



钰地半导体  
Tudi Semiconductor

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## Product Specification

TUDI-AD8656

28MHz, Low noise, precision CMOS amplifier

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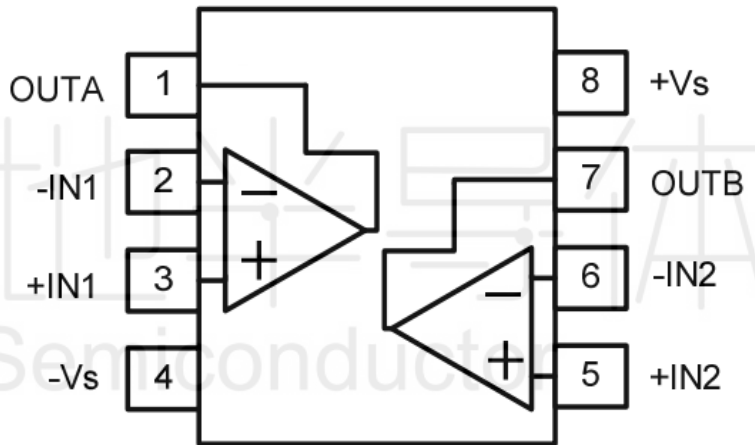
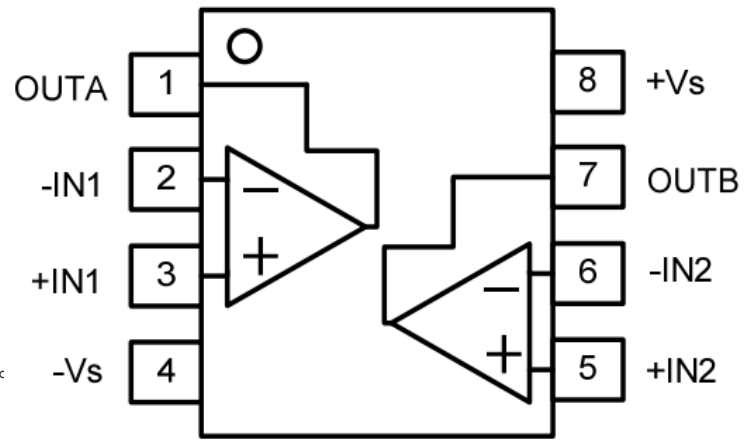


## Features

- Low noise: 2.7 nV/√Hz (f = 10 kHz)
- Low offset voltage: maximum 250 μV VCM
- Bandwidth: 28 MHz
- Rail-to-rail input/output
- Gain stable at unity
- Operating voltage range: 2.7 V to 5.5
- Operating temperature range: -40°C to 125°C

## Applications

- ADC and DAC buffer;
- audio;
- industrial control;
- precision filter;
- digital scale;
- PLL filter.



SOP8/MSOP8

## Description

The AD8656 is a precision CMOS amplifier with the lowest noise in its class.

The AD8656 provides low for low voltage applications. In addition, it has rail-to-rail input and output swing capability, allowing the designer to buffer an analog-to-digital converter (ADC) other wide dynamic range devices in single supply systems.

In low voltage applications, the high precision performance of the AD8656 can improve resolution and dynamic range. Audio applications as microphone preamplifiers and mixing consoles benefit from the AD8656's low noise, low distortion, and high output current capability to reduce system-level noise ensure audio fidelity.

The high precision and the rail-to-rail input and output of the AD8656 benefit data acquisition, process control, and PLL filter.

The AD8656 is rated for an operating temperature range of -40°C to 125°C.

The AD8656 available in lead-free, 8-lead MSOP and SOIC packages.



