

### Features

- Single-Supply Operation from +1.8V ~ +5.5V
  - Rail-to-Rail Input / Output
  - Gain-Bandwidth Product: 1MHz (Typ)
  - Low Input Bias Current: 10pA (Typ)
  - Low Offset Voltage: 0.6mV (Typ)
  - Quiescent Current: 44μA per Amplifier (Typ)
  - Embedded RF Anti-EMI Filter
  - Embedded RF Anti-EMI Filter
  - Operating Temperature: -40°C ~ +125°C
  - Small Package:
- GS6012B Available in SOP-8 and MSOP-8 Packages  
GS6014B Available in SOP-14 and TSSOP-14 Packages

### General Description

The GS6012B/6014B have a high gain-bandwidth product of 1MHz, a slew rate of 0.6V/μs, and a quiescent current of 44 μA/amplifier at 5V. The GS6012B/6014B is designed to provide optimal performance in low voltage and low noise systems. They provide rail-to-rail output swing into heavy loads. The input common mode voltage range includes ground, and the input offset voltage range is 0.4mV~0.8mV for GS6012B/6014B. They are specified over the extended industrial temperature range (-40°C to +125°C). The operating range is from 1.8V to 5.5V. The GS6012B/6014B dual is available in Green SOP-8 and MSOP-8 packages. The GS6014B Quad is available in Green SOP-14 and TSSOP-14 packages.

### Applications

- ASIC Input or Output Amplifier
- Sensor Interface
- Medical Communication
- Smoke Detectors
- Audio Output
- Piezoelectric Transducer Amplifier
- Medical Instrumentation
- Portable Systems

### Pin Configuration

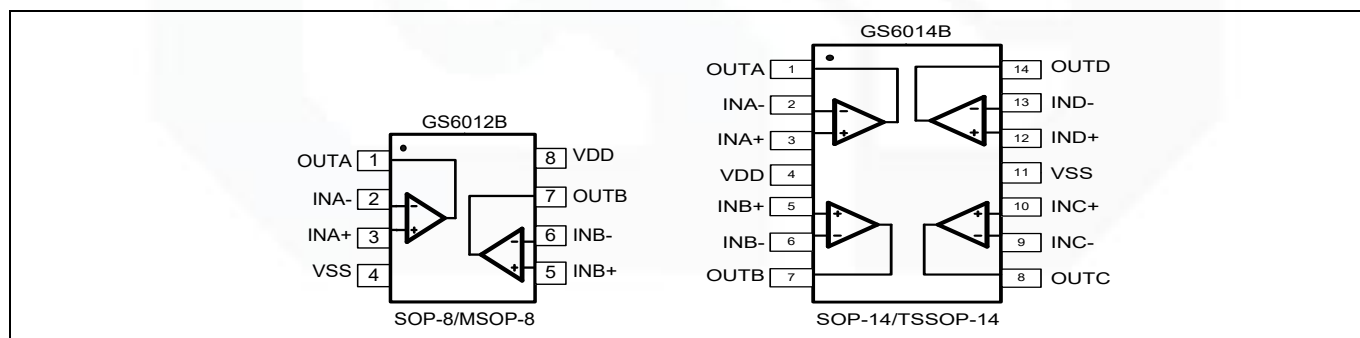


Figure 1. Pin Assignment Diagram

## Absolute Maximum Ratings

Condition	Min	Max
Power Supply Voltage (V <sub>DD</sub> to V <sub>SS</sub> )	-0.5V	+8V
Analog Input Voltage (IN+ or IN-)	V <sub>SS</sub> -0.5V	V <sub>DD</sub> +0.5V
PDB Input Voltage	V <sub>SS</sub> -0.5V	+7V
Operating Temperature Range	-40°C	+125°C
Junction Temperature	+160°C	
Storage Temperature Range	-55°C	+150°C
Lead Temperature (soldering, 10sec)	+260°C	
Package Thermal Resistance (T <sub>A</sub> =+25°C)		
SOP-14, θ <sub>JA</sub>	120°C/W	
TSSOP-14, θ <sub>JA</sub>	180°C/W	
SOP-8, θ <sub>JA</sub>	125°C/W	
MSOP-8, θ <sub>JA</sub>	216°C/W	
ESD Susceptibility		
HBM <sup>①</sup>	± 5KV	
CDM	± 2KV	
Latch up	± 350mA	

**Note:** Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability. ① OUT to  $V_{DD}+$  only pass 2kV.

## Package/Ordering Information

MODEL	CHANNEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION
GS6012B	Dual	GS6012B-SR	SOP-8	Tape and Reel,4000	GS6012B
		GS6012B-MR	MSOP-8	Tape and Reel,3000	GS6012B
GS6014B	Quad	GS6014B-TR	TSSOP-14	Tape and Reel,3000	GS6014B
		GS6014B-SR	SOP-14	Tape and Reel,2500	GS6014B

## Electrical Characteristics

(At  $V_S = +5V$ ,  $R_L = 100k\Omega$  connected to  $V_S/2$ , and  $V_{OUT} = V_S/2$ ,  $T_A = 25^\circ C$ , unless otherwise noted.)

