4.5ns，Rail to Rail，High Speed Comparator

FEATURES

- ♦ High Speed: 4.5ns
- ♦ Rail to Rail I/O
- ♦ Supply Voltage: 1.8Vto 5.5V
- ♦ Push Pull CMOS Output Stage
- ♦ Shutdown
- ♦ Low Supply Current: 3.2mA
- ♦ 6-Pin SOT23 Package

APPLICATIONS

Test equipment

Wireless base stations

Threshold detectors

Zero crossing detectors

Window comparators

Test equipment

GENERAL INFORMATION

Ordering information

Part Number	Description	MPQ
TLV3501AIDBVR	SOT23-6L	3K

Package dissipation rating

Package	RθJA（℃/W）
SOT23-6L	200

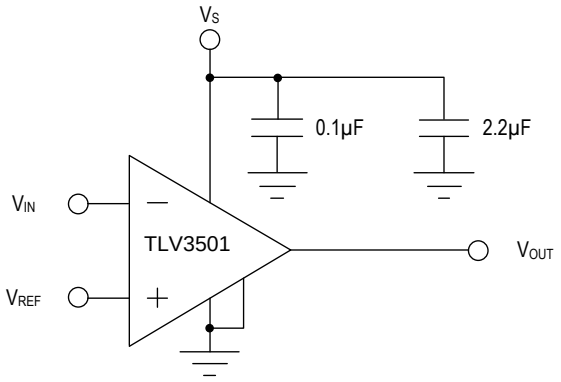
Absolute maximum ratings

Parameter	Value
Supply voltage	-0.7to 6V
IN+、IN- signal pin voltage	(V-)-0.3 to (V+)+0.3
IN+、IN- signal pin current	10mA max
OUT pin short circuit	70mA
Junction temperature	150℃
Storage temperature, Tstg	-55 to 150℃
Leading temperature （soldering，10secs）	260℃
ESD Susceptibility HBM	±2000V

Recommended operating condition

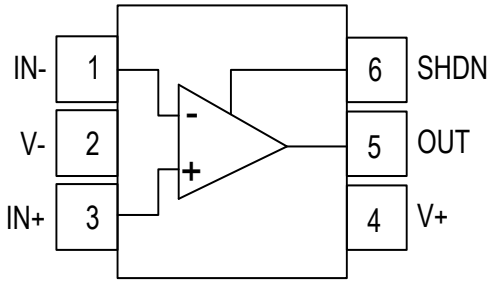
Symbol	Parameter	Range
Supply	Supply voltage	2.7-5.5V
Junction temperature		-40~125℃
PD	Power dissipation	0.50W

TYPICAL APPLICATION



Basic connections for the TLV3501AIDBVR

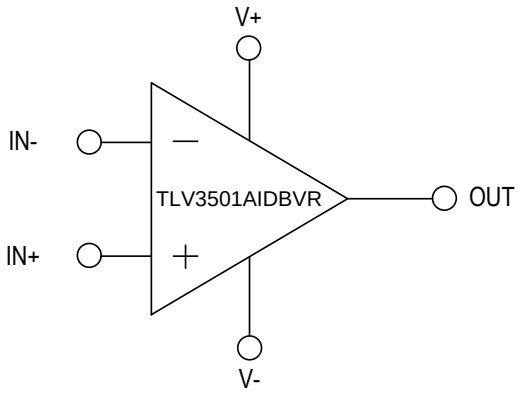
TERMINAL ASSIGMENTS



Pin information

PIN NO.	PIN name	Description
1	**IN-**	Negative (inverting) input
2	**V-**	Negative (lowest) power supply
3	**IN+**	Positive (noninverting) input
4	**V+**	Positive （highest）power supply
5	**OUT**	Output
6	**SHDN**	Shutdown (the device is idle when this pin is not in use)