## Pin Functions

Name	Description	Note
+Vs	Positive power supply	A bypass capacitor of 0.1 $\mu$ F as close to the part as possible should be placed between power supply pins or between supply pins and ground.
-Vs	Negative power supply or ground	If it is not connected to ground, bypass it with a capacitor of 0.1 $\mu$ F as close to the part as possible
-IN	Negative input	Inverting input of the amplifier. Voltage range of this pin can go from -Vs-0.3V to +Vs-1V.
+IN	Positive input	Non-inverting input of the amplifier. This pin has the same voltage range as -IN.
OUT	Output	The output voltage range extends to within millivolts of each supply rail.
NC	No connection	

## Product Specification

### Recommended Operating Conditions

Parameter	Rating	Unit
DC Supply Voltage	2.7V-5.5V	V
Input common-mode voltage range	-Vs ~ +Vs	V
Operating ambient temperature	-40~125	$^{\circ}$ C

### Thermal Data

Parameter	Rating	Unit
Package Thermal Resistance	206(MSOP8) 155(SOP8) 125(DIP8) 105(TSSOP8)	$^{\circ}$ C/W



### Electrical Characteristics

(+Vs=+5V,-Vs=0,VeM=Vs/2,TA=+25°C,RL=10kΩto Vs/2,unless otherwise noted)

Parameter	Symbol	Conditions	Min	Typ	Max	Unit
<b>Out Characteristics</b>						
Output Voltage Swing from Rail		RL=100kΩ		1		mV
		RL=10kΩ		8		mV
		RL=2kΩ		40		mV
Short-Circuit Current	IoUT	Sourcing		21		mA
		Sinking		22		mA
<b>Input Characteristics</b>						
Input Offset Voltage	Vos			±0.6	±5	mV
Input Offset Voltage Drift	ΔVos/ΔT	-40 to 125°C		±2.0		μV°C
Input Bias Current	IB			±2.5		pA
Input Offset Current	Ios			±2.5		pA
Common-Mode Voltage Range	VcM	Vs =5.5V	-0.1		5.6	V
Common-Mode Rejection Ratio	CMRR	VcM=0.1V to 4.9V		125		dB
Open-Loop Voltage Gain	AOL	Vo=0.2V to 4.8V		120		dB
<b>Dynamic Performance</b>						
Gain Bandwidth Product	GBWP	G=+1		28		MHz
Slew Rate	SR	G=+1,2V Output Step		7.5		V/μs
<b>Noise Performance</b>						
Voltage Noise Density	en	f=1kHz		12		nV/√Hz
<b>Power Supply</b>						
Operating Voltage Range			2.7		5.5	V
Power Supply Rejection Ratio	PSRR	VS = 2.7V-5.5V	80	100		dB
Quiescent Current /Amplifier	Io			650		μA



### Absolute Maximum Ratings

Parameter	Rating	Units
Power Supply:+Vs to-Vs	6.0	V
Input Voltage	-Vs -0.5V to+Vs+0.5V	V
Input Current(2)	10	mA
Storage Temperature Range	-65 to 150	°C
Junction Temperature	150	°C
Operating Temperature Range	-40 to 125	°C
ESD Susceptibility,HBM	2000	V

(1) Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

(2) Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails should be current-limited to 10mA or less.

## Application Notes

### Driving Capacitive Loads

Driving large capacitive loads can cause stability problems for voltage feedback op amps. As the load capacitance increases, the feedback loop's phase margin decreases, and the closed loop bandwidth is reduced. This produces gain peaking in the frequency response, with overshoot and ringing in the step response. A unity gain buffer ( $G = +1$ ) is the most sensitive to capacitive loads, but all gains show the same general behavior.

When driving large capacitive loads with these op amps (e.g.,  $> 100$  pF when  $G = +1$ ), a small series resistor at the output ( $R_{iso}$  in Figure 1) improves the feedback loop's phase margin (stability) by making the output load resistive at higher frequencies. It does not, however, improve the bandwidth. To select  $R_{iso}$ , check the frequency response peaking (or step response overshoot) on the bench. If the response is reasonable, you do not need  $R_{iso}$ . Otherwise, start  $R_{iso}$  at 1k and modify its value until the response is reasonable.