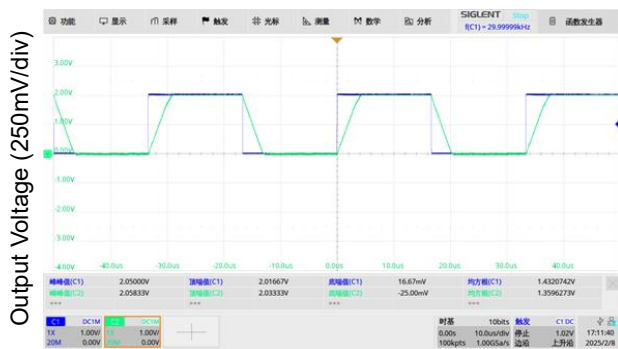
PARAMETER	SYMBOL	CONDITIONS				
			TYP	MAX	MIN	UNITS
INPUT CHARACTERISTICS						
Input Offset Voltage	V <sub>OS</sub>	V <sub>CM</sub> = 0V to (V <sub>S</sub> -1.2V)	0.6	0.8	0.4	mV
Input Bias Current	I <sub>B</sub>		10			pA
Input Offset Current	I <sub>OS</sub>		1			pA
Common-Mode Voltage Range	V <sub>CM</sub>	V <sub>S</sub> = 5.5V	-0.1 to +5.6			V
Common-Mode Rejection Ratio	CMRR	V <sub>S</sub> = 5.5V, V <sub>CM</sub> = -0.1V to 4V	70		62	dB
		V <sub>S</sub> = 5.5V, V <sub>CM</sub> = -0.1V to 5.6V	68		56	
Open-Loop Voltage Gain	A <sub>OL</sub>	R <sub>L</sub> = 5kΩ, V <sub>O</sub> = +0.1V to +4.9V	80		70	dB
		R <sub>L</sub> = 10kΩ, V <sub>O</sub> = +0.1V to +4.9V	100		94	
Input Offset Voltage Drift	ΔV <sub>OS</sub> /ΔT		2.7			μV/°C
OUTPUT CHARACTERISTICS						
Output Voltage Swing from Rail	V <sub>OH</sub>	R <sub>L</sub> = 100kΩ	4.997			mV
	V <sub>OL</sub>	R <sub>L</sub> = 100kΩ	5	30		mV
	V <sub>OH</sub>	R <sub>L</sub> = 10kΩ	4.992			mV
	V <sub>OL</sub>	R <sub>L</sub> = 10kΩ	8	30		mV
Output Current	I <sub>SINK</sub>	R <sub>L</sub> = 10Ω to V <sub>S</sub> /2	60		40	mA
	I <sub>Source</sub>		60		40	
POWER SUPPLY						
Operating Voltage Range					1.8	V
				5.5		V
Power Supply Rejection Ratio	PSRR	V <sub>S</sub> = +2V to +5V, V <sub>CM</sub> = +0.5V	82		60	dB
Quiescent Current / Amplifier	I <sub>Q</sub>		44	75	30	μA
DYNAMIC PERFORMANCE (CL = 100pF)						
Gain-Bandwidth Product	GBP		1			MHz
Slew Rate	SR	G = +1, 2V Output Step	0.6			V/μs
Settling Time to 0.1%	t <sub>s</sub>	G = +1, 2V Output Step	5.3			μs
Overload Recovery Time		V <sub>IN</sub> · Gain = V <sub>S</sub>	2.6			μs
NOISE PERFORMANCE						
Voltage Noise Density	e <sub>n</sub>	f = 1kHz	150			nV / √Hz
		f = 10kHz	70			nV / √Hz

## Typical Performance characteristics

At  $T_A=+25^{\circ}\text{C}$ ,  $V_S=5\text{V}$ ,  $R_L=100\text{K}\Omega$  connected to  $V_S/2$  and  $V_{OUT}=V_S/2$ , unless otherwise noted.