BLOCK DIAGRAM



Electrical characteristics

（TA=25℃，VDD=2.7-5.5V, unless otherwise noted）

Symbol	Parameter	Test condition	Min	Typ.	Max	Unit
OFFSET VOLTAGE						
Input offset voltage	V_{OS}	VCM=0V，IO=0mA		±1	±6.5	mV
Input offset voltage vs temperature	dV_{OS}/d_T	TA=-40℃ to +125℃		±5		μV/℃
Input offset voltage vs power supply	PSRR	VS=1.8Vto 5.5V		100	400	μV/V
Input hysteresis				6		mV
INPUT BASIC CURENT						
Input basic current	I_B	VCM=VCC/2		±2	±10	pA
Input offset current	I_{OS}	VCM=VCC/2		±2	±10	pA
INPUT VOLTAGE RANGE						
Common mode voltage range			(V-)-0.2		(V+)-0.2	V
Common mode rejection ratio	CMRR	V_{CM}=-0.2V to (V+)+0.2V		57	70	dB
		T_A=-40℃ to +125℃		55		dB
INPUT IMPEDENCE						
Common mode				$10^{13}/2$		Ω/pF
differential				$10^{13}/4$		Ω/pF
OUTPUT						
Voltage output swing from rail	V_{OH}，V_{OL}	I_{OUT}=±1mA		30	50	mV
SHUTDOWN						
Shutdown turn-off time	T_{OFF}			30		ns
Shutdown turn-on time	T_{ON}			100		ns
SHDN low threshold	V_L	Comparator disabled			(V+)-1.7	V
SHDN high threshold	V_H	Comparator disabled	(V+)-0.9			V
Input bias current of shutdown pin				2		pA
Quiescent current in shutdown	I_{QSD}			2		μA
POWER SUPPLY						
Specified voltage	V_S		2.7		5.5	V
Operating voltage range		Higher end		1.8		V
		Lower end		5.5		V
Quiescent current	I_Q	V_S=5V，V_O=High		3.2	5	mA
SWITCHING CHARACTERISTICS						
Propagation delay time	TPD ΔVIN=100mV，	TA=25℃		4.5	6.5	ns
		TA=-40℃ to +125℃			7	ns

	overdrive=20mV					
	TPD ΔVIN=100mV， overdrive=5mV	TA=25℃		7.5	10	ns
		TA=-40℃ to +125℃			12	ns
Propagation delay skew	ΔT_{SKEW}	ΔVIN=100mV， overdrive=20mV		0.5		ns
Maximum toggle frequency	F_{max}	Overdrive=50mV，V_S=5V		80		MHz
Rise time	T_R			1.5		ns
Fall time	T_F			1.5		ns

Characteristic plots

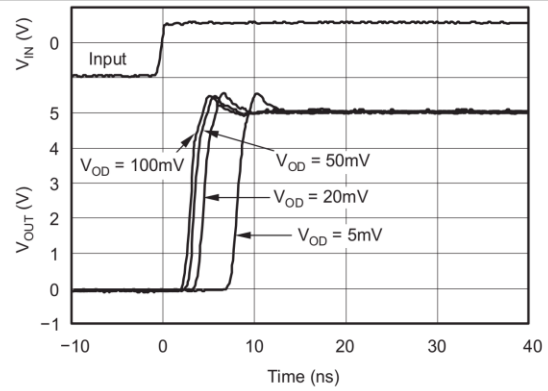


Figure 1. Output Response for Various Overdrive Voltages (Rising)