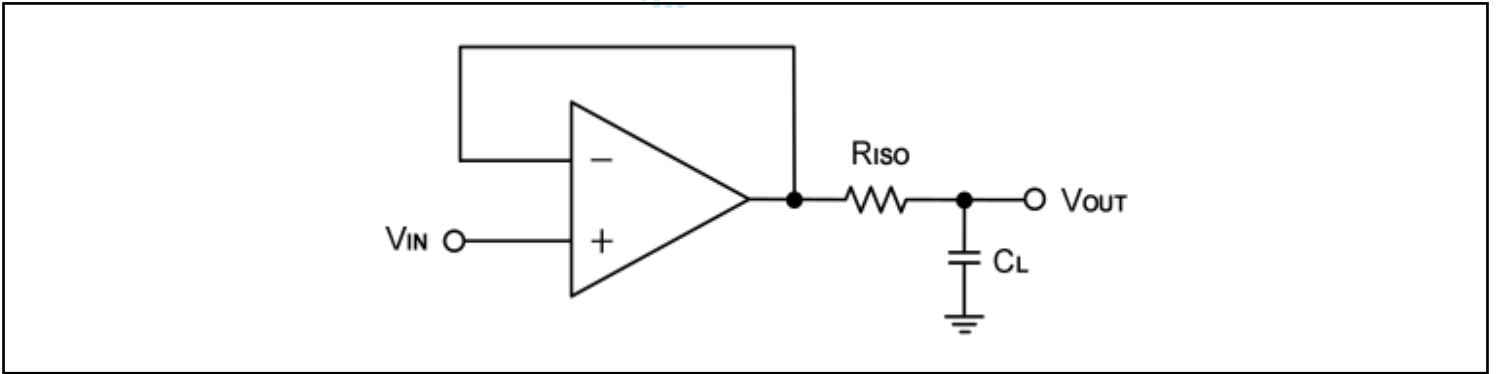


Figure 1. Indirectly Driving Heavy Capacitive Load

An improvement circuit is shown in Figure 2. It provides DC accuracy as well as AC stability.  $R_F$  provides the DC accuracy by connecting the inverting signal with the output,  $C_F$  and  $R_{iso}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving phase margin in the overall feedback loop.