Large Signal Transient Response

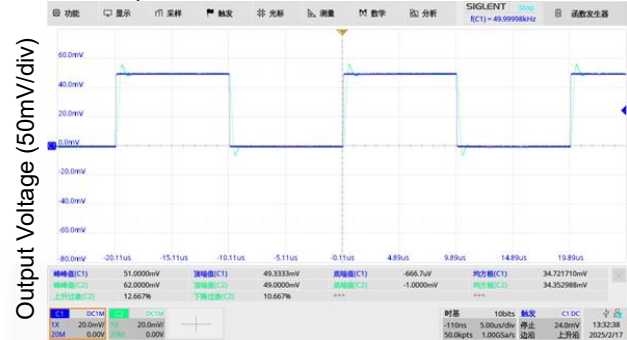
$C_L=100\text{pF}$   $R_L=100\text{K}\Omega$



Time(10μs/div)

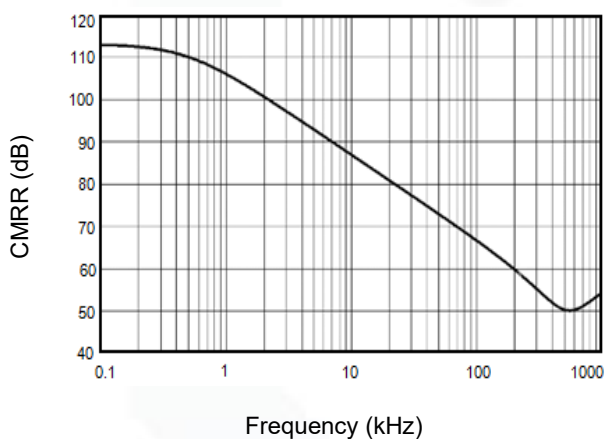
Small Signal Transient Response

$C_L=100\text{pF}$   $R_L=100\text{K}\Omega$



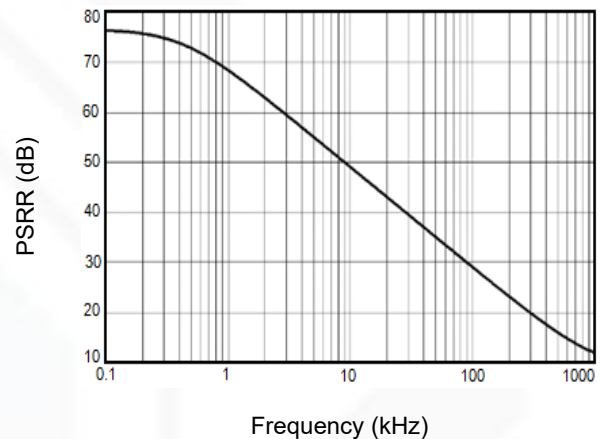
Time(5μs/div)

CMRR vs. Frequency



Frequency (kHz)

PSRR vs. Frequency



Frequency (kHz)

Overload Recovery Time

$V_S=5\text{V}$   $G=-10$



Time(20μs/div)

Overload Recovery Time

$V_S=5\text{V}$   $G=-10$

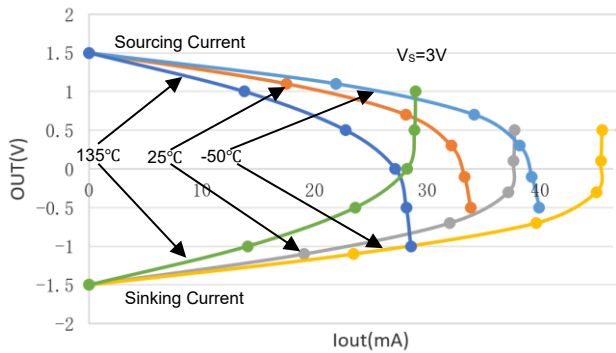


Time(20μs/div)

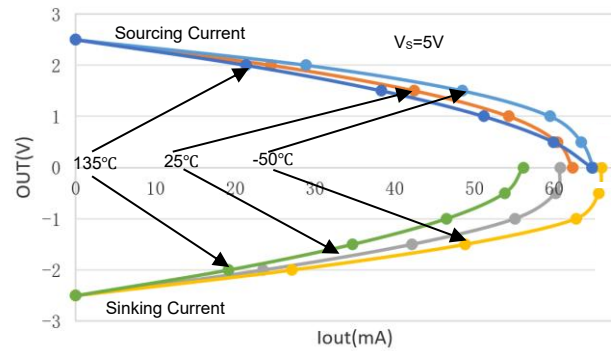
## Typical Performance characteristics

At  $T_A=+25^{\circ}\text{C}$ ,  $R_L=100\text{K}\Omega$  connected to  $V_S/2$  and  $V_{OUT}=V_S/2$ , unless otherwise noted.

