

## 1.Features

- $V_{DS(V)}=60V$
- $I_D=12A$
- $R_{DS(ON)}<35m\Omega(V_{GS}=5V)$
- Power Supplies
- Converters
- Power Motor Controls
- Bridge Circuits

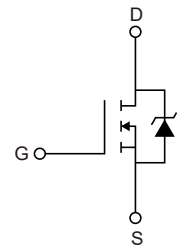
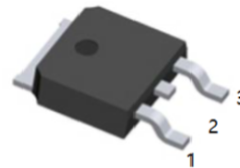
## 2.Applications

- Tighter  $V_{SD}$  Specification
- Lower Diode Reverse Recovery Time
- Lower Reverse Recovery Stored Charge
- Lower  $R_{DS(ON)}$
- Lower  $V_{DS(ON)}$

## 3.Pinning information

Pin	Symbol	Description
1	G	GATE
2	D	SOURCE
3	S	DRAIN

TO-252(DPAK)  
top view



## 4.Absolute Maximum Ratings $T_J=25^{\circ}C$

Parameter	Symbol	Rating	Units
Drain-to-Source Voltage	$V_{DSS}$	60	$V_{dc}$
Drain-to-Gate Voltage ( $R_{GS}=1M\Omega$ )	$V_{DGR}$	60	$V_{dc}$
Gate-to-Source Voltage, Continuous	$V_{GS}$	$\pm 15$	$V_{dc}$
- Non-Repetitive ( $t_p \leq 10$ ms)	$V_{GS}$	$\pm 20$	$V_{dc}$
Drain Current - Continuous @ $T_A=25^{\circ}C$	$I_D$	12	$A_{dc}$
- Continuous @ $T_A=100^{\circ}C$	$I_D$	10	$A_{dc}$
- Single Pulse ( $t_p \leq 10$ $\mu s$ )	$I_{DM}$	45	$A_{pk}$



Total Power Dissipation @ $T_A = 25^\circ\text{C}$	$P_D$	48	W
Derate above $25^\circ\text{C}$		0.32	W/ $^\circ\text{C}$
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ (Note 1)		2.1	W
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ (Note 2)		1.5	W
Operating and Storage Temperature Range	$T_J, T_{STG}$	-55 to 175	$^\circ\text{C}$
Single Pulse Drain-to-Source Avalanche Energy-Starting $T_J = 25^\circ\text{C}$ ( $V_{DD} = 25\text{Vdc}$ , $V_{GS} = 5\text{Vdc}$ $I_{L(pk)} = 11\text{A}$ , $L = 1\text{mH}$ , $V_{DS} = 60\text{Vdc}$ )	$E_{AS}$	61	mJ
Thermal Resistance-Junction-to-Case	$R_{\theta JC}$	3.13	$^\circ\text{C/W}$
-Junction-to-Ambient (Note 1)	$R_{\theta JA}$	71.4	$^\circ\text{C/W}$
-Junction-to-Ambient (Note 2)	$R_{\theta JA}$	100	$^\circ\text{C/W}$
Maximum Lead Temperature for Soldering Purposes, 1/8 in. from case for 10 seconds	$T_L$	260	$^\circ\text{C}$

Maximum ratings are those values beyond which device damage can occur. Maximum ratings applied to the device are individual stress limit values (not normal operating conditions) and are not valid simultaneously. If these limits are exceeded, device functional operation is not implied, damage may occur and reliability may be affected.

1. When surface mounted to an FR4 board using 1 in pad size (Cu area =  $1.127\text{ in}^2$ ).
2. When surface mounted to an FR4 board using the minimum recommended pad size (Cu area =  $0.412\text{ in}^2$ )



