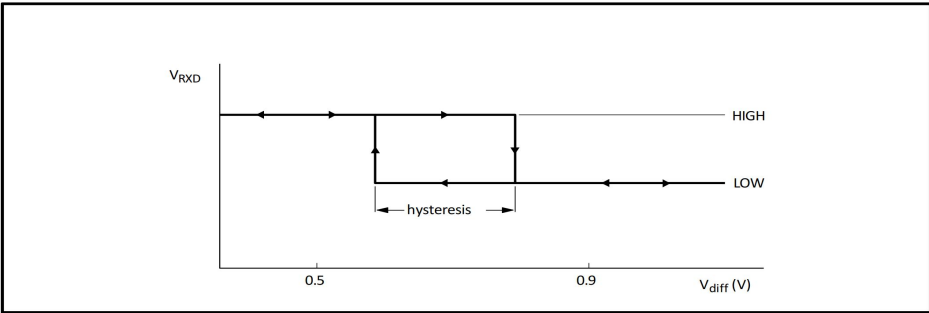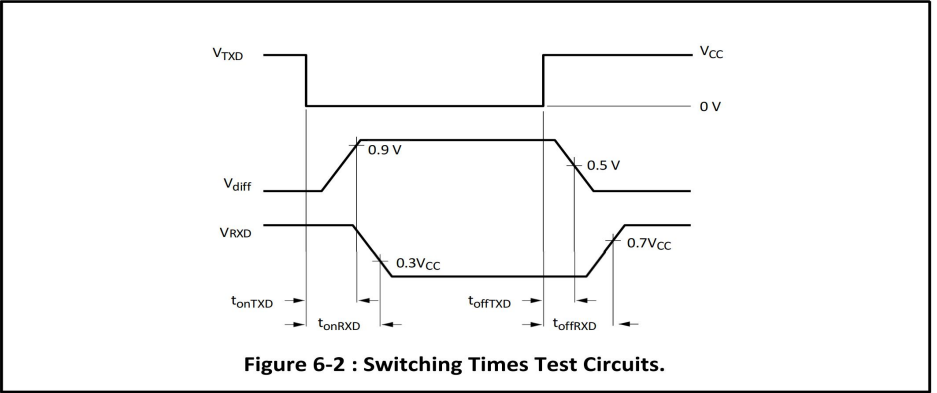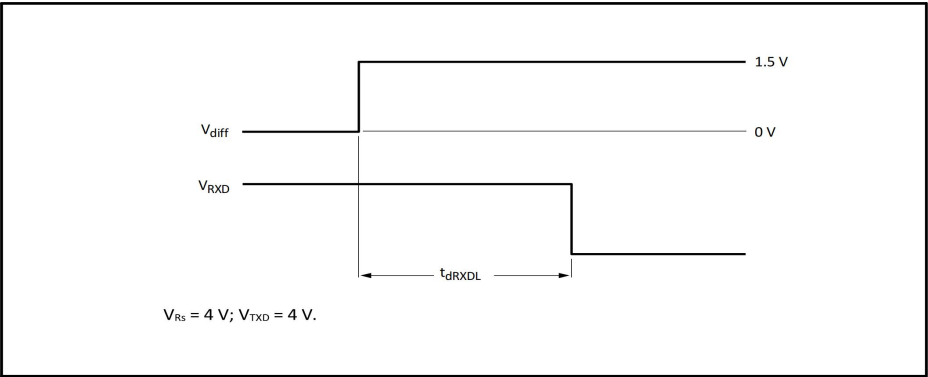
Figure 6-5. Timing diagram for bus dominant to RXD LOW

The waveforms of the applied transients shall be in accordance with *"ISO 7637 part 1"*, test pulses 1, 2, 3a and 3b.

Figure 6-6. Test circuit for transients

7. ORDERING INFORMATION

Ordering Information

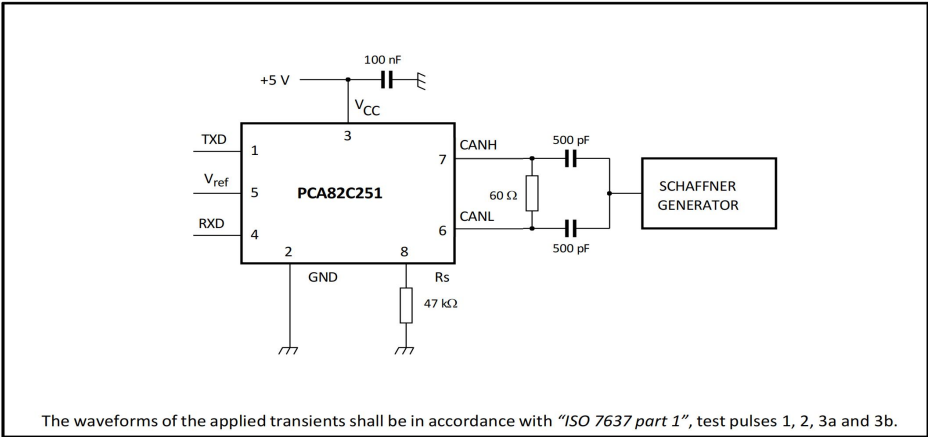| Part Number	Device Marking	Package Type	Body size (mm)	Temperature (°C)	MSL	Transport Media	Package Quantity
PCA82C251	PCA82C251	SOP8	4.90 * 3.90	- 40 to 85	MSL3	T&R	2500
PCA82C251	PCA82C251	DIP8	9.25 * 6.38	- 40 to 85	MSL3	Tube 50	2000

8. DIMENSIONAL DRAWINGS

SOP8

UNIT:mm

	MIN	NOM	MAX
A	1.450	1.550	1.650
A1	0.100	0.150	0.200
A2	1.300	1.400	1.500
A3	0.600	0.650	0.700
b	0.380	—	0.510
e	1.240	1.270	1.300
D	4.800	4.900	5.000
E	5.800	6.000	6.200
E1	3.800	3.900	4.000
L	0.450	0.600	0.750
L1	—	0.25BSC	—

DIP8

PIN1

UNIT:mm

	MIN	NOM	MAX
A	3.600	3.800	4.000
A1	3.786	3.886	3.986
A2	3.200	3.300	3.400
A3	1.550	1.600	1.650
b	0.440	—	0.490
e	2.510	2.540	2.570
D	9.150	9.250	9.350
E	7.800	8.500	9.200
E1	6.280	6.380	6.480
L	3.000	—	—