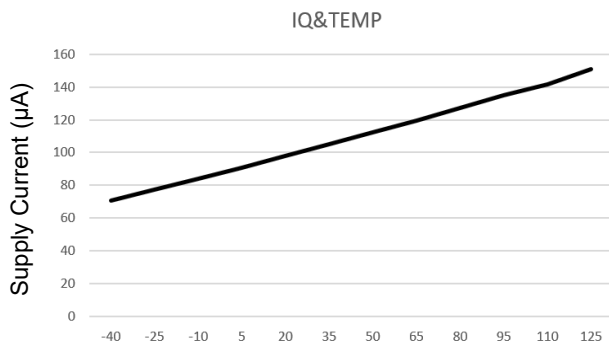
Vout VS Iout



Vout VS Iout



Supply Current vs. Temperature



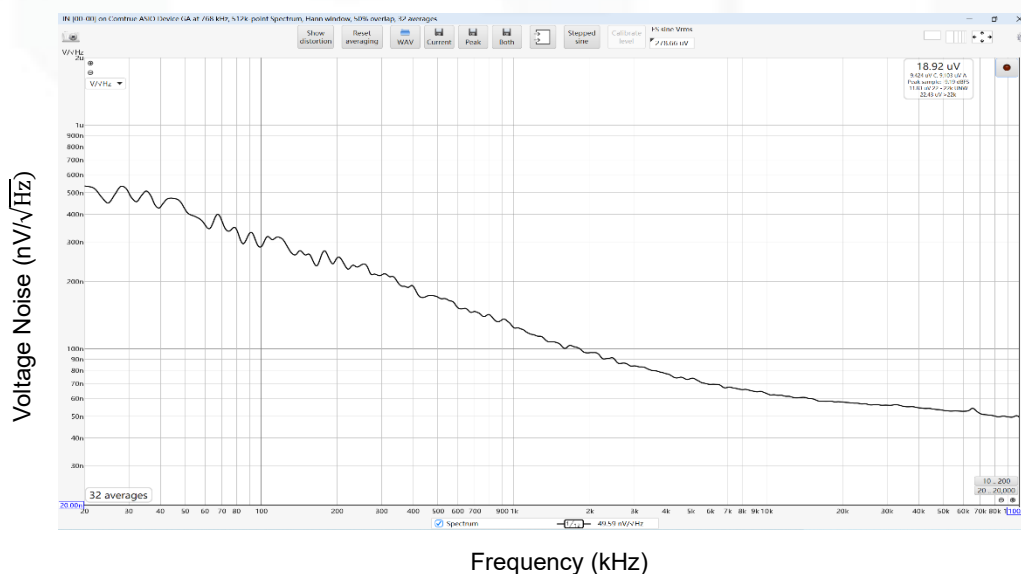
Open Loop Gain, Phase Shift vs. Frequency



Temperature (°C)

Frequency (kHz)

Input Voltage Noise Spectral Density vs. Frequency



## Application Note

### Size

GS6012B/6014B series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The small footprints of the GS6012B/6014B packages save space on printed circuit boards and enable the design of smaller electronic products.

### Power Supply Bypassing and Board Layout

GS6012B/6014B series operates from a single 1.8V to 5.5V supply or dual  $\pm 0.9V$  to  $\pm 2.75V$  supplies. For best performance, a 0.1 $\mu F$  ceramic capacitor should be placed close to the  $V_{DD}$  pin in single supply operation. For dual supply operation, both  $V_{DD}$  and  $V_{SS}$  supplies should be bypassed to ground with separate 0.1 $\mu F$  ceramic capacitors.

### Low Supply Current

The low supply current (typical 44 $\mu A$  per channel) of GS6012B/6014B will help to maximize battery life. They are ideal for battery powered systems.

### Operating Voltage

GS6012B/6014B operates under wide input supply voltage (1.8V to 5.5V). In addition, all temperature specifications apply from  $-40^{\circ}C$  to  $+125^{\circ}C$ . Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-Ion battery lifetime.

### Rail-to-Rail Input

The input common-mode range of GS6012B/6014B extends 100mV beyond the supply rails ( $V_{SS}-0.1V$  to  $V_{DD}+0.1V$ ). This is achieved by using complementary input stage. For normal operation, inputs should be limited to this range.

### Rail-to-Rail Output

Rail-to-Rail output swing provides maximum possible dynamic range at the output. This is particularly important when operating in low supply voltages. The output voltage of GS6012B/6014B can typically swing to less than 10mV from supply rail in light resistive loads ( $>100k\Omega$ ), and 30mV of supply rail in moderate resistive loads (10k $\Omega$ ).