## 5. Electrical Characteristics @ $T_J = 25^\circ\text{C}$ (unless otherwise specified)

Parameter	Symbol	Conditions	Min	Typ	Max	Units
OFF CHARACTERISTICS						
Drain-to-Source Breakdown Voltage (Temperature Coefficient Positive) (Note 3)	$V_{(BR)DSS}$	$V_{GS}=0Vdc, I_D=250\mu Adc$	60	71.3		Vdc
				71.2		mV/°C
Zero Gate Voltage Drain Current	$I_{DSS}$	$V_{DS}=60Vdc, V_{GS}=0Vdc$			1	$\mu Adc$
		$V_{DS}=60Vdc, V_{GS}=0Vdc, T_J=150^{\circ}C$			10	$\mu Adc$
Gate-Body Leakage Current	$I_{GSS}$	$V_{GS}=\pm 15Vdc, V_{DS}=0Vdc$			$\pm 100$	nAdc
ON CHARACTERISTICS						
Gate Threshold Voltage(Note 3) (Negative Temperature Coefficient)	$V_{GS(th)}$	$V_{DS}=V_{GS}, I_D=250\mu Adc$	1	1.6	2	Vdc
				4.6		mV/°C
Static Drain-Source On-State Resistance	$R_{DS(ON)}$	$V_{GS}=4.5Vdc, I_D=10Adc$		35	45	mΩ
Static Drain-to-Source On-Resistance (Note 3)	$V_{DS(ON)}$	$V_{GS}=5Vdc, I_D=20Adc$		0.81	1.66	Vdc
		$V_{GS}=5Vdc, I_D=10Adc, T_J=150^{\circ}C$		0.72		Vdc
Forward Transconductance (Note 3)	$g_{FS}$	$V_{DS}=4Vdc, I_D=10Adc$		17.5		Mhos
DYNAMIC CHARACTERISTICS						
Input Capacitance	$C_{iss}$	$V_{DS}=25Vdc, V_{GS}=0Vdc$  $f=1MHz$		707	990	pF
Output Capacitance	$C_{oss}$			224	320	pF
Transfer Capacitance	$C_{rss}$			72	105	pF
SWITCHING CHARACTERISTICS						
Turn-On Delay Time	$t_{D(on)}$	$V_{DD}=30Vdc, I_D=20Adc$  $V_{GS}=5Vdc, R_G=9.1\Omega$ (Note 3)		9.6	20	ns
Rise Time	$t_r$			98	200	ns
Turn-Off Delay Time	$t_{D(off)}$			25	50	ns
Fall Time	$t_f$			62	120	ns
Gate Charge	$Q_T$	$V_{DS}=48Vdc, I_D=20Adc$  $V_{GS}=5Vdc$ (Note 3)		16.6	32	nC
	$Q_1$			5.5		nC
	$Q_2$			8.5		nC



DRAIN-SOURCE DIODE CHARACTERISTICS						
Forward On-Voltage	$V_{SD}$	$I_S=20A_{dc}, V_{GS}=0V$ (Note 3)		0.97	1.2	Vdc
		$I_S=20A_{dc}, V_{GS}=0V, T_J=150^{\circ}C$		0.85		Vdc
Reverse Recovery Time	$t_{rr}$	$I_S=20A, dI_S/dt=100 A/\mu s$ (Note 3) $V_{GS}=0V_{dc}$		42		ns
	$t_a$			30		ns
	$t_b$			12		ns
Reverse Recovery Stored Charge	$Q_{rr}$			0.66		$\mu C$

Notes:

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

3. Pulse Test: Pulse Width  $\leq 300 \mu s$ , Duty Cycle  $\leq 2\%$ .

4. Switching characteristics are independent of operating junction temperatures.



## 6.1 Typical Characteristics

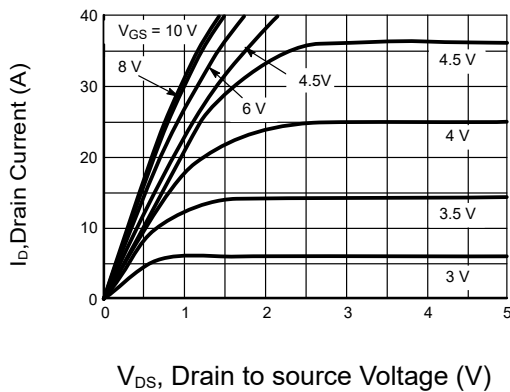


Figure 1: On-Region Characteristics

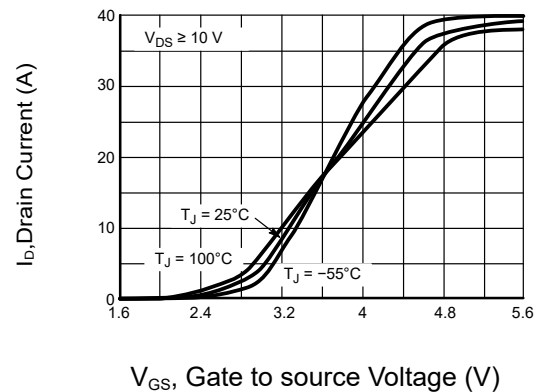


Figure 2: Transfer Characteristics

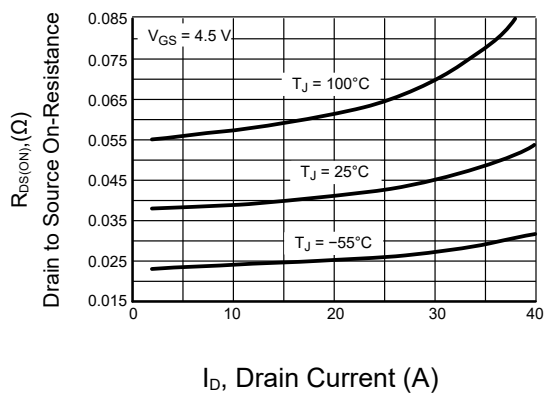


Figure 3: On-Resistance versus Drain Current and Temperature

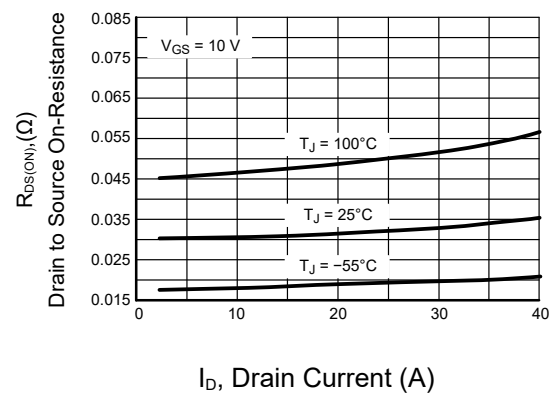


Figure 4: On-Resistance versus Drain Current and Gate Voltage

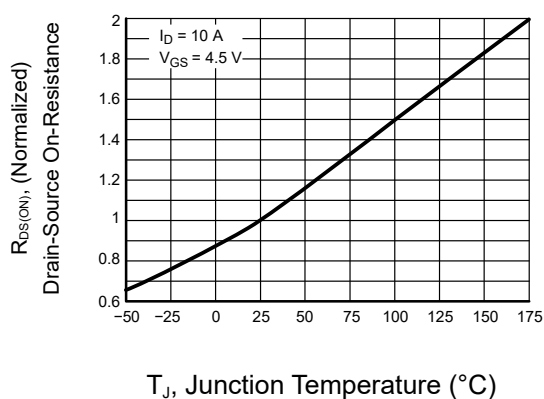


Figure 5: On-Resistance Variation with Temperature

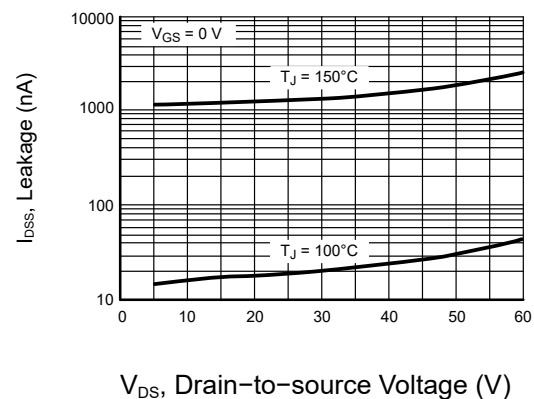


Figure 6: Drain-To-Source Leakage Current versus Voltage



## 6.2 Typical Characteristics

Switching behavior is most easily modeled and predicted by recognizing that the power MOSFET is charge controlled. The lengths of various switching intervals ( $t$ ) are determined by how fast the FET input capacitance can be charged by current from the generator.