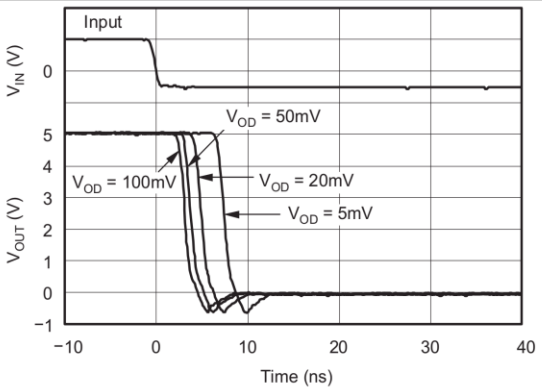
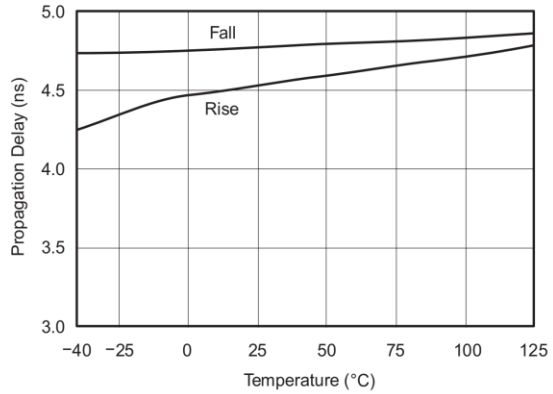
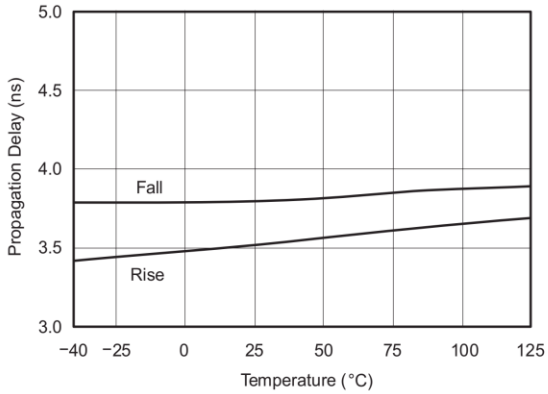


Figure 2. Output Response for Various Overdrive Voltage (Falling)