### Capacitive Load Tolerance

The GS6012B/6014B is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain. Figure 2 shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

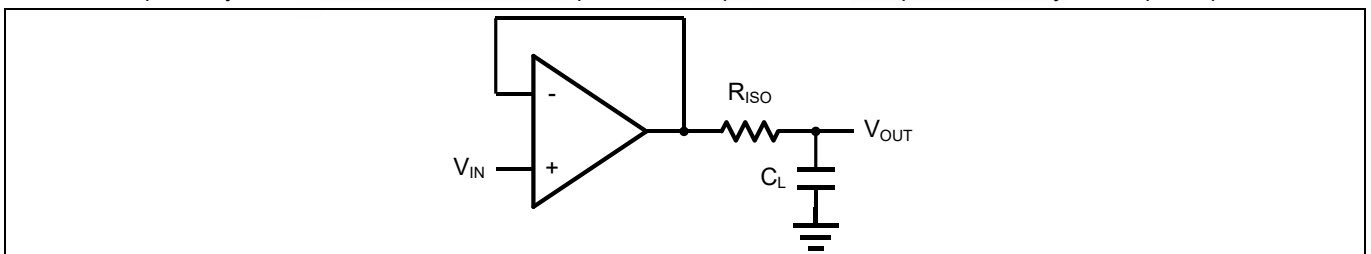


Figure 2 Indirectly Driving a Capacitive Load Using Isolation Resistor

The bigger the  $R_{ISO}$  resistor value, the more stable  $V_{OUT}$  will be. However, if there is a resistive load  $R_L$  in parallel with the capacitive load, a voltage divider (proportional to  $R_{ISO}/R_L$ ) is formed, this will result in a gain error.

The circuit in Figure 3 is an improvement to the one in Figure 2.  $R_F$  provides the DC accuracy by feed-forward the  $V_{IN}$  to  $R_L$ .  $C_F$  and  $R_{ISO}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of  $C_F$ . This in turn will slow down the pulse response.

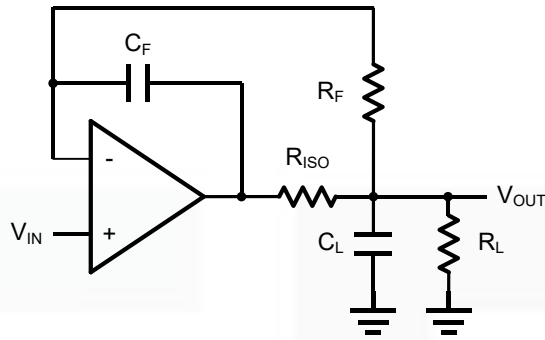


Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy

## Typical Application Circuits

### Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common to the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4. shown the differential amplifier using GS6012B/6014B.

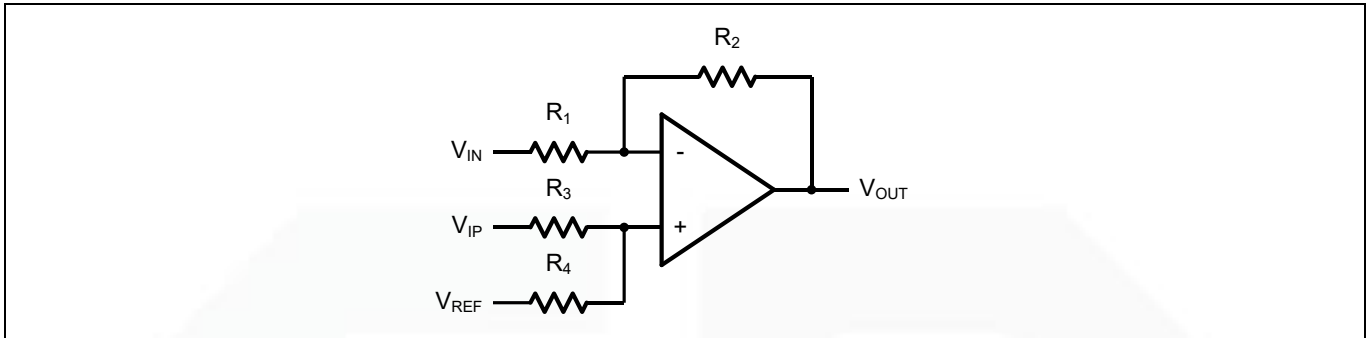


Figure 4. Differential Amplifier

$$V_{OUT} = \left( \frac{R_1 + R_2}{R_3 + R_4} \right) \frac{R_4}{R_1} V_{IN} - \frac{R_2}{R_1} V_{IP} + \left( \frac{R_1 + R_2}{R_3 + R_4} \right) \frac{R_3}{R_1} V_{REF}$$

If the resistor ratios are equal (i.e.  $R_1 = R_3$  and  $R_2 = R_4$ ), then

$$V_{OUT} = \frac{R_2}{R_1} (V_{IP} - V_{IN}) + V_{REF}$$

### Low Pass Active Filter

The low pass active filter is shown in Figure 5. The DC gain is defined by  $-R_2/R_1$ . The filter has a -20dB/decade roll-off after its corner frequency  $f_c = 1/(2\pi R_3 C_1)$ .