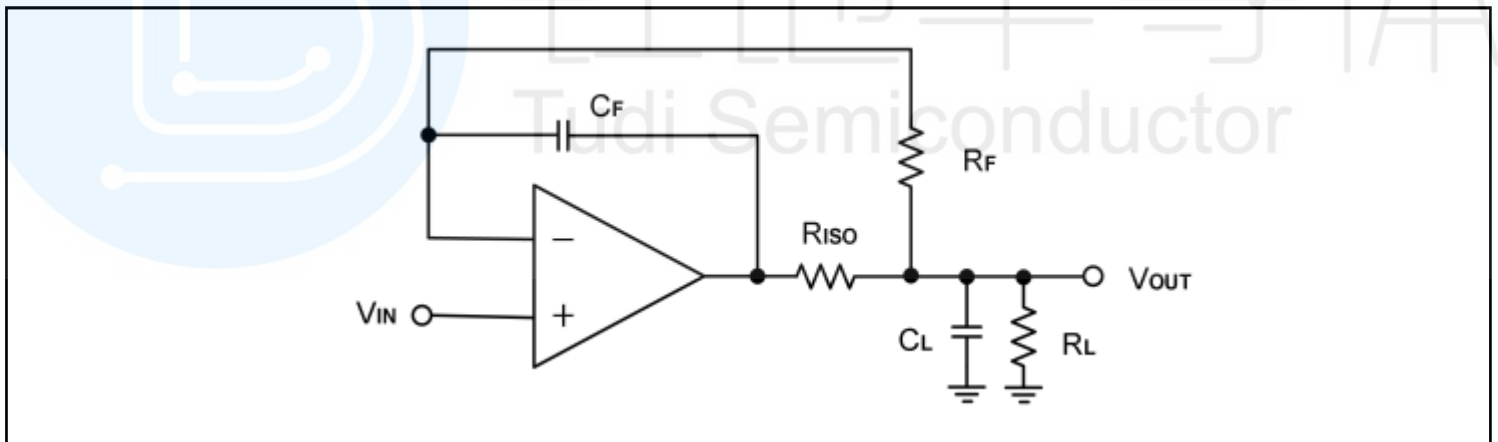


Figure 2. Indirectly Driving Heavy Capacitive Load with DC Accuracy

For non-inverting configuration, there are two others ways to increase the phase margin:  
(a) by increasing the amplifier's gain or  
(b) by placing a capacitor in parallel with the feedback resistor to counteract the parasitic capacitance associated with inverting node, as shown in Figure 3.

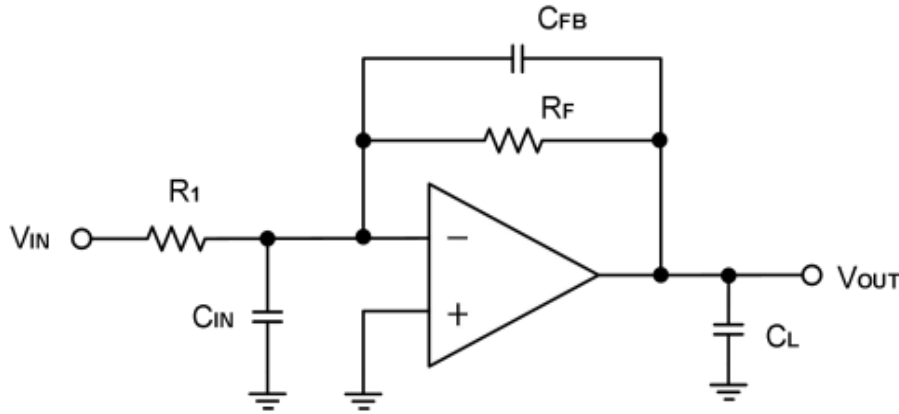


Figure 3. Adding a Feedback Capacitor in the Non-inverting Configuration

### Power-Supply Bypassing and Layout

The 8656 operates from a single +2.7V to +5.5V supply or dual  $\pm 1.05\text{V}$  to  $\pm 2.75\text{V}$  supplies. For single-supply operation, bypass the power supply +Vs with a  $0.1\mu\text{F}$  ceramic capacitor which should be placed close to the +Vs pin. For dual-supply operation, both the +Vs and the -Vs supplies should be bypassed to ground with separate  $0.1\mu\text{F}$  ceramic capacitors.  $2.2\mu\text{F}$  tantalum capacitor can be added for better performance.

The length of the current path is directly proportional to the magnitude of parasitic inductances and thus the high frequency impedance of the path. High speed currents in an inductive ground return create an unwanted voltage noise. Broad ground plane areas will reduce the parasitic inductance. Thus a ground plane layer is important for high speed circuit design.

### Typical Application Circuits

#### Differential Amplifier

The circuit shown in Figure 4 performs the differential function. If the resistors ratios are equal ( $R_4 / R_3 = R_2 / R_1$ ), then  $V_{OUT} = (V_{IP} - V_{IN}) \times R_2 / R_1 + V_{REF}$ .