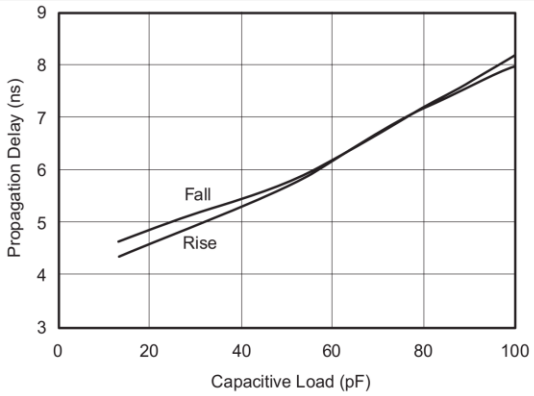
V_{OD} = 20 mV

Figure 3. Propagation Delay vs Temperature



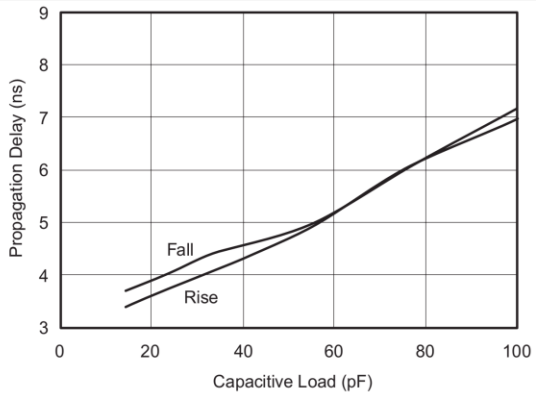
V_{OD} = 50 mV

Figure 4. Propagation Delay vs Temperature



V_{OD} = 20 mV

Figure 5. Propagation Delay vs Capacitive Load



V_{OD} = 50 mV

Figure 6. Propagation Delay vs Capacitive Load

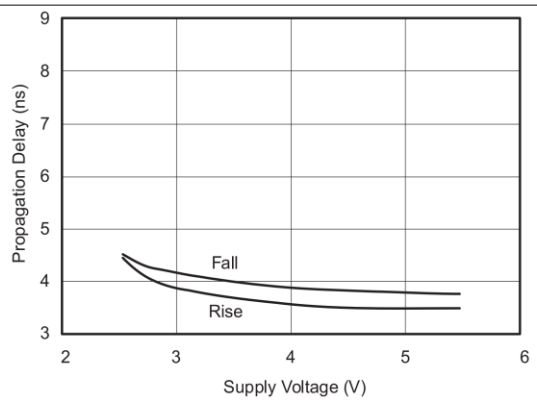


Figure 7. Propagation Delay vs Supply Voltage

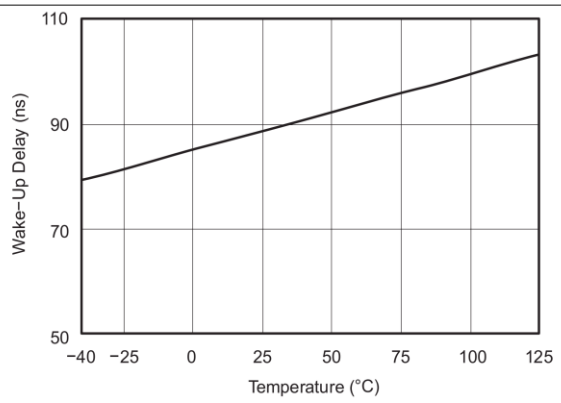


Figure 8. Wake-Up Delay vs Temperature

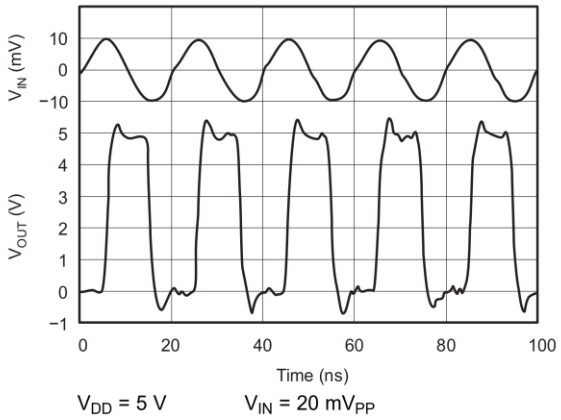


Figure 9. Response to 50-MHz Sine Wave

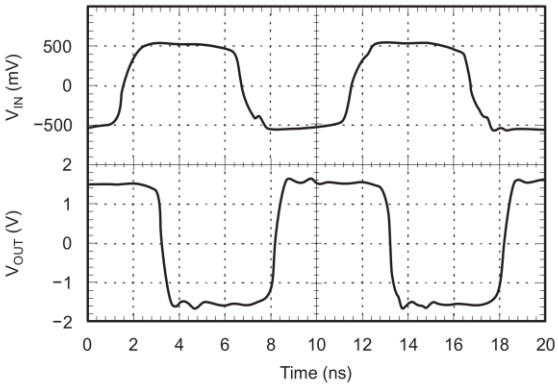


Figure 10. Response to 100-MHz Sine Wave (±2.5-V Dual Supply into 50-Ω Oscilloscope Input)