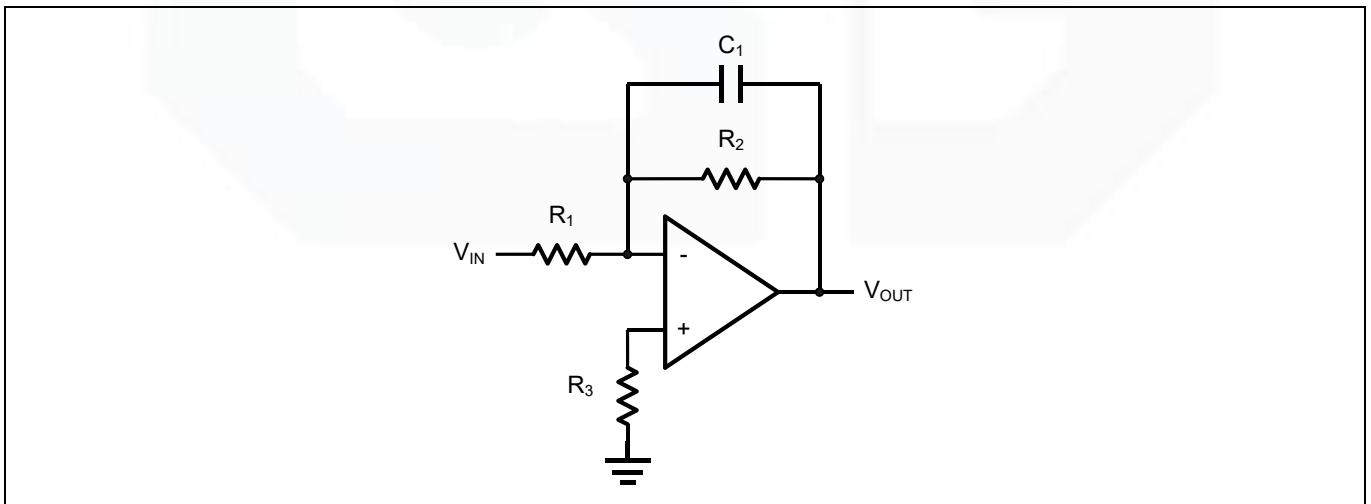


Figure 5. Low Pass Active Filter



### Instrumentation Amplifier

The triple GS6012B/6014B can be used to build a three-op-amp instrumentation amplifier as shown in Figure 6. The amplifier in Figure 6 is a high input impedance differential amplifier with gain of  $R_2/R_1$ . The two differential voltage followers assure the high input impedance of the amplifier.

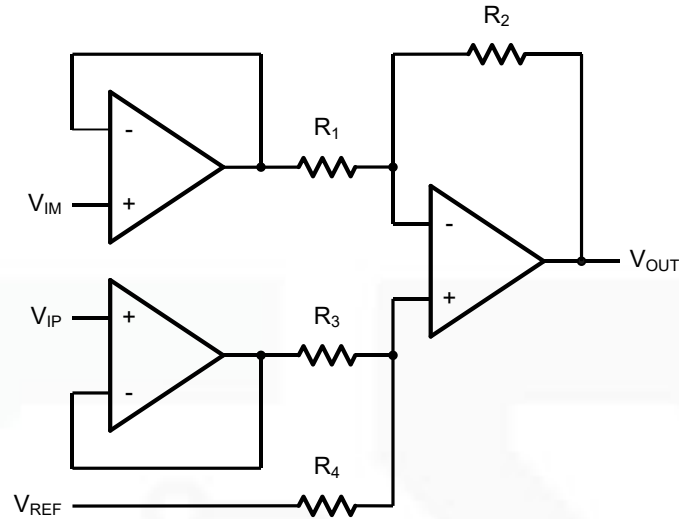
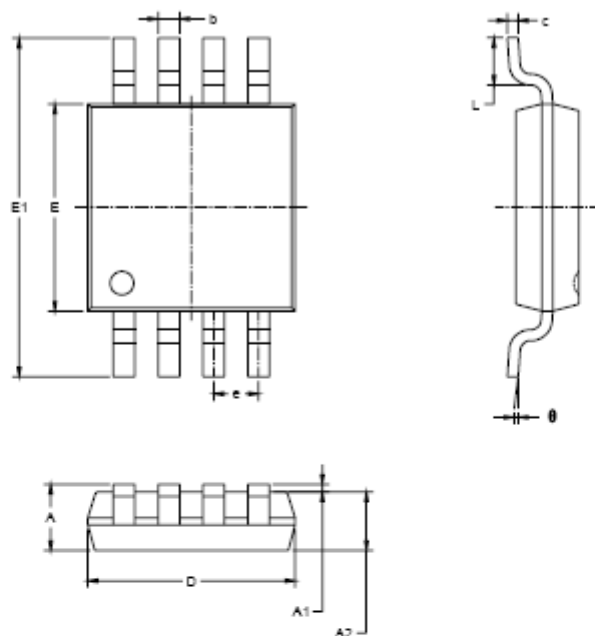