The published capacitance data is difficult to use for calculating rise and fall because drain-gate capacitance varies greatly with applied voltage. Accordingly, gate charge data is used. In most cases, a satisfactory estimate of average input current ( $I_{G(AV)}$ ) can be made from a rudimentary analysis of the drive circuit so that

$$t = Q/I_{G(AV)}$$

During the rise and fall time interval when switching a resistive load,  $V_{GS}$  remains virtually constant at a level known as the plateau voltage,  $V_{GSP}$ . Therefore, rise and fall times may be approximated by the following:

$$t_r = Q_2 \times R_G / (V_{GG} - V_{GSP})$$

$$t_f = Q_2 \times R_G / V_{GSP}$$

where

$V_{GG}$  = the gate drive voltage, which varies from zero to

$V_{GG}$   $R_G$  = the gate drive resistance

and  $Q_2$  and  $V_{GSP}$  are read from the gate charge curve.

During the turn-on and turn-off delay times, gate current is not constant. The simplest calculation uses appropriate values from the capacitance curves in a standard equation for voltage change in an  $R_C$  network. The equations are:

$$t_{d(on)} = R_G C_{iss} \ln [V_{GG}/(V_{GG} - V_{GSP})]$$

$$t_{d(off)} = R_G C_{iss} \ln (V_{GG}/V_{GSP})$$

The capacitance ( $C_{iss}$ ) is read from the capacitance curve at a voltage corresponding to the off-state condition when calculating  $t_{d(on)}$  and is read at a voltage corresponding to the on-state when calculating  $t_{d(off)}$ .

At high switching speeds, parasitic circuit elements complicate the analysis. The inductance of the MOSFET source lead, inside the package and in the circuit wiring which is common to both the drain and gate current paths, produces a voltage at the source which reduces the gate drive current. The voltage is determined by  $L di/dt$ , but since  $di/dt$  is a function of drain current, the mathematical solution is complex. The MOSFET output capacitance also complicates the mathematics. And finally, MOSFETs have finite internal gate resistance which effectively adds to the resistance of the driving source, but the internal resistance is difficult to measure and, consequently, is not specified.

The resistive switching time variation versus gate resistance (Figure 9) shows how typical switching performance is affected by the parasitic circuit elements. If the parasitics were not present, the slope of the curves would maintain a value of unity regardless of the switching speed. The circuit used to obtain the data is constructed to minimize common inductance in the drain and gate circuit loops and is believed readily achievable with board mounted components. Most power electronic loads are inductive; the data in the figure is taken with a resistive load, which approximates an optimally snubbed inductive load. Power MOSFETs may be safely operated into an inductive load; however, snubbing reduces switching losses.