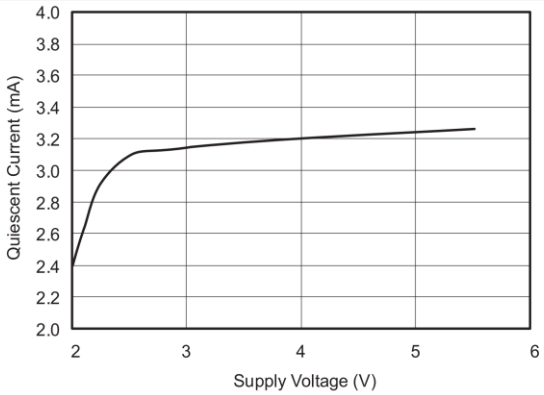


Figure 11. Quiescent Current vs Supply Voltage

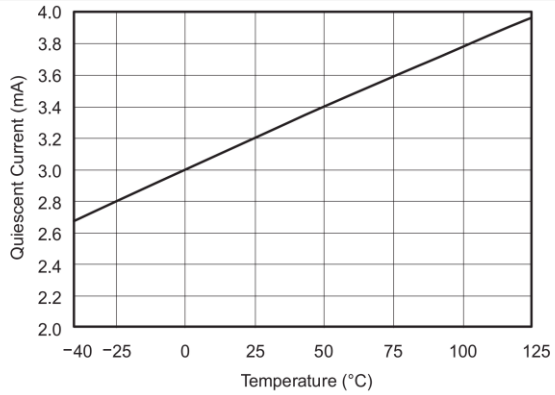


Figure 12. Quiescent Current vs Temperature

Operation description

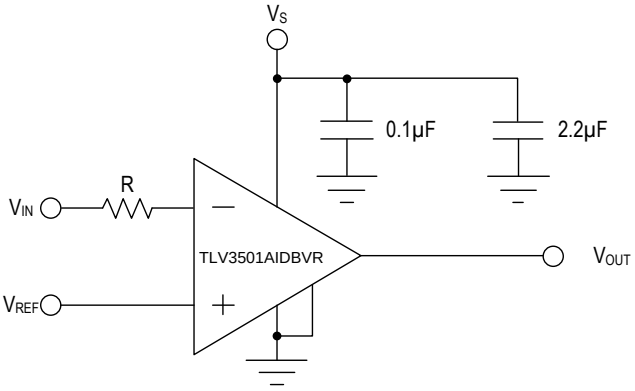
The TLV3501XXX device both feature high speed and include 6mV of internal hysteresis for improved noise immunity with an input common mode range that extends 0.2Vbeyond the power supply rails.

Operating Voltage

The TLV3501XX comparator is specified for use on a single supply from 1.8Vto 5.5V (or a dual supply from ±1.35V to ±2.75V) over a temperature range of -40℃ to +125℃. This device continue to function below this range, but performance is not specified.

Input Over voltage Protection

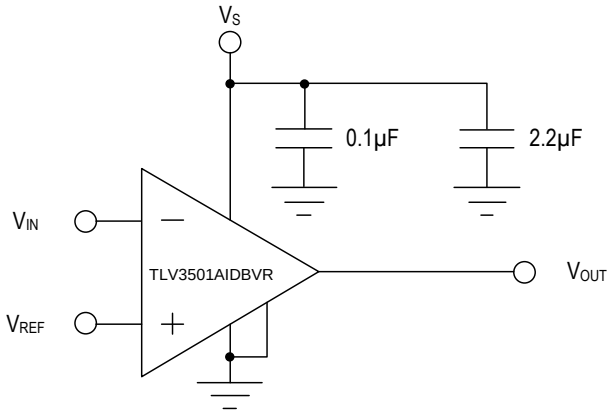
Device inputs are protected by electrostatic discharge diode that conduct if the input voltages exceed the power supplies by more than approximately 300mV. Momentary voltages greater than 300mV beyond the power supply can be tolerated if the input current is limited to 10mA. This limiting is easily accomplished with a small input resistor in series with the comparator, as shown in the below figure.



Input current protection for voltage exceeding the supply voltage

Shutdown

A shutdown pin allows the device to go into idle when it is not in use. When the shutdown pin is high, the device draws approximately 2uA, and the output goes to high impedance. When the shutdown pin is low, the TLV3501xx is active. When the TLV3501xx shutdown feature is not used, connect the shutdown pin to the most negative supply, as shown in the below figure. Exiting shutdown mode requires approximately 100ns.