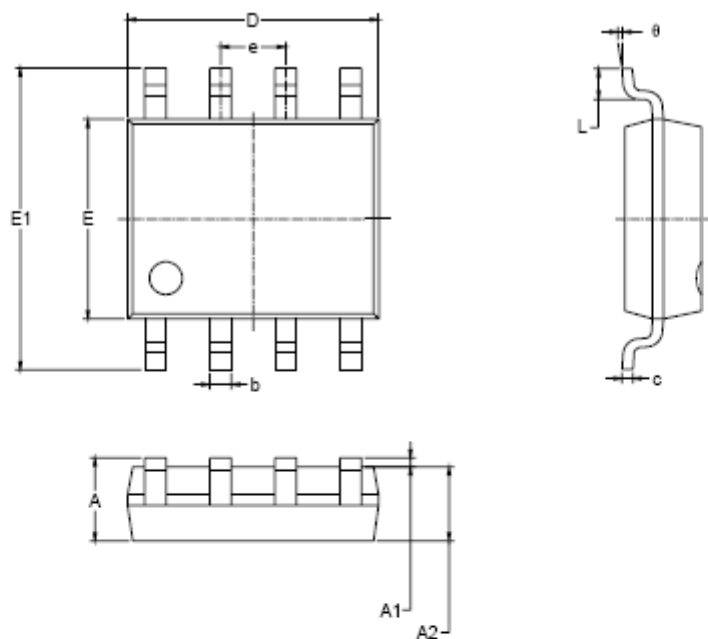
Figure 6. Instrument Amplifier

## Package Information

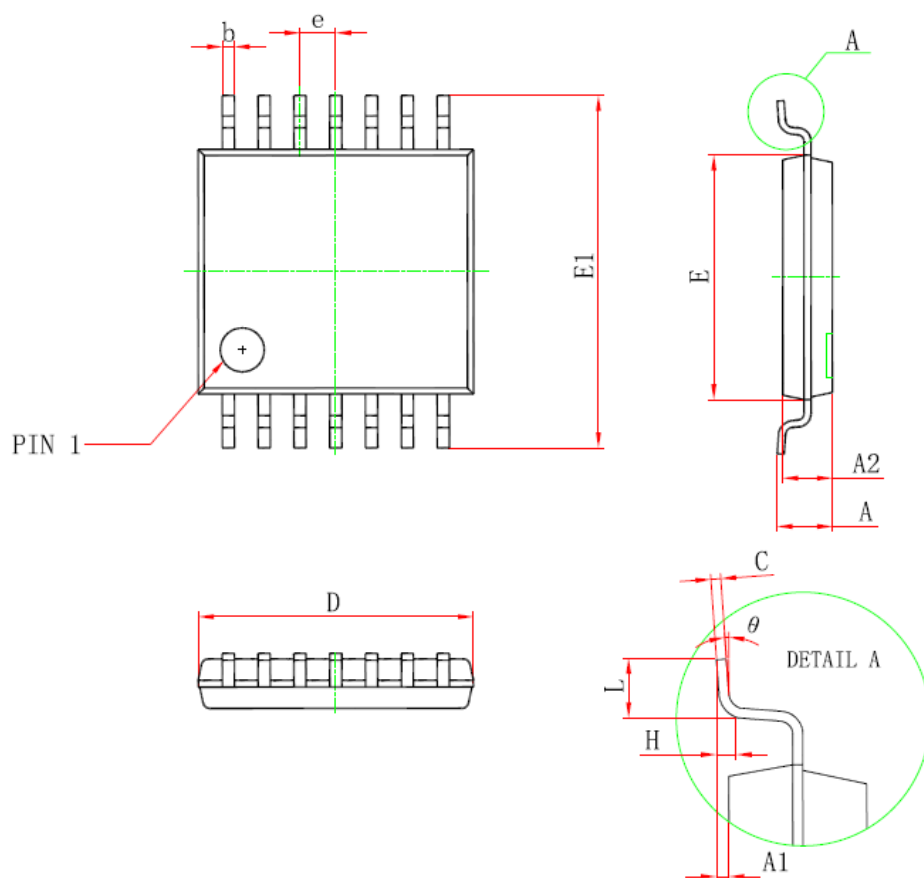
### MSOP-8



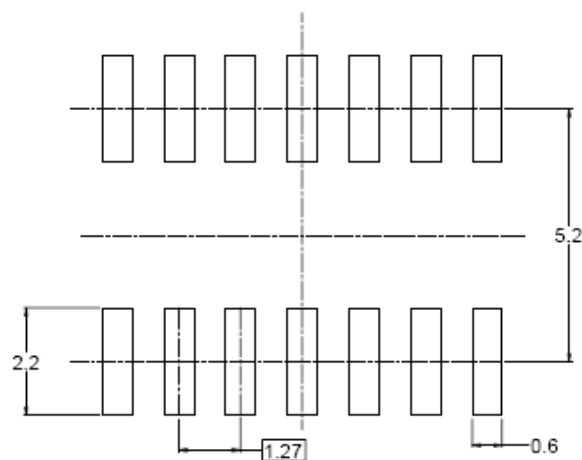
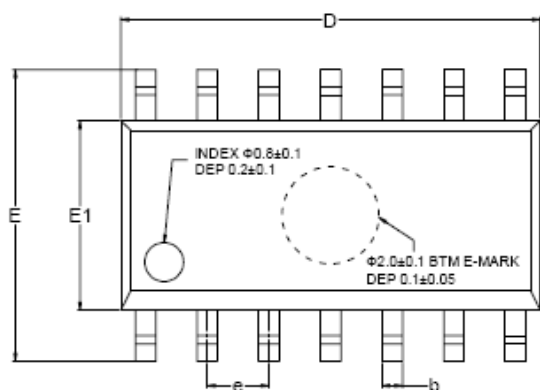
Symbol	Dimensions In Millimeters		Dimensions In Inches	
	MIN	MAX	MIN	MAX
A	0.820	1.100	0.032	0.043
A1	0.020	0.150	0.001	0.008
A2	0.750	0.950	0.030	0.037
b	0.250	0.380	0.010	0.015
c	0.090	0.230	0.004	0.009
D	2.900	3.100	0.114	0.122
E	2.900	3.100	0.114	0.122
E1	4.750	5.050	0.187	0.199
e	0.650 BSC		0.026 BSC	
L	0.400	0.800	0.016	0.031
θ	0°	8°	0°	8°

**SOP-8**


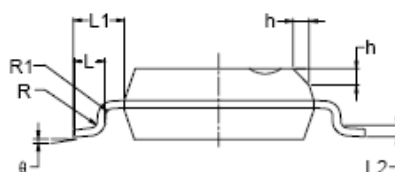
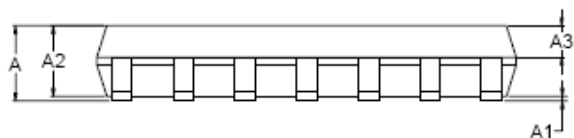
Symbol	Dimensions In Millimeters		Dimensions In Inches	
	MIN	MAX	MIN	MAX
A	1.350	1.750	0.053	0.069
A1	0.100	0.250	0.004	0.010
A2	1.350	1.550	0.053	0.061
b	0.330	0.510	0.013	0.020
c	0.170	0.250	0.006	0.010
D	4.700	5.100	0.185	0.200
E	3.800	4.000	0.150	0.157
E1	5.800	6.200	0.228	0.244