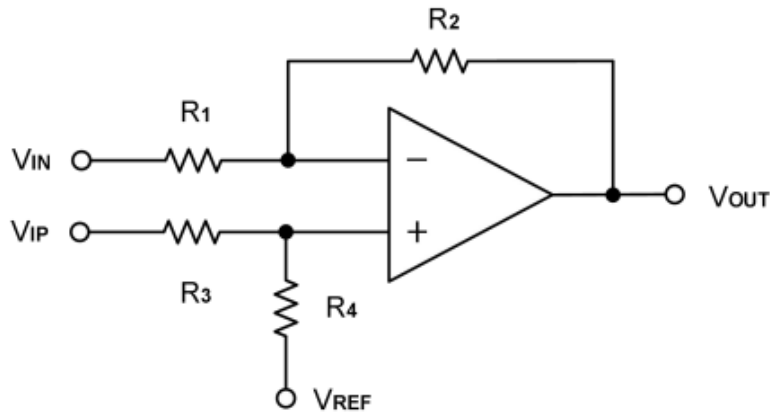


Figure 4. Differential Amplifier

### Low Pass Active Filter

When receiving low-level signals, limiting the bandwidth of the incoming signals into the system is often required. The simplest way to establish this limited bandwidth is to place an RC filter at the noninverting terminal of the amplifier. If even more attenuation is needed, a multiple pole filter is required. The Sallen-Key filter can be used for this task, as Figure 5. For best results, the amplifier should have a bandwidth that is 8 to 10 times the filter frequency bandwidth. Failure to follow this guideline can result in reduction of phase margin. The large values of feedback resistors can couple with parasitic capacitance and cause undesired effects such as ringing or oscillation in high-speed amplifiers. Keep resistors value as low as possible and consistent with output loading consideration.

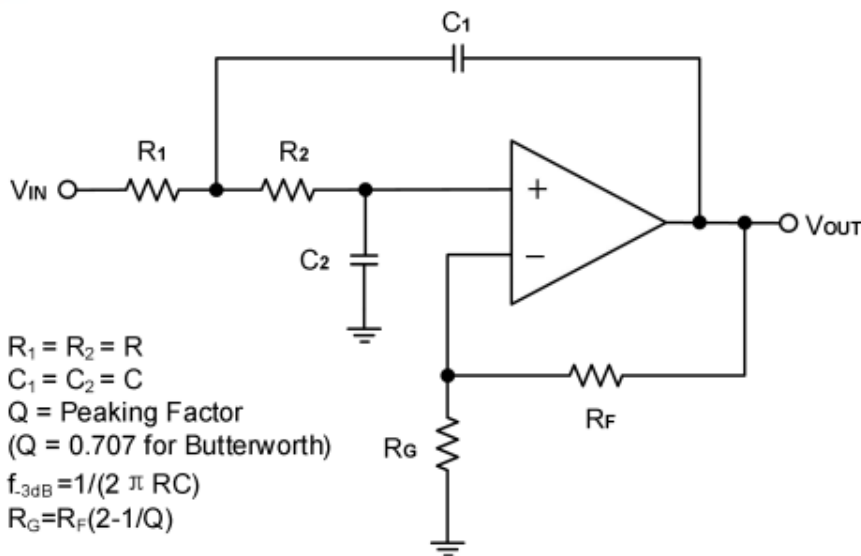
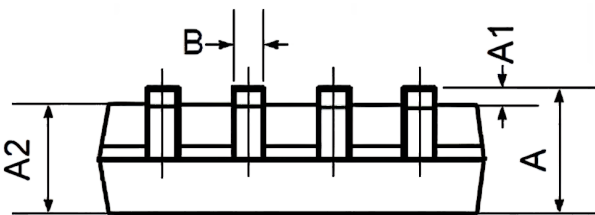
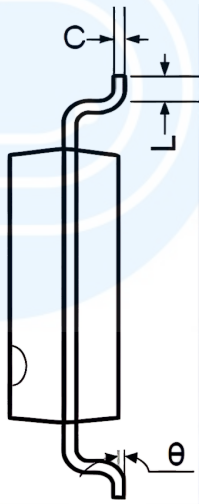
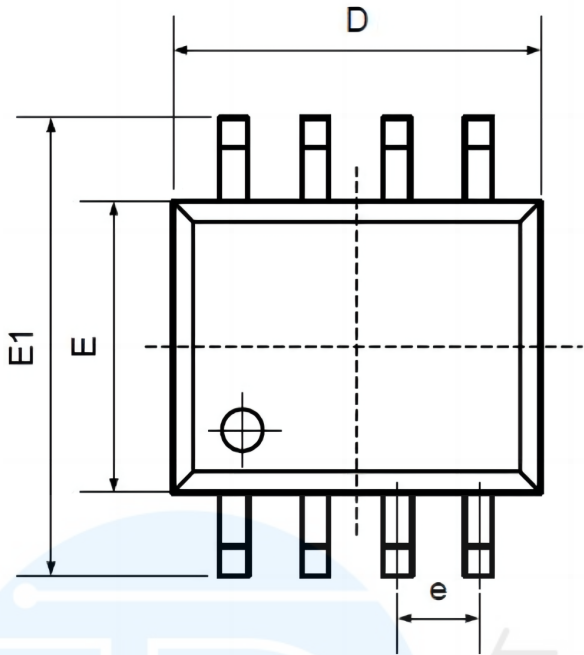