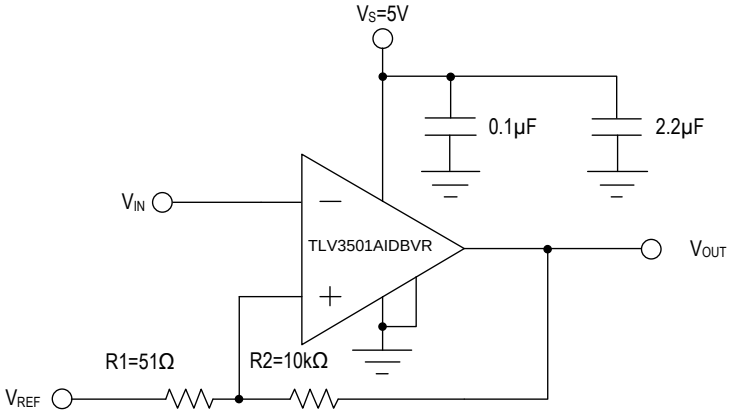
Basic connection for the TLV3501AIDBVR

Application Information

The TLV3501xx has a robust performance when used with a good layout. However, comparator inputs have little noise immunity within the range of a specified offset voltage. For slow moving or noisy input signals, the comparator output can cause an undesirable switch state as input signals move through the switching threshold. In such applications, the 6mV of internal hysteresis of the TLV3501xx might not be sufficient. For greater noise immunity, external hysteresis can be added by connecting a small amount of feedback to the positive input. The next figure shows a typical topology used to introduce 25mV of additional hysteresis, for a total of 31mV hysteresis when operating from a single 5V supply. Use the following equation to calculate the approximate total hysteresis.

$$V_{HYST} = \frac{(V+) \times R_1}{R_1 + R_2} + 6mV$$

The total hysteresis sets the value of the transition voltage required to switch the comparator output, by enlarging the threshold region, thereby reducing sensitivity to noise.



Power Supply Recommendations

Place bypass capacitors close to the power supply pins to reduce noise coupling in from noisy or high impedance power supplies. For more information on bypass capacitor placement, see Layout Guidelines.

Layout Guidelines

For any high speed comparator or amplifier, proper design and printed circuit board layout are necessary for optimal performance. Excess stray capacitance on the active input, or improper grounding, can limit the maximum performance of high speed circuitry.

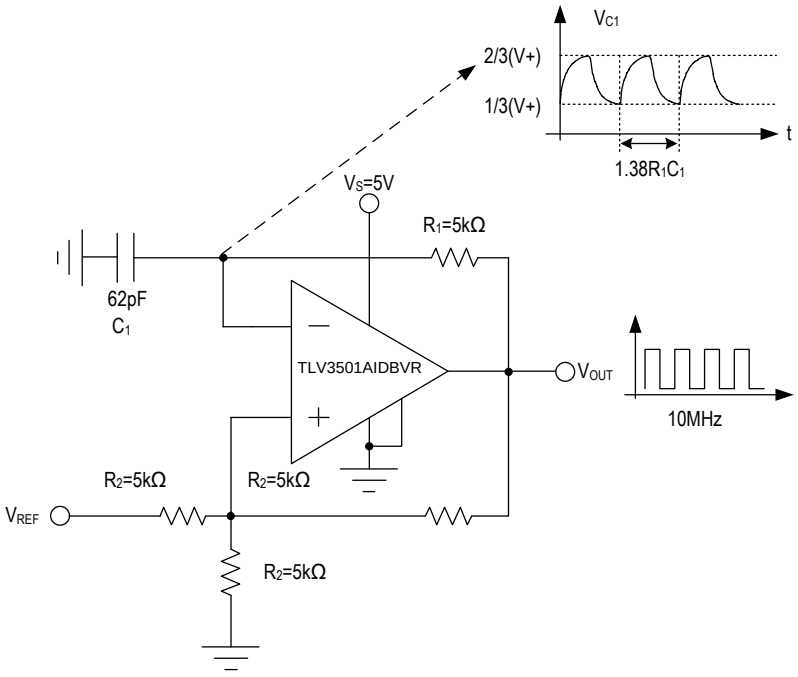
Minimizing resistance from source to the comparator input is necessary to minimize the propagation delay of the complete circuit. The source resistance, along with input and stray capacitance, creates an RC filter that delays voltage transitions at the input, and reduces the amplitude of high frequency signals. The input capacitance of the TLV3501xx, along with stray capacitance from an input p in to ground, results in several picofarads of capacitance.

The location and type of capacitance used for power supply bypassing are critical to high speed comparators. The suggested 2.2uF tantalum capacitor does not need to be as close to the device as the 0.1uF capacitor, and may be shared with other devices. The 2.2uF capacitor buffers the power supply line against ripple, and the 0.1uFcapacitor provides a charge for the comparator during high frequency switching.