e	1.27 BSC		0.050 BSC	
L	0.400	1.270	0.016	0.050
θ	0°	8°	0°	8°



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
D	4.900	5.100	0.193	0.201
E	4.300	4.500	0.169	0.177
b	0.190	0.300	0.007	0.012
c	0.090	0.200	0.004	0.008
E1	6.250	6.550	0.246	0.258
A		1.200		0.047
A2	0.800	1.000	0.031	0.039
A1	0.050	0.150	0.002	0.006
e	0.65 (BSC)		0.026 (BSC)	
L	0.500	0.700	0.020	0.028
H	0.25(TYP)		0.01(TYP)	
$\theta$	1°	7°	1°	7°

**SOP-14**


RECOMMENDED LAND PATTERN (Unit: mm)



Symbol	Dimensions In Millimeters			Dimensions In Inches		
	MIN	MOD	MAX	MIN	MOD	MAX
A	1.35		1.75	0.053		0.069
A1	0.10		0.25	0.004		0.010
A2	1.25		1.65	0.049		0.065
A3	0.55		0.75	0.022		0.030
b	0.36		0.49	0.014		0.019
D	8.53		8.73	0.336		0.344
E	5.80		6.20	0.228		0.244
E1	3.80		4.00	0.150		0.157
e	1.27 BSC			0.050 BSC		
L	0.45		0.80	0.018		0.032
L1	1.04 REF			0.040 REF		
L2	0.25 BSC			0.01 BSC		
R	0.07			0.003		
R1	0.07			0.003		
h	0.30		0.50	0.012		0.020
θ	0°		8°	0°		8°