Figure 5. Two-Pole Low-Pass Sallen-Key Active Filter



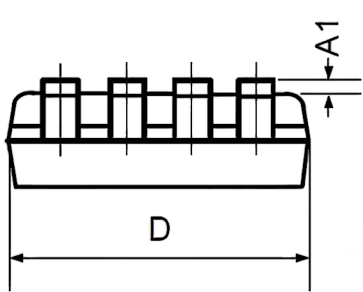
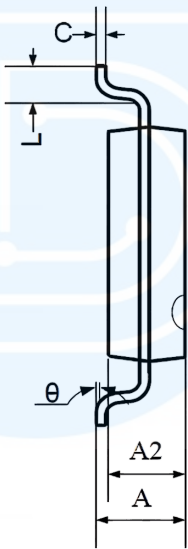
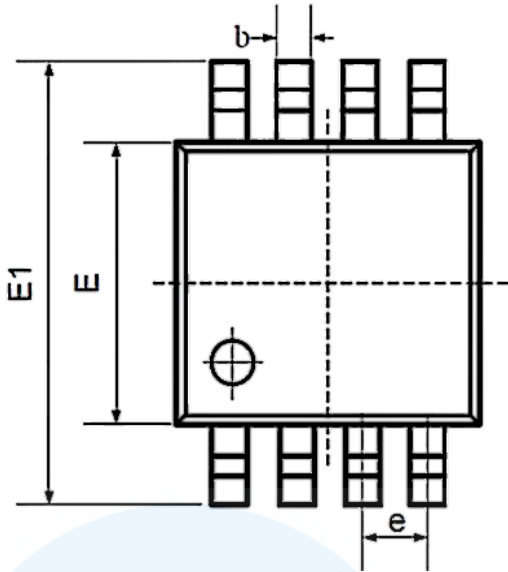
Package SOP8



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	1.350	1.750	0.053	0.069
A1	0.100	0.250	0.004	0.010
A2	1.350	1.550	0.053	0.061
B	0.330	0.510	0.013	0.020
C	0.190	0.250	0.007	0.010
D	4.780	5.000	0.188	0.197
E	3.800	4.000	0.150	0.157
E1	5.800	6.300	0.228	0.248
e	1.270TYP		0.050TYP	
L	0.400	1.270	0.016	0.050
θ	0°	8°	0°	8°



Package MSOP8



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	0.800	1.200	0.031	0.047
A1	0.000	0.200	0.000	0.008
A2	0.760	0.970	0.030	0.038
b	0.30 TYP		0.012 TYP	
C	0.15 TYP		0.006 TYP	
D	2.900	3.100	0.114	0.122
e	0.65 TYP		0.026 TYP	
E	2.900	3.100	0.114	0.122
E1	4.700	5.100	0.185	0.201
L	0.410	0.650	0.016	0.026
θ	0°	6°	0°	6°



## Order information

Order Number	Package	Package Quantity	Marking On The park	Temperature
AD8656ARZ-TUDI	SOP8	Tape,Reel,2500	AD8656A	- 40°C to 125°C
AD8656ARMZ-TUDI	MSOP8	Tape,Reel,2500	AOS	



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