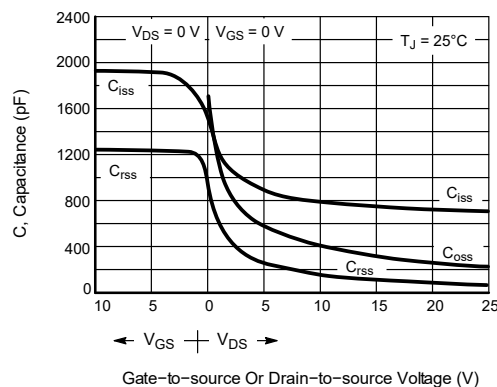
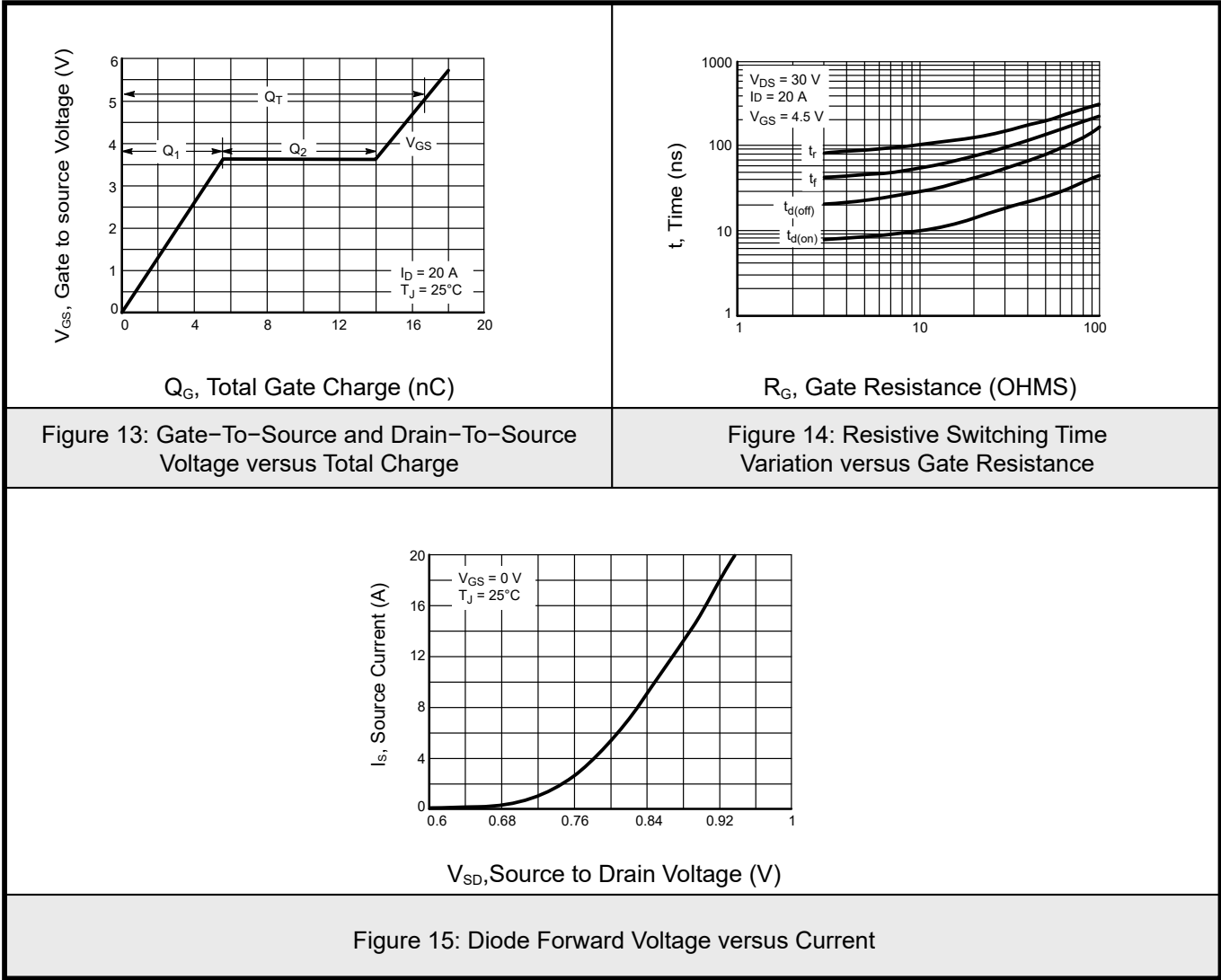


Figure 7. Capacitance Variation



6.3Typical Characteristics





## 6.4 Typical Characteristics

The Forward Biased Safe Operating Area curves define the maximum simultaneous drain-to-source voltage and drain current that a transistor can handle safely when it is forward biased. Curves are based upon maximum peak junction temperature and a case temperature ( $T_C$ ) of 25°C. Peak repetitive pulsed power limits are determined by using the thermal response data in conjunction with the procedures discussed in AN569, "Transient Thermal Resistance - General Data and Its Use."

Switching between the off-state and the on-state may traverse any load line provided neither rated peak current ( $I_{DM}$ ) nor rated voltage ( $V_{DSS}$ ) is exceeded and the transition time ( $t_{r,tf}$ ) do not exceed 10 s. In addition the total power averaged over a complete switching cycle must not exceed  $(T_{J(MAX)} - T_C)/(R_{JC})$ .

A Power MOSFET designated E -FET can be safely used in switching circuits with unclamped inductive loads. For