In a high speed circuit, fast rising and falling switching transients create voltage differences across lines that would be at the same potential at DC. To reduce this effect, use a ground plane to reduce difference in voltage potential within the circuit board. A ground plane has the advantage of minimizing the effect of stray capacitances on the circuit board by proving a more desirable path for the current to flow. With a signal trace over a ground plane, at high frequency the return current (in the ground plane) tends to flow right under the signal trace. Breaks in the ground plane (as simple as through hole leads and vias) in crease the inductance of the plane, making it less effective at higher frequencies. Breaks in the ground plane for necessary vias must be spaces randomly.

Typical application

The TLV3501xx can easily be configured as a simple and in expensive relaxation oscillator. In the next figure, the R_2 network sets the trip threshold at 1/3 and 2/3 of the supply. Because this circuit is a high speed circuit, the resistor values are low to minimize the effects of parasitic capacitance. The positive input alternates between 1/3 of V+ and 2/3 of V+， depending on whether the output is low or high. The time to charge (or discharge) is $0.69\times R_1C_1$. Therefore, the period is $1.38\times R_1C_1$. For 62pF and 1kohm as shown in the next figure, the output is calculated to 10.9MHz. An implementation of this circuit oscillator at 9.6MHz. Parasitic capacitance and component tolerance explain the difference between theory and actual performance.



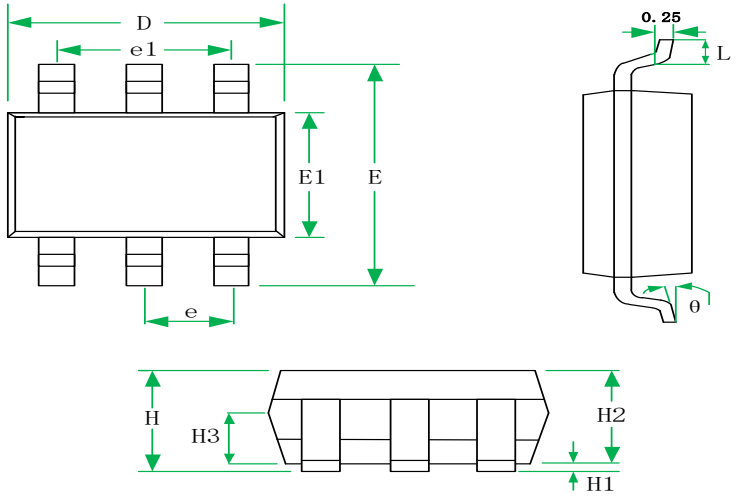
Relaxation Oscillator

For hysteresis of 1/3 of V+ and threshold levels between 1/3 of V+ and 2/3 of V+, the resistors connected to the comparator positive input must be equal in value. The resistor value must be kept low enough so it does not create additional time constant because of the input capacitor and board parasitic capacitor. The value of the charging resistor R_1must be relatively low for high frequency switching without drawing high current and affecting the output high and low level. The value of the charging capacitor must be high enough to avoid errors cause by parasitic capacitance.

For the positive input, $IN+=1/3V_{OUT}+1/3V+=1/3V+$, if V_{OUT} is low and assuming V_{OL} is very close to GND. Or, $IN+=1/3V_{OUT}+1/3V+=1/3V+=2/3V+$, if V_{OUT} is high and assuming V_{OH} is very close to V+.

For the negative input, the capacitor charges to 2/3V+ and discharges to 1/3 V+ exponentially at the same rate with a time constant of R_1C_1.

Package information



SYMBOL	MILLIMETERS			INCHES		
	MIN	NOM	MAX	MIN	NOM	MAX
H			1.45			0.057
H1	0.04		0.15	0.0016		0.0059
H2	1.00	1.10	1.20	0.039	0.043	0.047
H3	0.55	0.65	0.75	0.022	0.026	0.029
D	2.72	2.92	3.12	0.107	0.115	0.123
E	2.60	2.80	3.00	0.102	0.110	0.118
E1	1.40	1.60	1.80	0.055	0.063	0.071
e	0.95BSC			0.037BSC		
e1	1.90BSC			0.074BSC		
L	0.30		0.60	0.012		0.024
θ	0		8°	0		8°

SOT23-6 for TLV3501AIDBVR