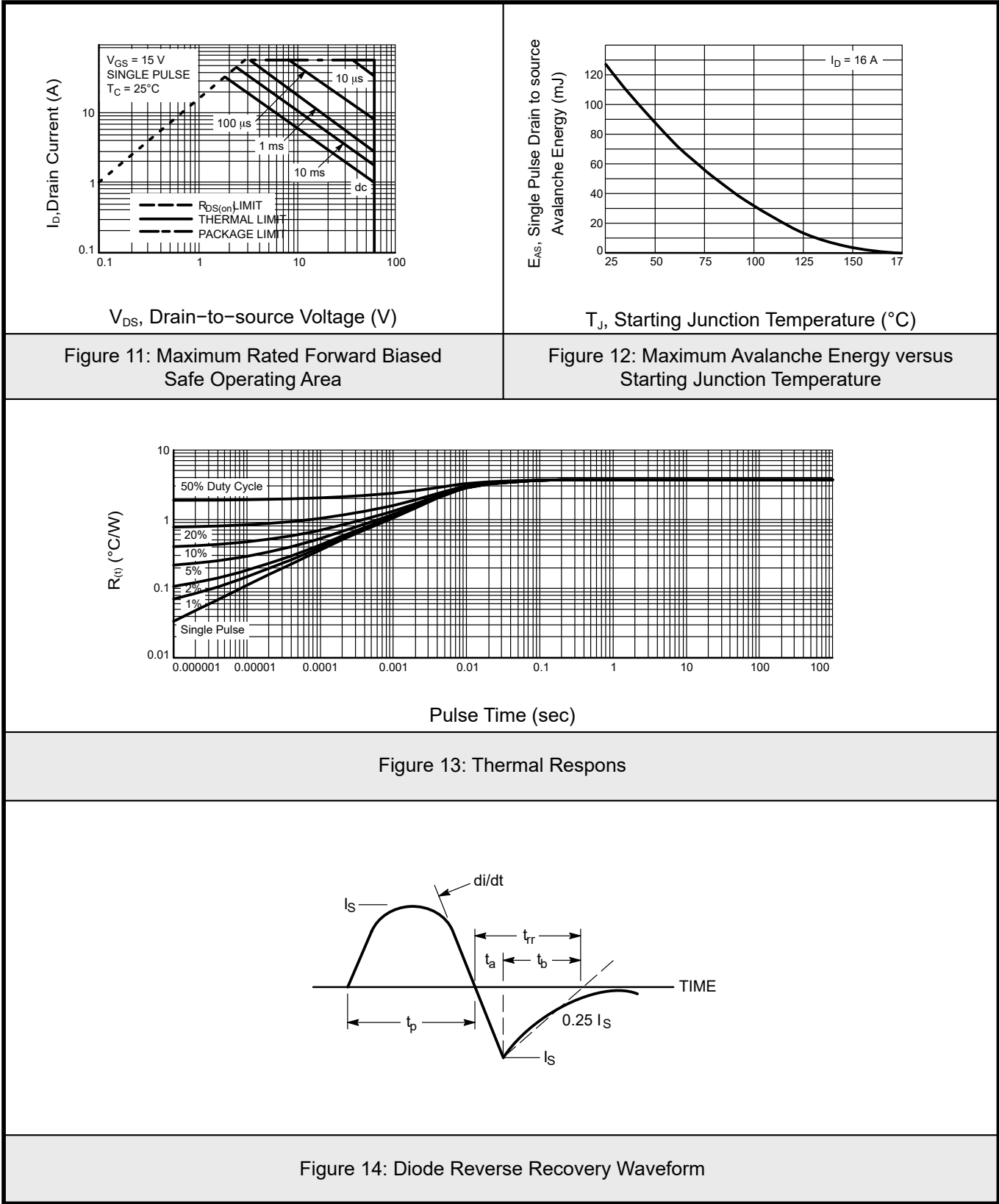
reliable operation, the stored energy from circuit inductance dissipated in the transistor while in avalanche must be less than the rated limit and adjusted for operating conditions differing from those specified. Although industry practice is to rate in terms of energy, avalanche energy capability is not a constant. The energy rating decreases non-linearly with an increase of peak current in avalanche and peak junction temperature.

Although many E-FETs can withstand the stress of drain-to-source avalanche at currents up to rated pulsed current ( $I_{DM}$ ), the energy rating is specified at rated continuous current ( $I_D$ ), in accordance with industry custom. The energy rating must be derated for temperature as shown in the accompanying graph (Figure 12). Maximum energy at currents below rated continuous  $I_D$  can safely be assumed to equal the values indicated.



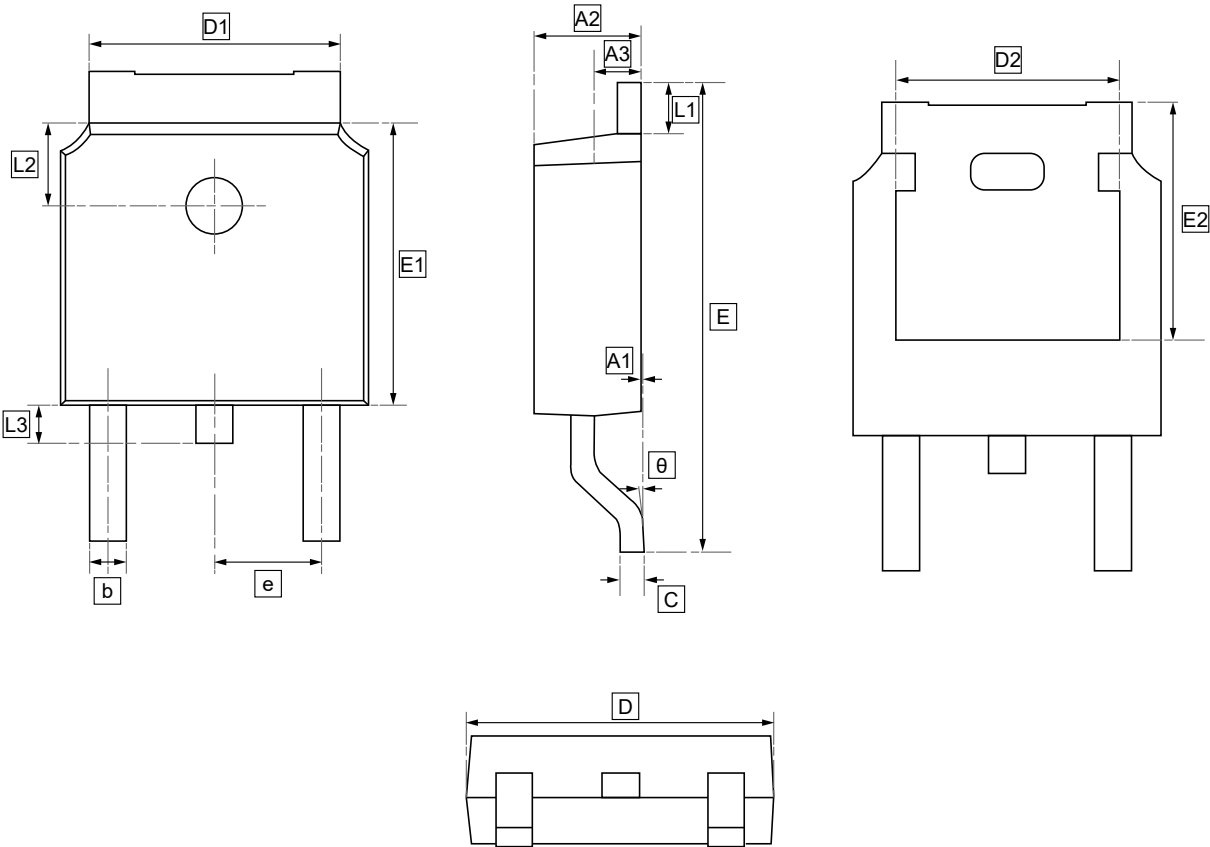


6.5Typical Characteristics





7.TO-252 Package Outline Dimensions

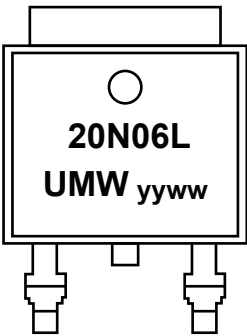


DIMENSIONS (mm are the original dimensions)

Symbol	A1	A2	A3	b	c	D	D1	D2	E	E1	E2	e	L1	L2	L3	θ
Min	0.00	2.18	0.90	0.65	0.46	6.35	4.95	4.32	9.40	5.97	5.21	2.286 BSC	0.89	1.70	0.60	0.00
Max	0.13	2.39	1.10	0.85	0.61	6.73	5.46	4.90	10.41	6.22	5.38		1.27	1.90	1.00	8.00



8.Ordering information



yy: Year Code  
ww: Week Code

Order Code	Package	Base QTY	Delivery Mode
UMW NTD20N06LT4G	TO-252	2500	Tape and reel



## 9.Disclaimer

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